

Appendix A Example Planar Overview

A.1 The Collateral Material

The information in these appendixes are for reference only, and consists of example planar design information, example firmware, and additional information. It is believed to be correct, but is presented as is, and may contain errors. It is intended to help companies develop their own products using the MCM.

The example planar, along with the firmware and additional information, can be used:

- As a development system.
- To test new boot code, operating systems and/or applications.
- For performance measuring.

The example planar and firmware are:

- Tested for functionality to the level of software internal to IBM as of 12/31/96.
- Prototypes, which may contain prototype ASICs, errata, and/or wiring changes. The example planar may also have errata.

The example planar and firmware are not:

- A fully debugged, complete, and market ready design
- Tested for compliance to FCC and other regulatory requirements
- Intended to be used as a product.
- Sold by IBM.

A.2 Example Planar

In order to offer more robust support to the MCM user, IBM has designed an example planar for the MCM (see Figure A-1). The example planar is described (schematics, BOM, mechanical drawings, and functional information) in these appendixes. The example planar is a working design that is in use internally at IBM, and that was used in the design and development of the MCM.

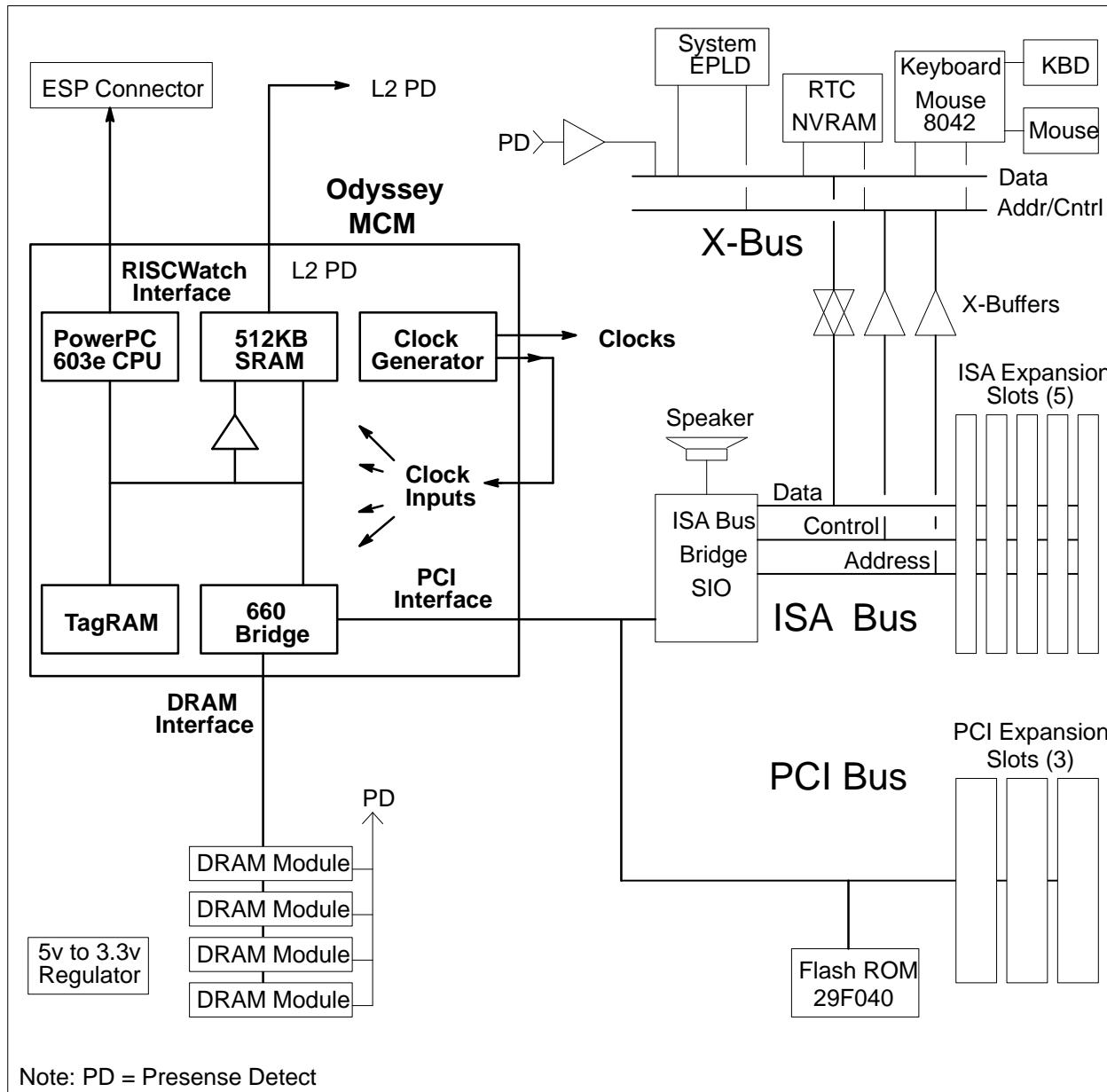


Figure A-1. 100 MHz PPC 603e Example Planar Block Diagram

The example planar is designed to an industry standard BabyAT (8.6 in. by 13 in.) form factor, although the enclosure may need to be modified to provide additional cooling and/or additional space for I/O cards. It requires +5V to power most of the components. The motherboard also requires ± 12 V to support some of the peripheral features. Components

that require +3.3V (such as the PCI bus agents) are supported using a regulator mounted on the motherboard to convert +5V to +3.3V. It is sized to fit within a BabyAT form factor enclosure.

A.2.1 System Memory

The example planar memory subsystem can support up to 128M of DRAM memory on four 72 pin modules via sockets. Each SIMM socket can support an 8M or 32M 72 pin SIMM. The DRAM subsystem is 72 bits wide: 64 data bits and eight parity bits. One parity bit is generated for each byte of data written. The MCM can also be configured to perform ECC memory data checking and correction using the same standard parity DRAM modules. Or it can be configured to disable DRAM parity checking for systems using non-parity DRAM.

A.2.2 PCI Bus

The MCM includes the interface between the PCI bus and the rest of the system. The example planar provides three standard PCI expansion slots.

A.2.3 Flash ROM

The example planar uses an AMD AM29F040-120 Flash™ ROM to contain the POST and boot code. Vital Product Data (VPD) such as the motherboard speed and native I/O complement be programmed into in this device. It is possible to program the Flash before or during the manufacturing process.

A.2.4 ISA Bus

The ISA bridge function is provided by an Intel 82378ZB chip (SIO). It provides a PCI to ISA bus bridge where the native I/O and the ISA slots reside, and it provides system services such as ISA bus DMA, PCI bus arbitration, and interrupt control. The example planar provides five PCI expansion slots.

A.2.5 System I/O EPLD

The system I/O EPLD is a programmable logic device that uses the X-bus signals and the partial decode signals from the SIO to decode chip selects for various components.

A.2.6 X-Bus Agents

The example planar uses a Dallas Semiconductor™ DS1385S to provide the real time clock (TOD or RTC) function. This device is PC compatible and resides on the X-bus. It features an additional 4K of NVRAM and a replaceable battery.

The example planar uses an Intel 8042AH as a keyboard and mouse controller.

A.2.7 Power Management

Some components are included on the example planar which can be used for power management hardware. Power management is not implemented on the example planar, and is not intended to be a feature of the example planar.

A.3 Other Help

Several other system, firmware, and miscellaneous issues of interest to the designer are included here. For more information, refer to the IBM PowerPC World Wide Web pages at <http://www.chips.ibm.com> and ftp://ftp.austin.ibm.com/pub/PPC_support. Also contact your IBM PowerPC Embedded Processor Solutions Technical Representative for details.

Consult specific manufacturer technical specifications for information on specific I/O devices such as hard drives, CD-ROMs, L2 cache cards, video cards, etc. that comprise a total system.

A.3.1 Example Firmware

IBM has developed firmware to support the MCM and example planar. The source code and the binary of this boot & POST firmware is available from IBM. Recommendations in this document for memory mappings, and various software issues may vary from the algorithms implemented in boot code or operating systems.

A.3.2 Design Files

The Gerber format physical design files of the example planar are available from IBM.

A.3.3 Quickstart Peripheral List

The example planar is intended to use typical PC peripherals. Products from a large number of manufacturers should work satisfactorily.

Table A-1 outlines generic requirements for peripherals and gives examples of some devices that have been used for testing. It is not a recommendation of any particular vendor. Table A-1 does not include cables for a parallel port, indicators, a switch, or a speaker.

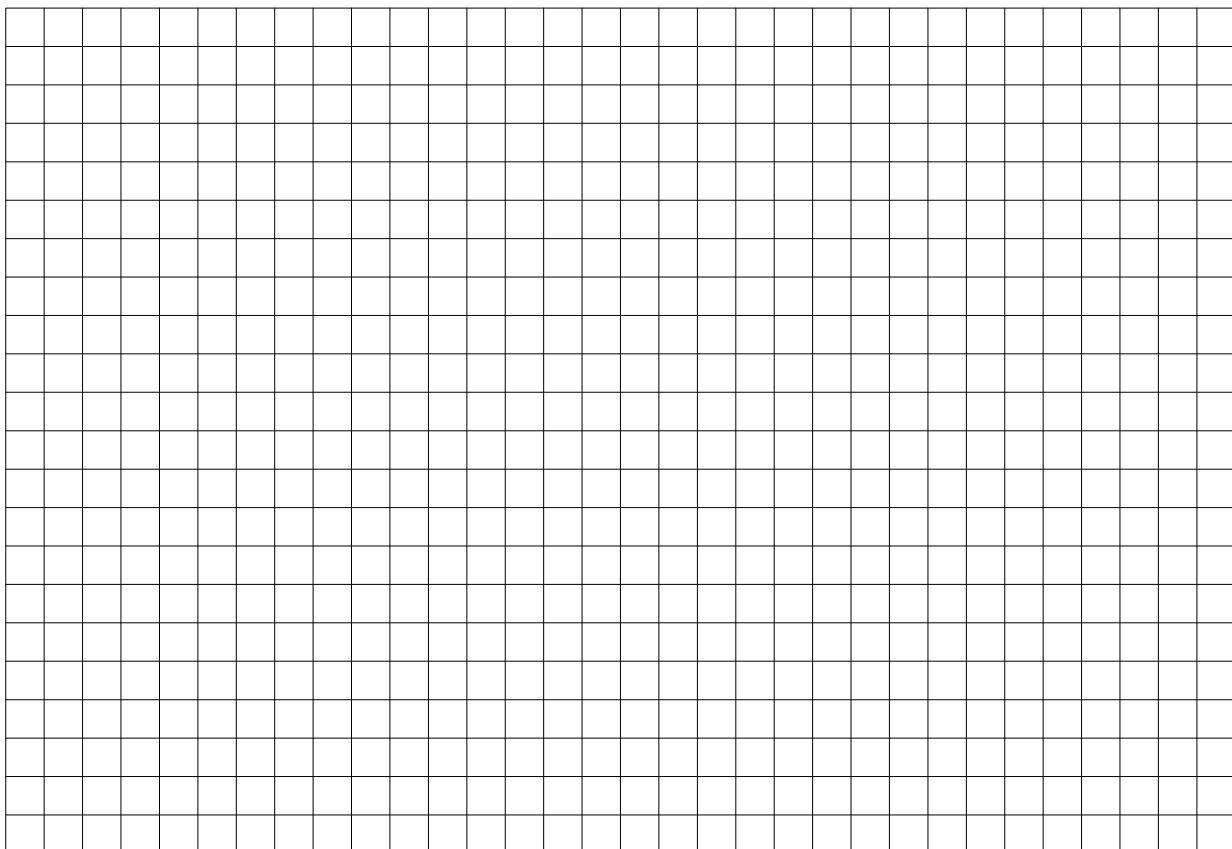
An IBM 3101 asynchronous terminal or equivalent is required for testing with the Bring Up Driver (BUD) code. Settings are 8-bit, no parity, one stop bit, and 9600 baud. VT100 or VT52 emulator terminals may work. It is desirable to also have a video monitor for BUD tests. The boot code will boot with either an async console, a video on motherboard, or both.

Table A-1. Quickstart Peripheral List

Generic Description	Example Device
L2 SRAM card, 256KB	Alliance Semiconductor AS7M64P3256-15C
Loctite 384 adhesive, 300ml cartridge	Loctite Corp. 17041
Loctite 384 activator, 10 liter can	Loctite Corp. 17101
Video adapter card, PCI S3	Diamond Stealth (S3) 864
SCSI adaptor card, PCI	NCR 8100S with 609-039-1635 controller
Super I/O adaptor (IDE, floppy, serial ports, parallel ports, etc.)	Acculogic sIDE-4/HP (110-00139-00E00)
Audio adaptor card	Creative Labs Soundblaster 16
Floppy disk drive, 3.5"x1.44MB	Alps DFR723F, IBM 73G4514, Mitsubishi MF355F-258UG
Hard disk drive, SCSI-2, 8 bit	Quantum LPS270/5405, Maxtor MXT-540SL, IBM WDS-3200 (79F4042)
Hard disk drive, SCSI 1GB	IBM 94G3187
Hard disk drive, IDE 1GB	Maxtor HDMC71260AC
CD_ROM drive, internal SCSI	Toshiba XM-4101BMY
CD_ROM drive, internal SCSI, 4x	Toshiba 5301-4x
CD_ROM drive, internal IDE, 4x	Chinnon CDS5451

Table A-1. Quickstart Peripheral List (Continued)

Generic Description	Example Device
Chassis, Baby AT	Olsen Metal Products CC300249-17
Power supply, 200W Energy Star	API-3186S, IBM 06H2968
Box fan	Panaflow FBA08T12M
Box fan shock mounts	IBM 81F7977
Internal cables, floppy, SCSI, and CD-ROM	Standard cables
Speaker, internal 8Ω .5W 2pin	
LED, 2.5 ma drive	
Asynchronous terminal	IBM 3101
Super VGA monitor	IBM 6324, 6325, 6327, 9524, 9525, 9527, 9521
Keyboard, PS/2 compatible	
Mouse, PS/2 compatible	



Appendix B Planar I/O Subsystems

B.1 ISA Subsystem

The 100 MHz PPC 603e MCM example planar includes an ISA bus that is interfaced to the PCI bus by the ISA bus bridge. ISA expansion slots are provided by the example planar. The example planar uses a buffered subset of the ISA bus, called the X-bus, to host on-board native I/O, such as the real time clock and the keyboard and mouse controller.

B.1.1 The ISA Bridge

The ISA bridge function is provided by an Intel 82378ZB chip (SIO). It provides a PCI to ISA bus bridge, with the following major functions:

- Bridge between PCI and ISA
 - 8/16 bit ISA devices
 - 24 bit addressing on ISA
 - Partially decodes native I/O addresses
 - Unclaimed PCI memory address below 16MB forwarded to the ISA bus
 - Unclaimed PCI I/O address below 64K forwarded to the ISA bus
 - Powers up to an open condition (i.e., cycles may be passed to the ISA bus)
 - Generates ISA clock, with a programmable divide ratio of three or four
 - Allows ISA mastering and has programmable decodes that map ISA memory cycles to the PCI bus
 - 32-bit posted memory write data buffer (no I/O buffering)
- Seven channel ISA DMA controller
 - Function of two 83C37s with 32-bit extensions
 - Supports 8-bit or 16-bit devices on the ISA bus
 - Supports 32-bit addressing for ISA to PCI memory transfers
 - 8-byte bidirectional buffer for DMA data
- Timer block (function of 82C54)
- Interrupt Controller (function of two 8259s)
- PCI bus arbiter.
- Functions as PCI target during programming and ISA target cycles, and as busmaster during DMA or ISA master cycles

- Generates ISA_REFRESH# signal to refresh ISA bus DRAM.

B.1.2 Address Ranges

The ISA bus address ranges which may be separately enabled in the ISA bus bridge for forwarding to the PCI bus are:

- 0 - 512K
- 512K - 640K
- 640K - 768K
- 768K - 896K (in 8 ranges of 16 K each)
- 896K - 960K
- 1M - xM (where $x \leq 16$) with the hole (see the SIO data book). The hole may be 64K or 8M.

If an ISA DMA produces an address in the 0-16M range and this address is enabled in the ISA bridge for forwarding to the PCI, the ISA bridge will initiate a PCI transaction which the 660 bridge will forward to system memory.

The 660 uses medium timing when claiming ISA master originated cycles on the PCI. It does not use subtractive decoding. Hence, ISA masters can only communicate with other ISA devices or system memory. They may not communicate with PCI devices.

Warning: The software should not map any PCI memory at PCI addresses which ISA masters can create (those addresses between 0-16M which are programmed for forwarding from ISA to PCI). This is because contention would result between the device mapped at that address and the 660. Alternatively stated, ISA masters should not be allowed to create accesses to system memory using any address between 0 and 16M that is mapped to a PCI device, such as video.

B.1.3 ISA Bus Concurrency

ISA bus cycles which are not enabled for forwarding, including the hole, remain on the ISA bus. That is, DMA or ISA busmaster cycles on the ISA bus can run concurrently with PCI or CPU cycles.

B.1.4 ISA Busmasters and IGN_PCI_AD31

The ISA bridge supports ISA busmasters. System memory accesses from an ISA busmaster are designed to be mapped to the 0 to 16M range, and the ISA bridge forwards them to the PCI bus at the same range, which is not compliant to the PowerPC Reference Platform specification. Other PCI to system memory accesses however, are correctly mapped to the 2G to 4G range (for system memory address range from 0 to 2G). In some architectures this problem is handled by using the ISA_MASTER# signal, which is active during the ISA busmaster operation.

However, the ISA bridge allows ISA masters to run posted writes to system memory, without latching in the accompanying ISA_MASTER# signal. In this situation, the ISA_MASTER# signal is no longer synchronized to the ISA busmaster operation.

To overcome this challenge, the example system detects PCI memory transactions that are initiated by the SIO. The example planar ANDs together GNT0#, GNT1#, and GNT2#, to generate IGN_PCI_AD31, which is active high during the address phase of any PCI transaction that is not initiated by one of the three possible PCI agents (besides the 660 and the SIO). If the PCI transaction was not initiated by the 660, and if it is a memory transaction, then the 660 assumes that it is a system memory transaction initiated on the PCI bus by the SIO on behalf of an ISA busmaster, and so forwards it to the correct system memory address in the 0 to 16M range.

As a consequence of this design, the ISA bridge must be programmed to map ISA DMA (that is bound for system memory) to a PCI memory transaction using the 0 to 2G address range, rather than the apparently correct 2G to 4G range. Since the DMA sourced PCI transaction also causes IGN_PCI_AD31 to be asserted during the address phase of a PCI transaction initiated by the ISA bridge, the 660 will not do the usual inversion of the highest order address bit, but will forward the transaction to system memory in the 0 to 2G range.

Another consequence of the design is that the ISA bridge can not initiate peer to peer PCI memory transactions, because no matter what the PCI address is, it will be claimed by the 660 (if the address is that of a populated memory location) and mapped to system memory, possibly causing various inappropriate results.

These are the only limitations on the normal operation of the ISA bridge that are caused by the IGN_PCI_AD31 design, and there are no implications for other PCI or ISA agents, which are totally unaffected by the situation.

B.1.5 DMA

The DMA controller in the ISA bridge consists of the functionality of two 82C37A DMA controllers with 32-bit addressability extensions and enhanced functionality. The DMA request/grant lines are connected on the example planar as shown in Table B-1.

Table B-1. DMA Assignments

DMA Channel	Assignment or Connection
0	ISA connectors
1	ISA connectors
2	ISA connectors
3	ISA connectors
4	Cascade in
5	ISA connectors
6	ISA connectors
7	ISA connectors

B.1.5.1 Supported DMA Paths

DMA operations can be performed only:

- From ISA I/O mapped devices to ISA memory mapped devices, and
- From ISA I/O mapped devices to system memory (via the PCI bus). In these transfers, the system memory address must be mapped to the PCI address range 0 to 2G (see Section B.1.4).

The DMA source device can be located on the X-bus. If the DMA target is ISA memory mapped, it can also reside on the X-bus.

- ISA DMA to PCI I/O devices is not allowed.
- ISA DMA to PCI memory devices is not allowed.
- ISA DMA from ISA memory mapped devices is not allowed.

B.1.5.2 DMA Timing

The DMA controller runs compatible cycles for all ISA to ISA DMA transfers. Type A, type B and type F timing is available only for ISA I/O to system memory (via the PCI) DMA transfers.

B.1.5.3 Scatter-Gather

The example planar permits the use of independent scatter-gather (SG) operations on DMA channels 0-3 and 5-7. This operation chains together a number of DMA transfers to different memory locations so that they appear as one DMA transfer. The SG command, descriptor table, and status registers are relocatable via a configuration register in the ISA bridge. The termination of an operation may be signaled to the software by configuring any (or all) of the SG channels to use IRQ13 or by configuring the channel to signal end of process (aka Terminal Count) to the DMA device. See the SIO data book for details.

B.1.5.4 X-Bus

The X-bus is a utility bus, an 8 bit buffered version of the ISA bus, implemented on the reference board to support the native I/O devices that are located on the reference board. X-bus data transceiver U12 is controlled by the ISA bridge via XDIR (UBUSTR) and XDEN# (UBUSOE#). Various devices are located on the X-bus.

B.1.5.5 Control Signal Decodes

The System I/O CPLD (Chandra) is a programmable logic device that uses the X-bus signals and the partial decode signals from the ISA bridge to decode chip selects for various components. For more information on the System I/O CPLD, see that data sheet.

B.1.5.6 Keyboard/Mouse Controller

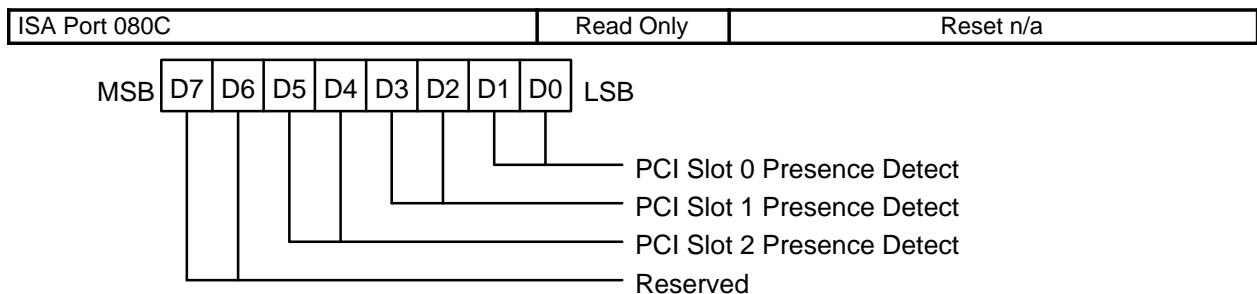
The example planar uses an Intel 8042AH as a keyboard and mouse controller, which resides on the X-bus. The code used is the same version as used in IBM Personal System/2 machines. This microcode may differ from other 8042 type keyboard controllers. These differences are usually only significant when porting AIX to the system (for more information contact your IBM representative. This device contains several registers. See the data sheet for more information.

B.1.5.7 Real Time Clock (RTC)

The example planar uses a Dallas Semiconductor™ DS1385S to provide the real time clock (TOD or RTC) function. This device is PC compatible and resides on the X-bus. It features an additional 4K of NVRAM and a replaceable battery. This device contains several registers. See the data sheet for more information.

B.1.5.8 PCI Adapter Card Presence Detect Register

The example planar uses U20 to buffer the PCI adaptor card presence detection bits onto the X-bus under control of the system I/O EPLD. These bits report in pairs, and do not contain any information about the identity of the card. They merely report on its presence.



Bits 7:6 Reserved

Bits 5:4 PCI Slot 2 Presence Detect Bits. 00 = Present, 11 = No PCI card installed.

Bits 3:2 PCI Slot 2 Presence Detect Bits. 00 = Present, 11 = No PCI card installed.

Bits 1:0 PCI Slot 2 Presence Detect Bits. 00 = Present, 11 = No PCI card installed.

B.1.5.9 L2 SRAM Identification Register

The example planar uses U19 to buffer the SRAM identification/presence detect bits from the SRAM socket onto the X-bus under control of the system I/O EPLD.

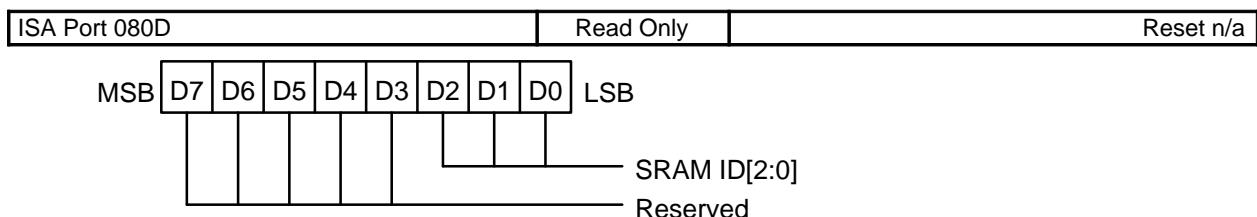
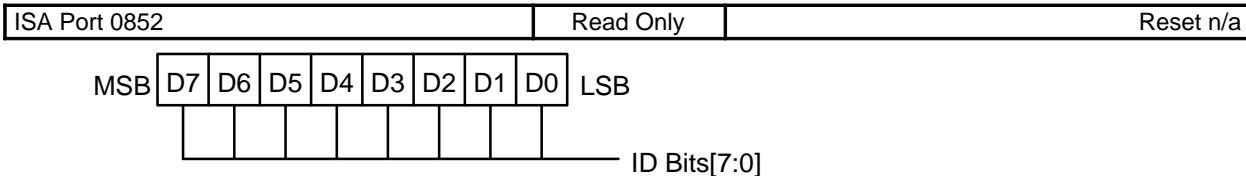


Table B-2. DRAM Module Presence Detect Bit Encoding

SRAM ID [2:0]			SRAM Module Identification
ID3	ID2	ID1	
0	0	0	512K Synchronous
0	0	1	256K Synchronous
0	1	0	Reserved
0	1	1	Reserved
1	0	0	512K Asynchronous
1	0	1	256K Asynchronous
1	1	0	1M Asynchronous
1	1	1	L2 SRAM module not present

B.1.5.10 Planar ID Detection Register

Revision information on the planar (motherboard) is buffered onto the X-bus by U18, under control of the system I/O EPLD.

**Table B-3. Planar ID Encoding**

Planar ID[7:0]	CPU	CPU Internal Clock/ CPU Bus Clock/ PCI Bus Clock	Schematic
0C	603	66/66/33	MPRH02SCU-01
0D	603e	99/66/33	MPRH02SCU-01
0E	604	132/66/33	MPRH02SCU-01
other	—	—	Reserved

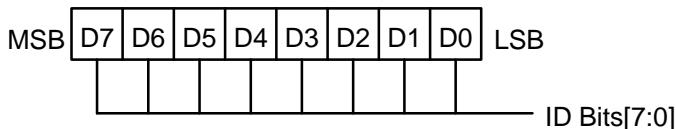
B.1.5.11 DRAM Presence Detection

DRAM module presence and identification data is hard coded into the pinout of the SIMM (or DIMM) by shorting particular pins to ground or no-connecting them on the SIMM itself. The example planar uses U17 and U40 to buffer the presence detect bits from the DRAM sockets onto the X-bus under control of signals from the system I/O EPLD. This information appears as the DRAM SIMM 1-2 Memory ID Register and the DRAM SIMM 3-4 Memory ID Register.

B.1.5.12 DRAM SIMM 1-2 Memory ID Register

ISA Port 0880	Read Only	Reset n/a
---------------	-----------	-----------

This register indicates the ID bits associated with SIMMs 1 and 2.



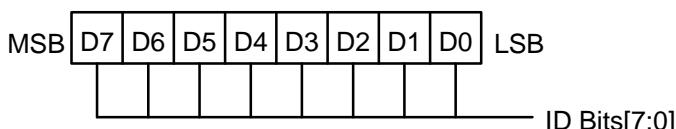
Bits 7:4 SIMM 2 ID bits. See Table B-4.

Bits 3:0 SIMM 1 ID bits. See Table B-4.

B.1.5.13 DRAM SIMM 3-4 Memory ID Register

ISA Port 0881	Read Only	Reset n/a
---------------	-----------	-----------

This register indicates the ID bits associated with SIMMs 3 and 4.



Bits 7:4 SIMM 4 ID bits. See Table B-4.

Bits 3:0 SIMM 3 ID bits. See Table B-4.

Table B-4. DRAM Module Presence Detect Bit Encoding

Presence Detect Bits				DRAM Module Identification
PD4	PD3	PD2	PD1	
1	0	0	0	4 MB (1M x 36b) 70ns
1	0	1	1	8 MB (2M x 36b) 70ns
1	0	1	0	16 MB (4M x 36b) 70ns
1	1	0	1	32 MB (8M x 36b) 70ns
1	1	1	1	No DRAM module installed.
Other encodings			Not currently defined.	

B.1.6 Miscellaneous

B.1.6.1 Speaker Support

The example planar has a connector for a small speaker. The speaker output is driven by the timer 2 signal from the ISA bridge. The intended speaker is a typical PC type, 8 ohms and .5W.

B.2 System EPLD

Note: The System I/O EPLD generates system control register access signals from X-bus I/O port transactions, supports power management, and provides various other system functions.

In this section, the system I/O EPLD is referred to as the EPLD.

The EPLD is a programmed Altera™ EPM5130QC100 electrically programmable logic device. For timing and electrical specifications, see that data sheet.

B.2.1 System Register Support

The EPLD supports both internal and external registers.

B.2.1.1 External Register Support

The EPLD supports a group of external registers, which are latches or other devices that are physically located in a device other than the EPLD. As shown in Figure B-1, the EPLD supplies control signals to the external registers, based on a decode of the address and control lines of the X-bus (or ISA bus). In response to the signals from the EPLD, the external register either reads or writes data to the X-bus.

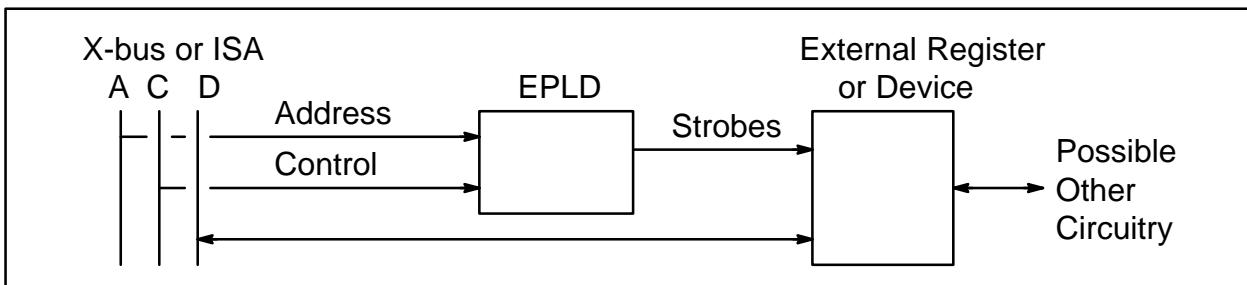


Figure B-1. Typical External Register

For the external registers that the EPLD supports, Table B-5 shows the external register, the ISA I/O port address, and the supplied control signal(s).

Table B-5. External Register Support

ISA Port Address	Register	Read/Write	Register Location	Signal	Strobe or Function
0060 or 0062 or 0064 or 0066	Keyboard Controller Registers	R/W	Keyboard Controller	KBD_CS# (asserted for an access to any of these addresses)	Address Decode
0070	RTC Address Latch Enable	W/O	RTC	RTC_ALE	Write strobe
0071	RTC Data	Write	RTC	RTCWR#	Write strobe
		Read		RTCD\$#	Read strobe
0074	NVRAM Address Low Byte	W/O	NVRAM	AS0#	Address Decode
0075	NVRAM Address High Byte	W/O	NVRAM	AS1#	Address Decode
0077	NVRAM Data	Write	NVRAM	NVRAMWE#	Write Strobe
		Read		NVRAMOE#	Read Strobe
080C	Equipment Present	R/O	Board	PRSNT_RD#	Read Strobe
080D	L2 Cache Status Register	R/O	Board	L2_STATUS_RD#	Read Strobe
0852	Planar ID	R/O	Board	PLANAR_ID_RD#	Read Strobe
0880	DRAM Presence Detect 1/2	R/O	Board	DRAM_PD_RD1#	Read Strobe
0881	DRAM Presence Detect 3/4	R/O	Board	DRAM_PD_RD2#	Read Strobe

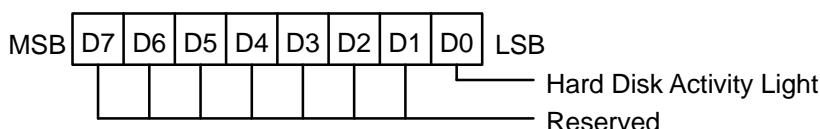
B.2.1.2 Internal Registers

The EPLD contains a group of internal registers, which are accessed via the ISA bus I/O port address shown for each register.

Storage Light Register

ISA Port 0808	Read/Write	Reset to xxxx xx0
---------------	------------	-------------------

This register controls the HDD_LED# output of the EPLD. This signal normally controls the hard disc drive activity LED.

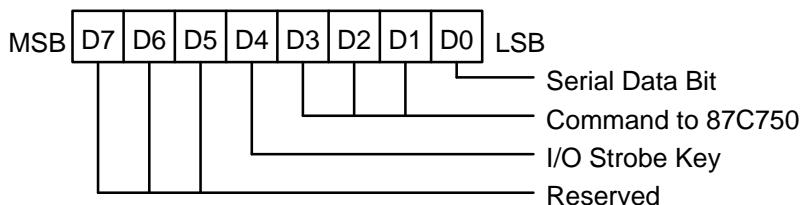


Bit 0 . . . Hard Disk Activity Light:

- 0 = Turn light off (negate HDD_LED#).
- 1 = Turn light on (assert HDD_LED#).

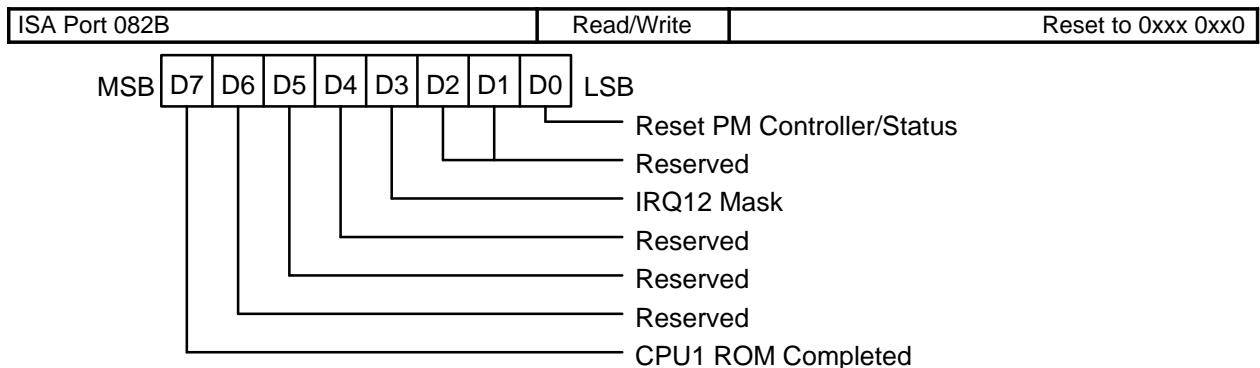
Power Management Control Register 1

ISA Port 082A	Read/Write	Reset to xxxx xxxx
---------------	------------	--------------------



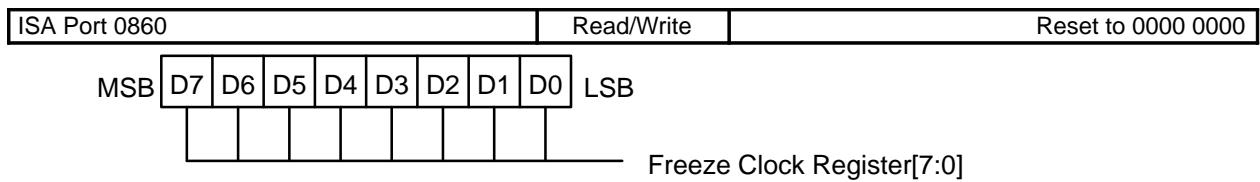
This register is part of the power management system. For a detailed functional description of the operation of this register, see the *603/604 Reference Design Power Management Specification*.

Power Management Control Register 2



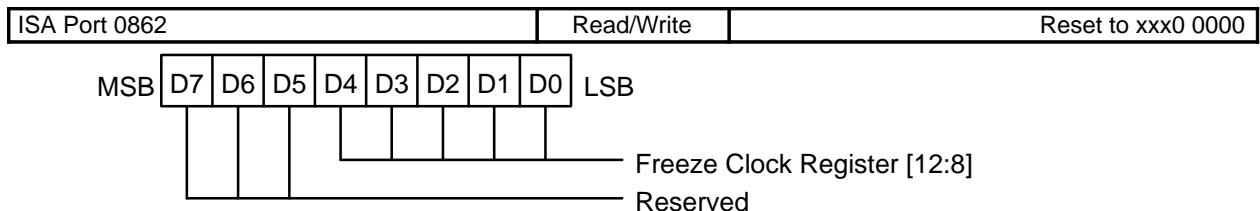
This register is part of the power management system. For a detailed functional description of the operation of this register, see *The 603/604 Reference Design Power Management Specification*.

Freeze Clock Register (FCR) Low



See FCR high.

Freeze Clock Register (FCR) High



The freeze clock register (FCR) is a 13 bit register that is accessed by the system via the X-bus as two 8-bit registers, as shown in this and the previous sections. Once triggered, the FCR data is shifted out as serial data on FRZ_DATA_OUT.

This function is intended for use with the MPC970 clock chip, as is done in the example planar. The MPC970 contains an input shift register, the input of which (Frz_Data) is connected to FRZ_DATA_OUT. ISA_CLK is used to clock the data from the EPLD to the MPC970.

EPLD will shift the freeze clock data out of the FCR in response to either one of two triggers:

1. A write to ISA I/O port 862, or

2. A low to high transition on the UNFREEZE input iff bit 0 and bit 1 of the FCR are high.

After triggering the data transfer, wait at least 2.3us for the transfer to complete before accessing the FCR or retriggering the data transfer.

This is a one way serial data transfer between the two devices. Reading the FCR returns the contents of the FCR, and does not cause a read of the data in the MPC970 register.

A 1 in a bit position freezes the corresponding clock output of the MPC970. For details of the data transfer operation and the meaning of the data, see the MPC970 data sheet.

B.2.2 Signal Descriptions

Table B-6 shows the active signals of the EPLD. Pins not shown should not be connected.

Table B-6. Signal Descriptions

Signal Name	Pin	I/O	Description
X-bus Interface Signals			
ECS[2:0]	21, 22, 59	I	Encoded Chip Select [2:0]. Encoded chip selects for peripheral devices supported by the ISA I/O Bridge. Used by EPLD X-bus I/O port address decoders. From SIO.
ECSEN#	60	I	Encoded Chip Select Enable. Asserted to enable the base decoder. Negated to select the option decoder. Used by EPLD X-bus I/O port address decoders. From SIO.
XA[7:0]	61 ,64, 65, 66, 67, 70, 71, 72.	I	X-address bus [7:0]. Used by EPLD X-bus I/O port address decoders.
XD[7:0]	31, 32, 33, 92, 73, 74, 34, 95.	I/O 24mA	X-data bus [7:0]. Used by EPLD to transfer data.
XIOW#	16	I	X-bus I/O Write. This signal indicates that the system is writing to an X-bus I/O device. Used by EPLD X-bus I/O port address decoders.
XIOR#	9	I	X-bus I/O Read. This signal indicates that the system is reading from an X-bus I/O device. Used by EPLD X-bus I/O port address decoders.
External Register Support Signals			
AS0#	50	O 6mA	NVRAM address register low byte write strobe. EPLD asserts this signal to write to X-bus port 0074.
AS1#	49	O 6mA	NVRAM address register high byte write strobe. EPLD asserts this signal to write to X-bus port 0075.
DRAM_PD_RD1#	57	O 6mA	DRAM SIMM presence detect read enable 1. EPLD asserts this signal to read X-bus port 0880.
DRAM_PD_RD2#	56	O 6mA	DRAM SIMM presence detect read enable 2. EPLD asserts this signal to read X-bus port 0881.
KYBD_CS#	98	O 6mA	Keyboard chip select. EPLD asserts this signal to read X-bus ports 0060, 0062, 0064, and 0066.
L2_STATUS_RD#	54	O 6mA	L2 cache status register read strobe. EPLD asserts this signal to read X-bus port 080D.
NVRAMOE#	1	O 6mA	NVRAM output enable (read strobe). EPLD asserts this signal to read X-bus port 0076. This normally causes a read of the NVRAM data stored at the location contained in the NVRAM address register.

Table B-6. Signal Descriptions (Continued)

Signal Name	Pin	I/O	Description
External Register Support Signals			
NVRAMWE#	53	O 6mA	NVRAM data write strobe. EPLD asserts this signal to write to X-bus port 0076. This normally causes the data associated with this write to be written into the NVRAM location contained in the NVRAM address register.
PLANAR_ID_RD#	52	O 6mA	Planar ID read. EPLD asserts this signal to read X-bus port 0852.
PRSNT_RD#	41	O 6mA	Equipment present register read. EPLD asserts this signal to read X-bus port 080C.
RTC_ALE	48	O 6mA	Real time clock address latch enable. EPLD asserts this signal to write X-bus port 0070.
RTCDS#	51	O 6mA	Real time clock read strobe. EPLD asserts this signal to read X-bus port 0071.
RTCWR#	35	O 6mA	Real time clock write strobe. EPLD asserts this signal to write to X-bus port 0071.
Interrupt Signals			
IRQ1	86	O	Interrupt request 1. Latched active when IRQ1_IN is asserted. Negated when KYBD_CS# is asserted.
IRQ1_IN	89	I	Keyboard interrupt. EPLD is designed to intercept the keyboard interrupt between the keyboard and the ISA bus bridge. Either connect IRQ1 and IRQ1_IN as shown in the example planar schematics, or disconnect both signals from the system (routing the keyboard interrupt to the ISA bus bridge), or see the <i>603/604 Reference Design Power Management Guide</i> .
IRQ12	36	I	Interrupt request 12 input. Connect to system IRQ12, the mouse interrupt. Also see the <i>603/604 Reference Design Power Management Guide</i> .
System Clock Interface Signals			
FRZ_DATA_OUT	90	O 6mA	Freeze data out. Serial data stream to MPC970 clock chip. See the MPC970 data sheet.
ISA_CLK	20	I	ISA clock. Used to clock the freeze serial data stream to the MPC970 clock chip. See the MPC970 data sheet. Also see the <i>603/604 Reference Design Power Management Guide</i> .
Power Management Signals (not used in release 2.0 or 2.1)			
83CX_RESET	91	O 6mA	Power management controller reset. No-connect or connect as shown in the <i>603/604 Reference Design Power Management Guide</i> .
ACTIVITY#	58	O 6mA	Activity. No-connect or connect as shown in the <i>603/604 Reference Design Power Management Guide</i> .
CMD_STATE#	8	I	Power management controller command state. No-connect or connect as shown in the <i>603/604 Reference Design Power Management Guide</i> .
EXT_ACTVITY#	3	I	External activity. No-connect or pull low or connect as shown in the <i>603/604 Reference Design Power Management Guide</i> .
IO_STROBE#	99	O 6mA	I/O strobe. No-connect or connect as shown in the <i>603/604 Reference Design Power Management Guide</i> .
PROC_RDY	83	I	Power management controller ready. No-connect or connect as shown in the <i>603/604 Reference Design Power Management Guide</i> .
RWD0	96	I/O 24mA	Power management controller serial read/write data bit. No-connect or connect as shown in the <i>603/604 Reference Design Power Management Guide</i> .

Table B-6. Signal Descriptions (Continued)

Signal Name	Pin	I/O	Description
Power Management Signals (not used in release 2.0 or 2.1)			
UNFREEZE	82	I	Unfreeze. No-connect or connect as shown in the 603/604 Reference Design Power Management Guide.
Other Signals			
HDD_LED#	100	O 6mA	Hard disk drive activity light. EPLD asserts this signal while bit 0 of the storage light register (port 0808) is 1. This signal normally indicates hard disk drive activity.
RESET#	10	I	System reset. Used by EPLD to reset internal state machines and internal registers.
VCC	18, 19, 43, 44, 68, 69, 93, 94	I	+5V:
GND	12, 13, 37, 38, 62, 63, 87, 88	I	GROUND:

B.2.3 EPLD Design Equations**B.2.3.1 Fit File**

```
-- MAX+plus II Compiler Fit File
-- Version 5.0 8/5/94
-- Compiled: 05/05/95 14:44:22
```

BEGIN

```
DEVICE = "EPM5130WC-1";
"A0" : INPUT_PIN = 72 ;
"A1" : INPUT_PIN = 71 ;
"A2" : INPUT_PIN = 70 ;
"A3" : INPUT_PIN = 67 ;
"A4" : INPUT_PIN = 66 ;
"A5" : INPUT_PIN = 65 ;
"A6" : INPUT_PIN = 64 ;
"A7" : INPUT_PIN = 61 ;
"/CMD_STATE" : INPUT_PIN = 8 ; -- LC8
"/ECSEN" : INPUT_PIN = 60 ;
"ECS0" : INPUT_PIN = 59 ;
"ECS1" : INPUT_PIN = 22 ;
"ECS2" : INPUT_PIN = 21 ;
"/EXT_ACTVTY" : INPUT_PIN = 3 ; -- LC3
"IRQ1" : INPUT_PIN = 89 ; -- LC103
"IRQ12" : INPUT_PIN = 36 ; -- LC38
"ISA_CLK" : INPUT_PIN = 20 ;
"PROC_RDY" : INPUT_PIN = 83 ; -- LC99
"/RESET" : INPUT_PIN = 10 ;
"UNFREEZE" : INPUT_PIN = 82 ; -- LC98
"/XIOR" : INPUT_PIN = 9 ;
"/XIOW" : INPUT_PIN = 16 ;
```

```

"/ACTIVITY"
"/AS0"
"/AS1"
"/DRAM_PD_RD1"
"/DRAM_PD_RD2"
"FRZ_DATA_OUT"
"/HDD_LED"
"/IO_STROBE"
"IRQ1_OUT"
"/KYBD_CS"
"/L2_STATUS_RD"
"/NVRAMOE"
"/NVRAMWE"
"/PLANAR_ID_RD"
"/PRSNT_RD"
"RTC_ALE"
"/RTCDS"
"/RTCWR"
"83CX_RESET"
"RWD0"
"XD0"
"XD1"
"XD2"
"XD3"
"XD4"
"XD5"
"XD6"
"XD7"
"CLKFF0"
"CLKFF1"
"CLKFF2"
"CLKFF3"
"CLKFF4"
"CLKFF5"
"CLKFF6"
"CLKFF7"
"CLKFF8"
"CLKFF9"
"CLKFF10"
"CLKFF11"
"CLKFF12"
"CNTR0"
"CNTR1"
"CNTR2"
"CNTR3"
"DOUBLE_FRZ"

: OUTPUT_PIN = 58 ; -- LC72
: OUTPUT_PIN = 50 ; -- LC56
: OUTPUT_PIN = 49 ; -- LC55
: OUTPUT_PIN = 57 ; -- LC71
: OUTPUT_PIN = 56 ; -- LC70
: OUTPUT_PIN = 90 ; -- LC104
: OUTPUT_PIN = 100 ; -- LC120
: OUTPUT_PIN = 99 ; -- LC119
: OUTPUT_PIN = 86 ; -- LC102
: OUTPUT_PIN = 98 ; -- LC118
: OUTPUT_PIN = 54 ; -- LC68
: OUTPUT_PIN = 1 ; -- LC1
: OUTPUT_PIN = 53 ; -- LC67
: OUTPUT_PIN = 52 ; -- LC66
: OUTPUT_PIN = 41 ; -- LC49
: OUTPUT_PIN = 48 ; -- LC54
: OUTPUT_PIN = 51 ; -- LC65
: OUTPUT_PIN = 35 ; -- LC37
: OUTPUT_PIN = 91 ; -- LC113
: BIDIR_PIN = 96 ; -- LC116
: BIDIR_PIN = 95 ; -- LC115
: BIDIR_PIN = 34 ; -- LC36
: BIDIR_PIN = 74 ; -- LC82
: BIDIR_PIN = 73 ; -- LC81
: BIDIR_PIN = 92 ; -- LC114
: BIDIR_PIN = 33 ; -- LC35
: BIDIR_PIN = 32 ; -- LC34
: BIDIR_PIN = 31 ; -- LC33
: LOCATION = LC48 ;
: LOCATION = LC96 ;
: LOCATION = LC95 ;
: LOCATION = LC94 ;
: LOCATION = LC47 ;
: LOCATION = LC46 ;
: LOCATION = LC45 ;
: LOCATION = LC44 ;
: LOCATION = LC43 ;
: LOCATION = LC93 ;
: LOCATION = LC92 ;
: LOCATION = LC91 ;
: LOCATION = LC112 ;
: LOCATION = LC111 ;
: LOCATION = LC110 ;
: LOCATION = LC109 ;
: LOCATION = LC108 ;
: LOCATION = LC90 ;

```

```

"DOUBLE_SNC" : LOCATION = LC107 ;
"FREEZEFFF" : LOCATION = LC106 ;
"GEN_STOP_BITFFF" : LOCATION = LC105 ;
"GPCS0" : LOCATION = LC89 ;
"PWR_REG21" : LOCATION = LC80; --IRQ12_MASK
"PWR_REG22" : LOCATION = LC79 ;
"SHIFT_ENFFF" : LOCATION = LC103 ; -- PIN 89
"SNC_FREEZE" : LOCATION = LC99 ; -- PIN 83
"SNC_SHIFT_ENFFF" : LOCATION = LC98 ; -- PIN 82
"SNC_UNFREEZE" : LOCATION = LC101 ; -- PIN 85
"START_SHIFTFFF" : LOCATION = LC100 ; -- PIN 84
"UNFREEZEFFF" : LOCATION = LC97 ; -- PIN 81

END;

```

B.2.3.2 TDF File

```
%*****AHDL SOURCE CODE FOR: SIO/XBUS INTERFACE CONTROL*****%
```

```

SUBDESIGN chandra(
%%
%      DEFINE PRIMARY INPUTS AND OUTPUTS %
%%
A[7..0]          :INPUT; % A7 is MSB and A0 is LSB %
ECS[2..0]         :INPUT; % ECS2 is MSB and ECS0 is LSB %
/ECSEN           :INPUT;
/XIOR            :INPUT;
/XIOW            :INPUT;
/RESET           :INPUT;
/CMD_STATE       :INPUT;
PROC_RDY         :INPUT;
/EXT_ACTVTY     :INPUT;
IRQ1             :INPUT;
IRQ12            :INPUT;% Freeze Clock %
ISA_CLK          :INPUT;
UNFREEZE        :INPUT;

/PLANAR_ID_RD   :OUTPUT;
/PRSNT_RD        :OUTPUT;
/L2_STATUS_RD   :OUTPUT;
/DRAM_PD_RD1    :OUTPUT;
/DRAM_PD_RD2    :OUTPUT;
/KYBD_CS         :OUTPUT;
IRQ1_OUT         :OUTPUT;
RTC_ALE          :OUTPUT;
/RTCDS           :OUTPUT;
/RTCWR           :OUTPUT;
/AS0              :OUTPUT;
/AS1              :OUTPUT;
/NVRAMWE         :OUTPUT;
/NVRAMOE         :OUTPUT;
/HDD_LED         :OUTPUT;

```

```

83CX_RESET      :OUTPUT;
/ACTIVITY       :OUTPUT;
/IO_STROBE      :OUTPUT;
FRZ_DATA_OUT    :OUTPUT;

XD[7..0]         :BIDIR;    % XD7 is MSB and XD0 is LSB %
RWD0             :BIDIR;
                      )

VARIABLE          % Misc. Variables%
GPCS0            :NODE;
HDD_LEDFF        :SRFF;
D[7..0]           :NODE;
XD_TRI_OE        :NODE;
LIGHT_STRB       :NODE;
INT1FF           :DFF;      % Keyboard interput latch %
CLRINT1          :NODE;      % Power Management Variables%
PWR_REG1_STRB   :NODE;
PWR_REG2_STRB   :NODE;
PWR_REG2[2..0]   :SRFF;
83CX_CS          :NODE;
RWD0_STRB        :NODE;
IRQ12_MASK       :NODE;      % Freeze Clock Variables %
CLKFF[12..0]     :DFFE;
SHIFT_ENFF       :SRFF;
CNTR[3..0]        :DFF;
SNC_SHIFT_ENFF  :DFFE;
START_SHIFTFF   :SRFF;
CLKFF_WR         :NODE;
CLKFF_STRB      :NODE;
CLKFF_SELL      :NODE;
CLKFF_SELH      :NODE;
GEN_START_BIT   :NODE;
GEN_STOP_BITFF  :DFF;
STOP_SHIFT       :NODE;
UNFREEZEFF      :DFFE;      % Unsynchronized UNFREEZE      %
SNC_UNFREEZE    :DFF;      % Synchronized UNFREEZE      %
DOUBLE_SNC       :DFF;      % Double Synchronized UNFREEZE %
FREEZEFF        :DFFE;      % Unsynchronized FREEZE      %
SNC_FREEZE      :DFF;      % Synchronized FREEZE      %
DOUBLE_FRZ       :DFF;      % Double Synchronized FREEZE %

BEGIN*****%
% CHANDRA uses SIO's General Purpose Register Decode 0 -- Software sets      %
% to 800-8FF to enable the following decoded signals.                          %
GPCS0 = LCELL (!ECS[2] & ECS[1] & !ECS[0] & /ECSEN);                         %

% Hard Disk Light I/O address range : 0808                                %
LIGHT_STRB = !A[7] & !A[6] & !A[5] & !A[4] & A[3] & !A[2]
              & !A[1] & !A[0] & GPCS0; HDD_LEDFF.s = XD[0] & LIGHT_STRB;
HDD_LEDFF.r = !XD[0] & LIGHT_STRB;
HDD_LEDFF.clk= GLOBAL(/XIOW);
HDD_LEDFF.clrn= (/RESET);

```

```

/HDD_LED      = !HDD_LEDFF.q;
%%
% Equipment present Read Command 1 (I/O address range: 080C) %
%%
/PRSNT_RD    = ! ( !A[7] & !A[6] & !A[5] & !A[4] & A[3] & A[2]
                  & !A[1] & !A[0] & GPCS0 & !/XIOR);
%%
% Equipment present Read Command 2 (I/O address range: 080D) %
%%
/L2_STATUS_RD = !( !A[7] & !A[6] & !A[5] & !A[4] & A[3] & A[2]
                  & !A[1] & A[0] & GPCS0 & !/XIOR);
%%
% External/Internal Planar ID I/O address range: 0852 %
%%
/PLANAR_ID_RD = !( !A[7] & A[6] & !A[5] & A[4] & !A[3] & !A[2]
                  & A[1] & !A[0] & GPCS0 & !/XIOR);
%%
% Keyboard Chip Select I/O address range = 0060, 0062, 0064, 0066- %
%%
/KYBD_CS     = !( !ECS[2] & ECS[1] & !ECS[0] & !/ECSEN);

% Keyboard interrupt (INT1) Clear = KBD_CS qualified with XIOR %

CLRINT1 = !ECS[2] & ECS[1] & !ECS[0] & !/ECSEN & !/XIOR; INT1FF.d=VCC;
INT1FF.clk = IRQ1;           % Set Keyboard INT1 latch on rising edge of IN1 %
INT1FF.clrn = /RESET & !CLRINT1;
IRQ1_OUT = INT1FF.q;

%%
% RTC_ALE I/O address 0070 %
%%
RTC_ALE = ( !A[2] & !A[1] & !A[0] & !ECS[2] & !ECS[1] & !/ECSEN & !/XIOW) ;
%%
% RTCWR I/O address 0071 %
%%
/RTCWR = !( !A[2] & !A[1] & A[0] & !ECS[2] & !ECS[1] & !/ECSEN & !/XIOW);
%%
% RTCDS I/O address 0071 %
%%
/RTCDS = !( !A[2] & !A[1] & A[0] & !ECS[2] & !ECS[1] & !/ECSEN & !/XIOR);
%%
% Nvram AS0 I/O address 0074 %
%%
/AS0      = !( A[2] & !A[1] & !A[0] & !ECS[2] & !ECS[1] & !/ECSEN & !/XIOW);
%%
% Nvram AS1 I/O address range: 0075 %
%%
/AS1      = !( A[2] & !A[1] & A[0] & !ECS[2] & !ECS[1] & !/ECSEN & !/XIOW);
%%
% Nvram RTC WER I/O address 0077 %
%%

```

```
/NVRAMWE = !(A[2] & A[1] & A[0] & !ECS[2] & !ECS[1] & !/ECSEN & !/XIOW);
%%
% Nvram RTC OER I/O address range: 0077 %
%%
/NVRAMOE = !(A[2] & A[1] & A[0] & !ECS[2] & !ECS[1] & !/ECSEN & !/XIOR)
; %%
%
% DRAM_PD_RD1 enables Present Detect I/O address range: 0880 %
%
/DRAM_PD_RD1 = !(A[7] & !A[6] & !A[5] & !A[4] & !A[3] & !A[2] & !A[1]
& !A[0] & GPCS0 & !/XIOR);
```

```
%%
% DRAM_PD_RD2 enables Present Detect I/O address range: 0881 %
%%
/DRAM_PD_RD2 = !(A[7] & !A[6] & !A[5] & !A[4] & !A[3] & !A[2] & !A[1]
& A[0] & GPCS0 & !XIOR);

%%
%*****POWER MANAGEMENT*****
%
% Write Power Control Register 1 I/O address: 082A
%
% (MSB) Bits 7-5 Reserved %
% Bit 4 I/O Strobe Key (W/O) %
% Bits 3-1 Data/Command B[3..1] to 83C750 (W/O) %
% (LSB) Bit 0 83C750 D0 (R/W) %

83CX_CS = (!A[7] & !A[6] & A[5] & !A[4] & A[3] & !A[2] & A[1] & GPCS0);
PWR_REG1_STRB = 83CX_CS & !A0;

%%
% -Write ZERO to 83C750 %
%%
RWDO_STRB = (PWR_REG1_STRB & !XIOW & !XD[0]);
RWDO = TRI (GND, RWDO_STRB);

%%
% 83C750 I/O Strobe Key %
%%
/IO_STROBE = !(XD[4] & PWR_REG1_STRB & !XIOW & PROC_RDY
& !83CX_RESET);

%IO_CHRDY = TRI (GND, IO_STRB); %

%*****Write Power Control Register 2 I/O address: 082B*****
%
% (MSB) Bit 7 CPU1 ROM Completed (Reset by /RESET, W/R) %
% Bit 6 Reserved %
% Bit 5 Reserved %
% Bit 4 Reserved %
% Bit 3 IRQ12 Mask (Reset by /RESET) (W/R) %
% Bits 2-1 Reserved %
% (LSB) Bit 0 Reset 83C750 (W/O) %

PWR_REG2_STRB = 83CX_CS & A0;
PWR_REG2[1].s = XD[3] & PWR_REG2_STRB;
PWR_REG2[1].r = !XD[3] & PWR_REG2_STRB;
PWR_REG2[1].clrn = (/RESET);
PWR_REG2[2].s = XD[7] & PWR_REG2_STRB;
PWR_REG2[2].r = !XD[7] & PWR_REG2_STRB;
PWR_REG2[2].clrn = (/RESET);
PWR_REG2[0].s = XD[0] & PWR_REG2_STRB;
PWR_REG2[0].r = !XD[0] & PWR_REG2_STRB;
```

```

PWR_REG2[0].clr_n = VCC;
PWR_REG2[].clk = GLOBAL(/XIOW);
83CX_RESET = PWR_REG2[0].q;
IRQ12_MASK = PWR_REG2[1].q;

%*****
% This output should be NORed with IRQ1 and EXTACTIV signals externally to%
% provide a real Activity Alert to the 87C350 (because of restriction of %
% CHANDRA's IO pins.)
%
%*****
/ACTIVITY = !(IRQ1 # !IRQ12_MASK & IRQ12 # !/EXT_ACTVTY);

%*****
% Freeze Clock Logic - Refer to the Motorola MC88LV970 PLL Clock Driver %
% Specification for information on Freeze Data protocol.
%
% Freeze Clock Logic contains a 12 Bit serial shift register %
% Decode the low order CLKFF[7..0] on addresses: 0860 -0861 %
% Decode the high order CLKFF[12..8] on addresses: 0862 -0863 %
%*****
CL
KFF_STRB = (!A[7] & A[6] & A[5] & !A[4] & !A[3] & !A[2] & GPCS0);
CLKFF_WR = CLKFF_STRB & !/XIOW;
CLKFF_SELL = CLKFF_STRB & !A[1];
CLKFF_SELH = CLKFF_STRB & A[1];
CLKFF[12..0].prn = VCC;
CLKFF[12..0].clr_n = /RESET;
CLKFF[12..0].ena = START_SHIFTFF.q # CLKFF_WR;
CLKFF[12..0].clk = ISA_CLK;

if START_SHIFTFF.q then                                % Shift with wraparound %
    CLKFF[11..0].d = CLKFF[12..1].q;
    CLKFF[12].d = CLKFF[0].q;
else if CLKFF_SELL then                            % Write to CLKFF[7..0] %
    CLKFF[7..0].d = XD[7..0];
    CLKFF[12..8].d = CLKFF[12..8].q;
    else if CLKFF_SELH then                         % Write to CLKFF[12..8] %
        CLKFF[7..0].d = CLKFF[7..0].q;
        CLKFF[12..8].d = XD[4..0];
    else
        CLKFF[12..0].d = CLKFF[12..0].q;          % Do nothing %
        end if;
    end if;
end if;

%*****
% UNFREEZE Flip Flop is cleared asynchronously when the UNFREEZE signal %
% makes a low to high transition. It is set once shifting has been enabled%
% The CHANDRA ignores the UNFREEZE signal if the 601 PCLK clocks are not %
% frozen.
%*****
UN-
FREEZEFF.prn = !SNC_SHIFT_ENFF.q & /RESET;
UNFREEZEFF.clr_n = VCC;
UNFREEZEFF.ena = !SHIFT_ENFF.q ;
UNFREEZEFF.d = !(CLKFF[0].q # CLKFF[1].q);

```

```

UNFREEZEFF.clk = UNFREEZE ;

%*****
% This signal is the UNFREEZE Flip Flop synchronized to the ISA clock. %
%*****
SNC_UNFREEZE.prn = /RESET;
SNC_UNFREEZE.clrn = VCC;
SNC_UNFREEZE.d = UNFREEZEFF.q;
SNC_UNFREEZE.clk = ISA_CLK;

%*****
%
% This signal is the UNFREEZE Flip Flop double synchronized to the ISA
clock.%*****
%
DOUBLE_SNC.prn = /RESET;
DOUBLE_SNC.clrn = VCC;
DOUBLE_SNC.d = SNC_UNFREEZE.q;
DOUBLE_SNC.clk = ISA_CLK;

%*****
% FREEZE Flip Flop is cleared asynchronously when the UNFREEZE signal %
% makes a high to low transition. It is set once shifting has been enabled%
% The CHANDRA ignores the UNFREEZE signal if the 601 PCLK clocks are %
% already frozen. %
%*****
FREEZEFF.prn = !SNC_SHIFT_ENFF.q & /RESET;
FREEZEFF.clrn = VCC;
FREEZEFF.ena = !SHIFT_ENFF.q ;
FREEZEFF.d = CLKFF[0].q & CLKFF[1].q;
FREEZEFF.clk = !UNFREEZE ;

%*****
% This signal is the FREEZE Flip Flop synchronized to the ISA clock. %
%*****
SNC_FREEZE.prn = /RESET;
SNC_FREEZE.clrn = VCC;
SNC_FREEZE.d = FREEZEFF.q;
SNC_FREEZE.clk = ISA_CLK;

%*****
%
% This signal is the FREEZE Flip Flop double synchronized to the ISA
clock.%*****
%
DOUBLE_FRZ.prn = /RESET;
DOUBLE_FRZ.clrn = VCC;
DOUBLE_FRZ.d = SNC_FREEZE.q;
DOUBLE_FRZ.clk = ISA_CLK;
SHIFT_ENFF.s = CLKFF_SELH;      % Start shifting upon write to 862      %
SHIFT_ENFF.r = GND;
SHIFT_ENFF.prn = DOUBLE_SNC.q & DOUBLE_FRZ;
SHIFT_ENFF.clrn = !GEN_STOP_BITFF.q & /RESET;  %Clr when it reaches 15 & rst%
SHIFT_ENFF.clk = /XIOW;SNC_SHIFT_ENFF.d = SHIFT_ENFF.q;

```

```

% One clock after SHIFT_ENFF%
SNC_SHIFT_ENFF.ena = /RESET;
SNC_SHIFT_ENFF.clrn = /RESET;
SNC_SHIFT_ENFF.clk = ISA_CLK;
CNTR[].clk = ISA_CLK;
CNTR[].clrn = /RESET;           % Counter resets to zero          %
If SNC_SHIFT_ENFF.q Then       % Count while SNC_SHIFT_ENFF is 1      %
  CNTR[].d = CNTR[].q + 1;
Else
  CNTR[].d = CNTR[].q;
End If;

GEN_START_BIT = !CNTR[3].q & !CNTR[2].q & !CNTR[1].q & CNTR[0].q;  % 1 %
STOP_SHIFT    = CNTR[3].q & CNTR[2].q & CNTR[1].q & !CNTR[0].q;  % 14 %
GEN_STOP_BITFF.d = STOP_SHIFT;  %Generate Stop Bit when counter is 15 %
GEN_STOP_BITFF.prn = VCC;
GEN_STOP_BITFF.clrn = /RESET;
GEN_STOP_BITFF.clk = ISA_CLK;
START_SHIFTFFF.s = GEN_START_BIT;
START_SHIFTFFF.r = STOP_SHIFT;
START_SHIFTFFF.clrn = /RESET;
START_SHIFTFFF.clk = ISA_CLK;

%*****%
% FRZ_DATA_OUT is a 1 when the counter = 0. It is also a 1 whenever      %
% START_SHIFTFFF is 1 and CLKFF[0] is a 1 and then when the counter is      %
% Fifteen to leave it in the high state.                                     %
%*****%
FRZ_DATA_OUT = !CNTR[3].q & !CNTR[2].q & !CNTR[1].q & !CNTR[0].q
# GEN_STOP_BITFF.q
# START_SHIFTFFF.q & !CLKFF[0].q;

%*****%
% Read Storage Light Status Register I/O address: 0808                  %
% (MSB)     Bits 7-1 Reserved                                         %
% (LSB)     Bit 0   Hard Disk Active Light          (W/R) %
% %
% Read Power Control Register 1 I/O address: 082A                      %
% (MSB)     Bits 7-1   Reserved                                         %
% (LSB)     Bit 0     83C750 D0          (W/R) %
% %
% Read Power Control Register 2 I/O address: 082B                      %
% (MSB)     Bit 7     CPU1 ROM Completed  (Reset by /RESET, R/W) %
%             Bit 6     Reserved                                         %
%             Bit 5     Reserved                                         %
%             Bit 4     Reserved                                         %
%             Bit 3     IRQ12 Mask        (Reset by /RESET, W/R) %
%             Bits 2-1   Reserved                                         %
% (LSB)     Bit 0     83C750 Status          (R/O) %
% %
%*****%
XD_TRI_OE = ((LIGHT_STRB
# PWR_REG1_STRB # PWR_REG2_STRB
# CLKFF_STRB) & !XIOR);

```

```
XD[0]      = TRI (D[0], XD_TRI_OE);
D[0]      = HDD_LEDFF & LIGHT_STRB
           # RWD0 & PWR_REG1_STRB
           # PROC_RDY & /CMD_STATE & PWR_REG2_STRB
           # CLKFF[0].q & CLKFF_SELL
           # CLKFF[8].q & CLKFF_SELH;

XD[1]      = TRI (D[1], XD_TRI_OE);
D[1]      = CLKFF[1].q & CLKFF_SELL
           # CLKFF[9].q & CLKFF_SELH;
XD[2]      = TRI (D[2], XD_TRI_OE);
D[2]      = CLKFF[2].q & CLKFF_SELL
           # CLKFF[10].q & CLKFF_SELH;
XD[3]      = TRI (D[3], XD_TRI_OE);
D[3]      = PWR_REG2[1].q & PWR_REG2_STRB
           # CLKFF[3].q & CLKFF_SELL
           # CLKFF[11].q & CLKFF_SELH;
XD[4]      = TRI (D[4], XD_TRI_OE);
D[4]      = CLKFF[4].q & CLKFF_SELL
           # CLKFF[12].q & CLKFF_SELH;
XD[5]      = TRI (D[5], XD_TRI_OE);
D[5]      = CLKFF[5].q & CLKFF_SELL;
XD[6]      = TRI (D[6], XD_TRI_OE);
D[6]      = CLKFF[6].q & CLKFF_SELL;
XD[7]      = TRI (D[7], XD_TRI_OE);
D[7]      = PWR_REG2[2].q & PWR_REG2_STRB
           # CLKFF[7].q & CLKFF_SELL;
END;
```

Appendix C Planar Set Up and Registers

This section describes the registers and setup considerations that are a function of the example planar, as opposed to the MCM. These settings and registers are not in general required by the MCM in every system. This information is included only for reference.

Designers of systems containing components similar to those found on the example planar may wish to start their development work using these settings.

C.1 ISA Bridge (SIO) Initialization

The example planar uses an Intel 82378ZB SIO as the ISA bridge.

The SIO chip must be configured prior to any other PCI bus agent. The SIO PCI arbiter is automatically enabled upon power-on reset. During power-on reset, the SIO drives the A/D(31:0), C/BE#(3:0), and par signals on the PCI bus.

The system I/O CPLD uses the decode circuits in the SIO that produce the signals EC-SADDR[2:0] and UBUSCOE# to decode the motherboard register addresses. For this reason, utility bus A and B decode registers must be initialized as shown in Table C-1.

The ISA clock divisor must be set as indicated prior to running any CPU to PCI transactions. If the configuration information is stored in Flash, this should pose no problem.

The SIO must be programmed so that interval timer 1 operates in mode 2 with a period of approximately 15 microseconds. This timer controls the ISA refresh interval. It must be programmed at least 200 microseconds before any access to ISA DRAM is attempted.

PCI memory write cycles destined for ISA can use a 32-bit posted write buffer in the SIO. Bit 2 of the PCI control register controls the enabling of the posted write buffer. The default (power-on reset) state for the posted write buffer is disabled. It is required that the posted write buffer be enabled.

Note that PCI burst transactions are not supported by the SIO. For burst transactions, the SIO will always target abort after the first data phase. The system will not allow the CPU to burst to the SIO (or any other PCI agent). No PCI master should be programmed to attempt burst transactions to the SIO.

The SIO defaults (after power-on reset) to the slow sampling point (bits 4:3 of the PCI Control Register) for its subtractive decode. Of the three choices for the sampling point: slow (5 PCI cycles), typical (4 PCI cycles) and fast (3 PCI cycles), one should be chosen that is one clock after the slowest I/O device on the PCI bus. If the PCI agents are all memory mapped above 16M Byte and all I/O mapped above 64K, then the fast sampling point for

the subtractive decode can be chosen. This insures that no other PCI agent except the SIO will claim these addresses. Configure PCI agents in this manner to improve performance.

The SIO automatically inserts a 4 ISA clock cycle delay between PCI originated back-to-back 8 and 16 bit I/O cycles to the ISA bus. In addition, the ISA Controller Recovery Timer Register (configuration register, address offset=4Ch) enables a number of additional ISA clock cycles of delay to be inserted between these types of back-to-back I/O cycles. The ISA Controller Recovery Timer Register defaults (after power-on reset) to 2 additional ISA clock cycles of delay, making the total delay equal to 6 ISA clock cycles, for both the 8 and 16 bit I/O recovery times. Since none of the native I/O devices on the example planar require such long recovery times, the additional cycles specified by the ISA Controller Timer Register can be disabled. If an ISA card requiring a long recovery time is supported, the driver should insure that the recovery time is met.

Disable scatter/gather mode and GAT mode.

Do not attempt to access DMA channel 4 address and byte count registers.

Always enable the ISA master and DMA buffers. In order to isolate slow ISA Bus I/O devices from the PCI bus, the DMA controller uses the DMA/ISA master Line Buffer. This buffer can operate in single transaction or in 8-byte mode. Bits 0-LE/7-BE and 1-LE/6-BE of the PCI Control register configure the line buffer for DMA and ISA masters separately. It is required that the 8-byte mode be enabled for both (Bits = 1,1).

The registers in Table C-1 must be set in order for the example planar I/O hardware to operate properly. Vendors use LE bit nomenclature, and nomenclature within CPU registers is BE.

Table C-1. Summary of SIO Register Setup (Configuration Address = 8080 08xx)

Register	Addr	Bit	Set To	Reset Value	DESCRIPTION
PCI Control Register	40h	2-LE 5-BE	1	0	Enable PCI Memory Posted Write Buffer.
PCI Control Register	40h	1-LE 6-BE	1	0	Enable ISA Master Line buffer.
PCI Control Register	40h	0-LE 7-BE	1	0	Enable DMA Line Buffer
PCI Arbiter Control (Config/PCI)	41h	0-LE 7-BE	0	0	Disable GAT (Guaranteed Access Time Mode). Note: GAT does not work in SIO.
ISA Clock Divisor (Config/PCI)	4Dh	5-LE 2-BE	0	0	Disable Coprocessor Error Support.
ISA Clock Divisor (Config/PCI)	4Dh	4-LE 3-BE	1	0	Enable IRQ12/M Mouse Support.
ISA Clock Divisor (Config/PCI)	4Dh	3-LE 4-BE	*	0	* This bit should be set to 1 before changing or loading the PCI ISA Clock Divisor value. Setting this bit to 1 will assert the RSTDVR signal (which resets the System I/O EPLD and any devices on the ISA bus slots). All these devices will require reconfiguration after this bit has been asserted. Software must guarantee that RSTDVR be asserted for a minimum of 1 ms after the clock divisor value is set.
ISA Clock Divisor (Config/PCI)	4Dh	2:0-LE 5:7-BE	*	0	* Set this field to 000b (divisor = 4). (If PCI clock is slower than 33 MHz, then this field would be 001b (divisor=3).

Table C-1. Summary of SIO Register Setup (Configuration Address = 8080 08xx) (Continued)

Register	Addr	Bit	Set To	Reset Value	DESCRIPTION
Utility Bus Chip Select A (Config/PCI)	4Eh	4-LE 4-BE	1	0	Disable generation of ECSADDR(2:0) and UBU-SOE# for the IDE and Floppy decode.
Utility Bus Chip Select A (Config/PCI)	4Eh	1-LE 6-BE	1	1	Enable keyboard addresses (60h, 62h, 64h, 66h).
Utility Bus Chip Select A (Config/PCI)	4Eh	0-LE 7-BE	1	1	Enable TOD Addresses (70h, 71h).
Utility Bus Chip Select B (Config/PCI)	4Fh	7-LE 0-BE	1	0	Enable access to the motherboard registers in the 0800-08FF address range.
Utility Bus Chip Select B (Config/PCI)	4Fh	6-LE 1-BE	1	1	Enable Port 92h access.
Utility Bus Chip Select B (Config/PCI)	4Fh	5:4-LE 2:3-BE	11	00	Disable generation of default address for Parallel Port.
Utility Bus Chip Select B (Config/PCI)	4Fh	3:2-LE 4:5-BE	11	00	Disable generation of default address for Serial Port B.
Utility Bus Chip Select B (Config/PCI)	4Fh	1:0-LE 6:7-BE	11	00	Disable generation of default address for Serial Port A.
Interrupt Controller 1 - ICW1 (I/O /PCI)	20h	3-LE 4-BE	0	x	Set Interrupt Controller 1 to edge triggered mode.
Interrupt Controller 1 - ICW1 (I/O /PCI)	20h	1-LE 6-BE	0	x	Set Interrupt Controller 1 to cascade mode.
Interrupt Controller 2 - ICW1 (I/O /PCI)	A0h	3-LE 4-BE	0	x	Set Interrupt Controller 2 to edge triggered mode.
Interrupt Controller 2 - ICW1 (I/O /PCI)	A0h	1-LE 6-BE	0	x	Set Interrupt Controller 2 to cascade mode.
NMI Status and Control (I/O /PCI)	61h	3-LE 4-BE	0	0	IOCHK# NMI enabled.
NMI Status and Control (I/O /PCI)	61h	2-LE 5-BE	0	0	PCI SERR# NMI enabled.
NMI Enable and TOD Address (I/O /PCI)	70h	7-LE 0-BE	0	1	NMI interrupt enabled.
DMA Command (I/O /PCI)	08h, D0h	7-LE 0-BE	0	0	DACK# Assert Level set to low.
DMA Command (I/O /PCI)	08h, D0h	6-LE 1-BE	0	0	DREQ Sense Level set to high.

C.1.1 Summary of SIO Configuration Registers

Table C-2. Summary of SIO Configuration Registers

Address	Description	Type	Reset Value	Set To *
8080 0800	Vendor Identification	R/O	86h	
8080 0801	Vendor Identification	R/O	80	
8080 0802	Device Identification	R/O	84	
8080 0803	Device Identification	R/O	04	
8080 0804	Command	R/W	07	0F
8080 0805	Command	R/W	00	00
8080 0806	Device Status	R/W	00	
8080 0807	Device Status	R/W	02	
8080 0808	Revision Identification	R/W	00	
8080 0840	PCI Control	R/W	20	21
8080 0841	PCI Arbiter Control	R/W	00	00
8080 0842	PCI Arbiter Priority Control	R/W	04	04
8080 0843	PCI Arbiter Priority Control Extension	R/W	00	00
8080 0844	MEMCS# Control	R/W	00	00
8080 0845	MEMCS# Bottom of Hole	R/W	10	10
8080 0846	MEMCS# Top of Hole	R/W	0F	0F
8080 0847	MEMCS# Top of Memory	R/W	00	00
8080 0848	ISA Address Decoder Control	R/W	01	F1
8080 0849	ISA Address Decoder ROM Block	R/W	00	00
8080 084A	ISA Address Bottom of Hole	R/W	10	10
8080 084B	ISA Address Top of Hole	R/W	0F	0F
8080 084C	ISA Controller Recovery Timer	R/W	56	56
8080 084D	ISA Clock Divisor	R/W	40	10
8080 084E	Utility Bus Chip Select A	R/W	07	07
8080 084F	Utility Bus Chip Select B	R/W	4F	FF
8080 0854	MEMCS# Attribute Register #1	R/W	00	
8080 0855	MEMCS# Attribute Register #2	R/W	00	
8080 0856	MEMCS# Attribute Register #3	R/W	00	
8080 0857	Scatter/Gather Relocation Base	R/W	04	
8080 0860	PIRQ Route Control 0	R/W	80	0F
8080 0861	PIRQ Route Control 1	R/W	80	0F
8080 0862	PIRQ Route Control 2	R/W	80	80
8080 0863	PIRQ Route Control 3 (unused)	R/W	80	80
8080 0880	BIOS Timer Base Address	R/W	78	
8080 0881	BIOS Timer Base Address	R/W	00	

Note: * If the entry in this column is blank, then the boot firmware does not write to this register.

C.2 Example Planar Combined Register Listing

C.2.1 Direct Access Registers

Table C-3 contains a summary listing of the registers that are physically located on the example planar. These registers are generally accessed using single CPU transfers. There is an additional set of registers (see Table 9-2) located in the 660 bridge, which are accessed using pairs of CPU transfers.

Table C-3. Combined Register Listing

ISA Port	Contiguous Mode Addr	Non-Contig Mode Addr	Description	R/W	Set To (6)	Loc (4)	Note
0000	8000 0000	8000 0000	DMA1 CH0 Base and Current Addr	R/W	—	SIO	
0001	8000 0001	8000 0001	DMA1 CH0 Base and Current Cnt	R/W	—	SIO	
0002	8000 0002	8000 0002	DMA1 CH1 Base and Current Addr	R/W	—	SIO	
0003	8000 0003	8000 0003	DMA1 CH0 Base and Current Cnt	R/W	—	SIO	
0004	8000 0004	8000 0004	DMA1 CH2 Base and Current Addr	R/W	—	SIO	
0005	8000 0005	8000 0005	DMA1 CH2 Base and Current Cnt	R/W	—	SIO	
0006	8000 0006	8000 0006	DMA1 CH3 Base and Current Addr	R/W	—	SIO	
0007	8000 0007	8000 0007	DMA1 CH3 Base and Current Cnt	R/W	—	SIO	
0008	8000 0008	8000 0008	DMA1 Status(R) Command(W)	R/W	—	SIO	
0009	8000 0009	8000 0009	DMA1 Soft Request	W	—	SIO	
000A	8000 000A	8000 000A	DMA1 Write Single Mask Bit	W	—	SIO	
000B	8000 000B	8000 000B	DMA1 Write Mode	W	—	SIO	
000C	8000 000C	8000 000C	DMA1 Clear Byte Pointer	W	—	SIO	
000D	8000 000D	8000 000D	DMA1 Master Clear	W	—	SIO	
000E	8000 000E	8000 000E	DMA1 Clear Mask	W	—	SIO	
000F	8000 000F	8000 000F	DMA1 R/W All Mask Register Bits	R/W	—	SIO	
0020	8000 0020	8000 1000	INT1 Control	R/W	—	SIO	
0021	8000 0021	8000 1001	INT1 Mask	R/W	—	SIO	
0040	8000 0040	8000 2000	Timer Counter 1 - Counter 0 Cnt	R/W	—	SIO	
0041	8000 0041	8000 2001	Timer Counter 1 - Counter 1 Cnt	R/W	—	SIO	
0042	8000 0042	8000 2002	Timer Counter 1 - Counter 2 Cnt	R/W	—	SIO	
0043	8000 0043	8000 2003	Timer Counter 1 Command Mode	W	—	SIO	
0060	8000 0060	8000 3000	Reset X-Bus (mse) IRQ12 and Kbd	R	—	SIO	
0061	8000 0061	8000 3001	NMI Status and Control	R/W	—	SIO	
0062	8000 0062	8000 3002	Reserved for Keyboard/Mouse	R/W	—	KBD (3)	
0064	8000 0064	8000 3004	Keyboard/Mouse	R	—	KBD (3)	
0066	8000 0066	8000 3006	Reserved for Keyboard/Mouse	R/W	—	KBD (3)	
0070	8000 0070	8000 3010	TOD Addr and NMI Enable	W	—	SIO	
0071	8000 0071	8000 3011	TOD Read/Write	R/W	—	RTC (3)	
0074	8000 0074	8000 3014	NV RAM Addr Strobe 0	W	—	NVR (3)	
0075	8000 0075	8000 3015	NV RAM Addr Strobe 1	W	—	NVR (3)	
0077	8000 0077	8000 3017	NV RAM Data Port	R/W	—	NVR (3)	
0078	8000 0078	8000 3018	BIOS Timer	R/W	—	SIO	
0079	8000 0079	8000 3019	BIOS Timer	R/W	—	SIO	
007A	8000 007A	8000 301A	BIOS Timer	R/W	—	SIO	
007B	8000 007B	8000 301B	BIOS Timer	R/W	—	SIO	
0080	8000 0080	8000 4000	DMA Page Register Reserved	R/W	—	SIO	
0081	8000 0081	8000 4001	DMA Channel 2 Page Register	R/W	—	SIO	
0082	8000 0082	8000 4002	DMA Channel 3 Page Register	R/W	—	SIO	
0083	8000 0083	8000 4003	DMA Channel 1 Page Register	R/W	—	SIO	
0084	8000 0084	8000 4004	DMA Page Register Reserved	R/W	—	SIO	
0085	8000 0085	8000 4005	DMA Page Register Reserved	R/W	—	SIO	
0086	8000 0086	8000 4006	DMA Page Register Reserved	R/W	—	SIO	
0087	8000 0087	8000 4007	DMA Channel 0 Page Register	R/W	—	SIO	
0088	8000 0088	8000 4008	DMA Page Register Reserved	R/W	—	SIO	

Table C-3. Combined Register Listing (Continued)

ISA Port	Contiguous Mode Addr	Non-Contig Mode Addr	Description	R/W	Set To (6)	Loc (4)	Note
0089	8000 0089	8000 4009	DMA Channel 6 Page Register	R/W	—	SIO	
008A	8000 008A	8000 400A	DMA Channel 7 Page Register	R/W	—	SIO	
008B	8000 008B	8000 400B	DMA Channel 5 Page Register	R/W	—	SIO	
008C	8000 008C	8000 400C	DMA Page Register Reserved	R/W	—	SIO	
008D	8000 008D	8000 400D	DMA Page Register Reserved	R/W	—	SIO	
008E	8000 008E	8000 400E	DMA Page Register Reserved	R/W	—	SIO	
008F	8000 008F	8000 400F	DMA Low Page Register Refresh	R/W	—	SIO	
0090	8000 0090	8000 4010	DMA Page Register Reserved	R/W	—	SIO	
0092	8000 0092	8000 4012	Special Port 92 Register	R/W	—	660	
0094	8000 0094	8000 4014	DMA Page Register Reserved	R/W	—	SIO	(2)
0095	8000 0095	8000 4015	DMA Page Register Reserved	R/W	—	SIO	
0096	8000 0096	8000 4016	DMA Page Register Reserved	R/W	—	SIO	
0098	8000 0098	8000 4018	DMA Page Register Reserved	R/W	—	SIO	
009C	8000 009C	8000 401C	DMA Page Register Reserved	R/W	—	SIO	
009D	8000 009D	8000 401D	DMA Page Register Reserved	R/W	—	SIO	
009E	8000 009E	8000 401E	DMA Page Register Reserved	R/W	—	SIO	
009F	8000 009F	8000 401F	DMA Low Page Register Refresh	R/W	—	SIO	
00A0	8000 00A0	8000 5000	INT2 Control Register	R/W	—	SIO	
00A1	8000 00A1	8000 5001	INT2 Mask Register	R/W	—	SIO	
00C0	8000 00C0	8000 6000	DMA2 CH0 Base and Current Addr	R/W	—	SIO	
00C2	8000 00C2	8000 6002	DMA2 CH0 Base and Current Cnt	R/W	—	SIO	
00C4	8000 00C4	8000 6004	DMA2 CH1 Base and Current Addr	R/W	—	SIO	
00C6	8000 00C6	8000 6006	DMA2 CH1 Base and Current Cnt	R/W	—	SIO	
00C8	8000 00C8	8000 6008	DMA2 CH2 Base and Current Addr	R/W	—	SIO	
00CA	8000 00CA	8000 600A	DMA2 CH2 Base and Current Cnt	R/W	—	SIO	
00CC	8000 00CC	8000 600C	DMA2 CH3 Base and Current Addr	R/W	—	SIO	
00CE	8000 00CE	8000 600E	DMA2 CH3 Base and Current Cnt	R/W	—	SIO	
00D0	8000 000D	8000 6010	DMA2 Status(R) Command(W)	R/W	—	SIO	
00D2	8000 00D2	8000 6012	DMA2 Soft Request	W	—	SIO	
00D4	8000 00D4	8000 6014	DMA2 Write Single Mask Bit	W	—	SIO	
00D6	8000 00D6	8000 6016	DMA2 Write Mode	W	—	SIO	
00D8	8000 00D8	8000 6018	DMA2 Clear Byte Pointer	W	—	SIO	
00DA	8000 00DA	8000 601A	DMA2 Master Clear	W	—	SIO	
00DC	8000 00DC	8000 601C	DMA2 Clear Mask	W	—	SIO	
00DE	8000 00DE	8000 601E	DMA2 R/W All Mask Register Bits	R/W	—	SIO	
00F0	8000 00F0	8000 7010	Coprocessor Error Reg - Reserved	R/W	—	SIO	
040B	8000 040B	8002 0006	DMA1 Extended Mode	W	—	SIO	
0410	8000 0410	8002 0010	CH0 Scatter/Gather Command	W	—	SIO	
0411	8000 0411	8002 0011	CH1 Scatter/Gather Command	W	—	SIO	
0412	8000 0412	8002 0012	CH2 Scatter/Gather Command	W	—	SIO	
0413	8000 0413	8002 0013	CH3 Scatter/Gather Command	W	—	SIO	
0415	8000 0415	8002 0015	CH5 Scatter/Gather Command	W	—	SIO	
0416	8000 0416	8002 0016	CH6 Scatter/Gather Command	W	—	SIO	
0417	8000 0417	8002 0017	CH7 Scatter/Gather Command	W	—	SIO	
0418	8000 0418	8002 0018	CH0 Scatter/Gather Status	R	—	SIO	

Table C-3. Combined Register Listing (Continued)

ISA Port	Contiguous Mode Addr	Non-Contig Mode Addr	Description	R/W	Set To (6)	Loc (4)	Note
0419	8000 0419	8002 0019	CH1 Scatter/Gather Status	R	—	SIO	
041A	8000 041A	8002 001A	CH2 Scatter/Gather Status	R	—	SIO	
041B	8000 041B	8002 001B	CH3 Scatter/Gather Status	R	—	SIO	
041D	8000 041D	8002 001D	CH5 Scatter/Gather Status	R	—	SIO	
041E	8000 041E	8002 001E	CH6 Scatter/Gather Status	R	—	SIO	
041F	8000 041F	8002 001F	CH7 Scatter/Gather Status	R	—	SIO	
0420	8000 0420	8002 1000	CH0 Scatter/Gather Pointer	R/W	—	SIO	
0421	8000 0421	8002 1001	CH0 Scatter/Gather Pointer	R/W	—	SIO	
0422	8000 0422	8002 1002	CH0 Scatter/Gather Pointer	R/W	—	SIO	
0423	8000 0423	8002 1003	CH0 Scatter/Gather Pointer	R/W	—	SIO	
0424	8000 0424	8002 1004	CH1 Scatter/Gather Pointer	R/W	—	SIO	
0425	8000 0425	8002 1005	CH1 Scatter/Gather Pointer	R/W	—	SIO	
0426	8000 0426	8002 1006	CH1 Scatter/Gather Pointer	R/W	—	SIO	
0427	8000 0427	8002 1007	CH1 Scatter/Gather Pointer	R/W	—	SIO	
0428	8000 0428	8002 1008	CH2 Scatter/Gather Pointer	R/W	—	SIO	
0429	8000 0429	8002 1009	CH2 Scatter/Gather Pointer	R/W	—	SIO	
042A	8000 042A	8002 100A	CH2 Scatter/Gather Pointer	R/W	—	SIO	
042B	8000 042B	8002 100B	CH2 Scatter/Gather Pointer	R/W	—	SIO	
042C	8000 042C	8002 100C	CH3 Scatter/Gather Pointer	R/W	—	SIO	
042D	8000 042D	8002 100D	CH3 Scatter/Gather Pointer	R/W	—	SIO	
042E	8000 042E	8002 100E	CH3 Scatter/Gather Pointer	R/W	—	SIO	
042F	8000 042F	8002 100F	CH3 Scatter/Gather Pointer	R/W	—	SIO	
0434	8000 0434	8002 1014	CH5 Scatter/Gather Pointer	R/W	—	SIO	
0435	8000 0435	8002 1015	CH5 Scatter/Gather Pointer	R/W	—	SIO	
0436	8000 0436	8002 1016	CH5 Scatter/Gather Pointer	R/W	—	SIO	
0437	8000 0437	8002 1017	CH5 Scatter/Gather Pointer	R/W	—	SIO	
0438	8000 0438	8002 1018	CH6 Scatter/Gather Pointer	R/W	—	SIO	
0439	8000 0439	8002 1019	CH6 Scatter/Gather Pointer	R/W	—	SIO	
043A	8000 043A	8002 101A	CH6 Scatter/Gather Pointer	R/W	—	SIO	
043B	8000 043B	8002 101B	CH6 Scatter/Gather Pointer	R/W	—	SIO	
043C	8000 043C	8002 101C	CH7 Scatter/Gather Pointer	R/W	—	SIO	
043D	8000 043D	8002 101D	CH7 Scatter/Gather Pointer	R/W	—	SIO	
043E	8000 043E	8002 101E	CH7 Scatter/Gather Pointer	R/W	—	SIO	
043F	8000 043F	8002 101F	CH7 Scatter/Gather Pointer	R/W	—	SIO	
0481	8000 0481	8002 4001	DMA CH2 High Page	R/W	—	SIO	
0482	8000 0482	8002 4002	DMA CH3 High Page	R/W	—	SIO	
0483	8000 0483	8002 4003	DMA CH1 High Page	R/W	—	SIO	
0487	8000 0487	8002 4007	DMA CH0 High Page	R/W	—	SIO	
0489	8000 0489	8002 4009	DMA CH6 High Page	R/W	—	SIO	
048A	8000 048A	8002 400A	DMA CH7 High Page	R/W	—	SIO	
048B	8000 048B	8002 400B	DMA CH5 High Page	R/W	—	SIO	
04D0	8000 04D0	8002 6010	Interrupt Control 1	R/W	—	SIO	
04D1	8000 04D1	8002 6011	Interrupt Control 2	R/W	—	SIO	
04D6	8000 04D6	8002 6016	DMA2 Extended Mode	W	—	SIO	
0808	8000 0808	8004 0008	HDD Light	R/W	—	EPLD	

Table C-3. Combined Register Listing (Continued)

ISA Port	Contiguous Mode Addr	Non-Contig Mode Addr	Description	R/W	Set To (6)	Loc (4)	Note
080C	8000 080C	8004 000C	Equipment Present	R	—	logic	(3)
080D	8000 080D	8004 000D	L2 Cache Status Reg	R	—	logic	(3)
0814	8000 0814	8004 0014	L2 Flush	W	—	660	
081C	8000 081C	8004 001C	System Control 81C	R/W	(7)	660	
0821	8000 0821	8004 1001	Memory Controller Misc	R/W	—	660	
082A	8000 082A	8004 100A	Power Mgmt Control Reg1	R/W	—	EPLD	(5)
082B	8000 082B	8004 100B	Power Mgmt Control Reg2	R/W	—	EPLD	(5)
0840	8000 0840	8004 2000	Memory Parity Error Status	R	—	660	
0842	8000 0842	8004 2002	L2 Error Status	R	—	660	
0843	8000 0843	8004 2003	L2 Parity Read & Clear	R	—	660	
0844	8000 0844	8004 2004	Unsupported Transfer Type Error	R	—	660	
0850	8000 0850	8004 2010	I/O Map Type	W	—	660	
0852	8000 0852	8004 2012	Board ID	R	—	logic	(3)
0860	8000 0860	8004 3000	Freeze Clock Reg Low	R/W	—	EPLD	(3)
0862	8000 0862	8004 3002	Freeze Clock Reg High	R/W	—	EPLD	(3)
0880	8000 0880	8004 4000	SIMM Presence Detect Slot 1/2	R	—	logic	(3)
0881	8000 0881	8004 4001	SIMM Presence Detect Slot 3/4	R	—	logic	(3)
	8000 0CF8		PCI/BCR Configuration Address	R/W	—	660	
	8000 0CF8		PCI/BCR Configuration Data	R/W	—	660	
	8080 08xx		PCI Type 0 Configuration Addr	R/W	—	660	
	8080 0800		Vendor Identification	R	—	SIO	
	8080 0801		Vendor Identification	R	—	SIO	
	8080 0802		Device Identification	R	—	SIO	
	8080 0803		Device Identification	R	—	SIO	
	8080 0804		Command	R/W	0F	SIO	
	8080 0805		Command	R/W	00	SIO	
	8080 0806		Device Status	R/W	—	SIO	
	8080 0807		Device Status	R/W	—	SIO	
	8080 0808		Revision Identification	R/W	—	SIO	
	8080 0840		PCI Control	R/W	21	SIO	
	8080 0841		PCI Arbiter Control	R/W	00	SIO	
	8080 0842		PCI Arbiter Priority Control	R/W	04	SIO	
	8080 0843		PCI Arbiter Priority Control Extension	R/W	00	SIO	
	8080 0844		MEMCS# Control	R/W	00	SIO	
	8080 0845		MEMCS# Bottom of Hole	R/W	10	SIO	
	8080 0846		MEMCS# Top of Hole	R/W	0F	SIO	
	8080 0847		MEMCS# Top of Memory	R/W	00	SIO	
	8080 0848		ISA Address Decoder Control	R/W	F1	SIO	
	8080 0849		ISA Address Decoder ROM Block	R/W	00	SIO	
	8080 084A		ISA Address Bottom of Hole	R/W	10	SIO	
	8080 084B		ISA Address Top of Hole	R/W	0F	SIO	
	8080 084C		ISA Controller Recovery Timer	R/W	56	SIO	
	8080 084D		ISA Clock Divisor	R/W	10	SIO	
	8080 084E		Utility Bus Chip Select A	R/W	07	SIO	
	8080 084F		Utility Bus Chip Select B	R/W	FF	SIO	

Table C-3. Combined Register Listing (Continued)

ISA Port	Contiguous Mode Addr	Non-Contig Mode Addr	Description	R/W	Set To (6)	Loc (4)	Note
	8080 0854		MEMCS# Attribute Register #1	R/W	—	SIO	
	8080 0855		MEMCS# Attribute Register #2	R/W	—	SIO	
	8080 0856		MEMCS# Attribute Register #3	R/W	—	SIO	
	8080 0857		Scatter/Gather Relocation Base	R/W	—	SIO	
	8080 0860		PIRQ Route Control 0	R/W	0F	SIO	
	8080 0861		PIRQ Route Control 1	R/W	0F	SIO	
	8080 0862		PIRQ Route Control 2	R/W	80	SIO	
	8080 0863		PIRQ Route Control 3 (unused)	R/W	80	SIO	
	8080 0880		BIOS Timer Base Address	R/W	—	SIO	
	8080 0881		BIOS Timer Base Address	R/W	—	SIO	
	8080 10xx		PCI Type 0 Configuration Addr	R/W	—	660	
	8080 20xx		PCI Type 0 Configuration Addr	R/W	—	660	
	8080 40xx		PCI Type 0 Configuration Addr	R/W	—	660	
	8080 80xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	8081 00xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	8082 00xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	8084 00xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	8088 00xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	8090 00xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	80A0 00xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	80C0 00xx		PCI Type 0 Configuration Addr	R/W	—	660	(5)
	BFFF EFFF	BFFF EFFF	System Error Addr	R	—	660	(5)
	BFFF FFFF	BFFF FFFF	Interrupt Vector	R	—	660	
	FFFF FFFF	FFFF FFFF	Flash Write Addr/Data	W	—	660	
	FFFF FFF1	FFFF FFF1	Flash Lock Out	W	—	660	

Notes:

1. The first 5 hex digits in the contiguous and non-contiguous mode columns represent the memory page number for which the protection attributes may be set in contiguous I/O mode. That is, devices having the same first five digits in this column will have the same attributes in the memory page table.
2. Port 94 may be used by certain video controllers (e.g. Weitek™ 9100). The SIO chip positively decodes this port. Therefore bus contention may arise when both devices claim the PCI cycle to this port address. Bus contention results in invalid data and possibly harm to the hardware.
3. The control signals for these ports are partially decoded by the SIO. The System I/O CPLD completes the decodes, and issues control signals to the registers, which are usually X-bus buffers.
4. KBD = Keyboard / Mouse Controller
RTC = Real Time Clock, also known as the TOD (Time Of Day clock)
NVR = Non-Volatile RAM, in the same package as the RTC
660 = The 660 Bridge.
5. Not used.
6. In the Set To column, a long dash — means that the initialization firmware does not write to this register. The register is either not used, not written to, or the value of it depends on changing circumstances.
If the word Memory appears, please refer to the System Memory section of the 660 User's Manual.
7. Set register 81C to C0h if an L2 is installed, else leave at reset value.

C.3 ISA Bus Register Suggestions

The following port assignments are designed to be compatible with the example planar firmware and Super I/O type chips. These registers and functions are not implemented on the example planar motherboard.



Table C-4. Compatible ISA Ports (Not on Reference Board)

ISA Port	Contiguous Mode Addr	Non-Contig Mode Addr	Description	Note	R/W
01F0	8000 01F0	8000 F010	IDE Data		R/W
01F1	8000 01F1	8000 F011	IDE Error/Features		R/W
01F2	8000 01F2	8000 F012	IDE Sector Count		R/W
01F3	8000 01F3	8000 F013	IDE Sector Number		R/W
01F4	8000 01F4	8000 F014	IDE Cylinder Low		R/W
01F5	8000 01F5	8000 F015	IDE Cylinder High		R/W
01F6	8000 01F6	8000 F016	IDE Drive Head		R/W
01F7	8000 01F7	8000 F017	IDE Status/Command		R/W
0278	8000 0278	8001 3018	Parallel Port 2	(1)	R/W
0279	8000 0279	8001 3019	Parallel Port 2		R/W
027A	8000 027A	8001 301A	Parallel Port 2		R/W
027B	8000 027B	8001 301B	Parallel Port 2		R/W
027C	8000 027C	8001 301C	Parallel Port 2		R/W
027D	8000 027D	8001 301D	Parallel Port 2		R/W
02F8	8000 02F8	8001 7018	Serial Port 2		R/W
02F9	8000 02F9	8001 7019	Serial Port 2		R/W
02FA	8000 02FA	8001 701A	Serial Port 2		R/W
02FB	8000 02FB	8001 701B	Serial Port 2		R/W
02FC	8000 02FC	8001 701C	Serial Port 2		R/W
02FD	8000 02FD	8001 701D	Serial Port 2		R/W
02FE	8000 02FE	8001 701E	Serial Port 2		R/W
02FF	8000 02FF	8001 701F	Serial Port 2		R/W
0370	8000 0370	8001 B010	Secondary Floppy Digital Output		W
0371	8000 0371	8001 B011	Secondary Floppy Digital Output		W
0372	8000 0372	8001 B012	Secondary Floppy Digital Output		W
0373	8000 0373	8001 B013	Secondary Floppy Digital Output		W
0374	8000 0374	8001 B014	Secondary Floppy Digital Output		W
0375	8000 0375	8001 B015	Secondary Floppy Digital Output		W
0376	8000 0376	8001 B016	Secondary Floppy Digital Output		W
0377	8000 0377	8001 B017	Secondary Floppy Digital Output		W
0376	8000 0376	8001 B016	Secondary IDE Alt Status/Device Ctl		R/W
0377	8000 0377	8001 B017	Secondary IDE drive address		W
0378	8000 0378	8001 B018	Parallel Port 1	(1)	W
0379	8000 0379	8001 B019	Parallel Port 1		W
037A	8000 037A	8001 B01A	Parallel Port 1		W
037B	8000 037B	8001 B01B	Parallel Port 1		W
037C	8000 037C	8001 B01C	Parallel Port 1		W
037D	8000 037D	8001 B01D	Parallel Port 1		W
0398	8000 0398	8001 C018	Super I/O Index Address		R/W
0399	8000 0399	8001 C019	Super I/O Data Address		R/W
03BC	8000 03BC	8001 D01C	Parallel Port 3		R/W
03BD	8000 03BD	8001 D01D	Parallel Port 3		R/W
03BE	8000 03BE	8001 D01E	Parallel Port 3		R/W
03F0	8000 03F0	8001 F010	Primary Floppy Digital Output (Media Sense)		W/O
03F1	8000 03F1	8001 F011	Primary Floppy Digital Output		W/O

Table C-4. Compatible ISA Ports (Not on Reference Board) (Continued)

ISA Port	Contiguous Mode Addr	Non-Contig Mode Addr	Description	Note	R/W
03F2	8000 03F2	8001 F012	Primary Floppy Digital Output		W
03F3	8000 03F3	8001 F013	Primary Floppy Digital Output (Also Media Sense)		W
03F4	8000 03F4	8001 F014	Primary Floppy Digital Output		W
03F5	8000 03F5	8001 F015	Primary Floppy Digital Output		W
03F6	8000 03F6	8001 F016	Primary Floppy Digital Output		W
03F7	8000 03F7	8001 F017	Primary Floppy Digital Output		W
03F6	8000 03F6	8001 F016	Primary IDE Alt Status/Device Ctl		R/W
03F7	8000 03F7	8001 F017	Primary IDE Drive Address		R
03F8	8000 03F8	8001 F018	Serial Port 1		R/W
03FA	8000 03FA	8001 F01A	Serial Port 1		R/W
03FB	8000 03FB	8001 F01B	Serial Port 1		R/W
03FC	8000 03FC	8001 F01C	Serial Port 1		R/W
03FD	8000 03FD	8001 F01D	Serial Port 1		R/W
03FE	8000 03FE	8001 F01E	Serial Port 1		R/W
03FF	8000 03FF	8001 F01F	Serial Port 1		R/W

Note:

1. This is a preferred location for this function.

Appendix D

Planar Electromechanical

D.1 Electrical

D.1.1 Power Requirements

This section sets out the power supply requirements for the example planar. They are achievable with low-cost PC power supplies (see Table D-1 for specifications and Table D-2 for approximate power consumption).

Table D-1. Power Supply Specification

Output	Tolerance	Maximum Ripple P-P
+5V	+5% -4%	50 mV
+12V	+5% -5%	120 mV
-12V	+10% -9%	120 mV
-5V* (1)	+10% -10%	120 mV
3.3V (2)	+5% -4%	50 mV

Table D-2. Approximate Power Consumption

Element	+5V (Amp)	+12V (Amp)	-12V (Amp)	-5V (Amp) (1)	3.3V (Amp) (2)
Base motherboard with 1 8M SIMM running typical code	3.0	0.1	0.02	0	0
8M DRAM standby/refresh (each)	0.18	0	0	0	0
32M DRAM standby/refresh (each)	0.16	0	0	0	0
256K SRAM L2 Cache	1.6	0	0	0	0
Each PCI Slot – allocation (3)	3.0 max	0.3 max	0.06 max	0	3.3 max
Each ISA Slot – allocation (3)(4)	4.5 max	1.5 max	0.3 max	0.2 max	0
Main Power Connector Capacity (at 5.0 amp./pin—20 amp. DC return)	20.0	5.0	5.0	0.5	0

Notes for Table D-1 and Table D-2:

1. The -5v is not used on the planer, but it is routed from the power supply connector to the ISA slots.
2. The 3.3v to the PCI bus slots comes from the power supply connector. The 3.3v for the CPU and the 660 bridge is a separate supply, generated from the +5v supply by a linear regulator.
3. These power requirements are allocated by the system designer. The currents specified per slot are also the maximum currents which may be consumed by the bus. Because of this, the total currents for all the cards used for either the PCI slot or the ISA slot must not exceed the amounts listed for those slots.
4. ISA slot information is taken from IEEE P996.

Other requirements are:

1. Overshoot on any supply voltage must be less than 10% of nominal and must decay to within the regulation band within 50 msec.
2. In any failure situation, the power supply must shut down before the +5v output reaches 6.5V to give the motherboard a reasonable chance of surviving; however, damage may occur at any voltage above 5.5V.
3. Power_Good Signal Requirements
 - The signal must be at a TTL down level when power is applied until >100 msec to 500 msec after the 5V supply has reached its minimum regulation level, and at TTL high level thereafter as long as outputs are within regulation.
 - At turn-off, the Power Good signal must drop to a TTL low level before any output drops below its regulation limits.
 - The driver must be capable of driving 400 microamps or sinking 5 milliamps. The rise time/fall time must be less than 1 usec., 10—90%
4. The +5V rise time (10%-90%) shall be 3 msec to 100 msec with a maximum slope of 0.75 volts/msec for voltages above 1.5 volts for all loadings.
5. All supply voltages shall track within 50 msec of each other measured at the 50% point.

D.1.2 Onboard 3.3V Regulator

There is a 3.3 volt regulator circuit on the example planar to support the CPU and the 660 Bridge. Note that the 3.3v on the PCI bus slots is sourced via the power supply connector, not the onboard 3.3v regulator.

Table D-3. Specifications for 3.3V Regulator on the Motherboard

Specification	Value
Output Voltage	3.3 V \pm 3%
Output Current	0.01 A to 5 A
Input voltage	4.75 V to 5.25 V
Pass element maximum case temperature	110 °C
Tracking	In regulation <1ms after +5 reaches 4.75V.
Overcurrent	No current limit feature

D.2 Physical Design Rules

These guidelines are given to aid designers with the physical design phase of their 100 MHz PPC 603e MCM planar. The guidelines are not intended to replace good physical design practices for the signal types and frequencies discussed, but to supplement standard practice by pointing out sensitive and critical areas. Some discussion of the IBM implementation of the example planar is also included to establish the context for the wiring guidelines.

The example planar was developed by adapting the Harley Reference Design to meet the needs of the MCM. The physical design rules for the example planar are identical to those for Harley except for any changes which are contained in the following areas.

D.2.1 Construction

The general construction of the example planar is shown in Figure D-1. It is constructed with two high frequency signal layers in the center, two power planes, and two external general purpose signal layers.

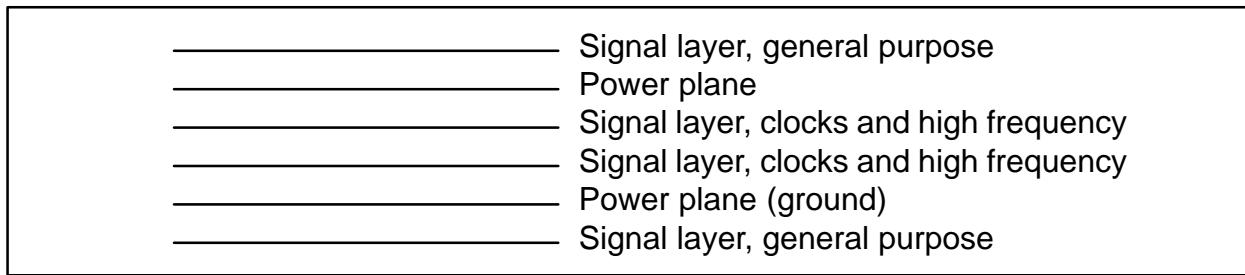


Figure D-1. Signal and Power Layers

A top view of a typical wiring channel, as implemented on the example planar, is shown in Figure D-2 (all dimensions are shown in inches). Minimum trace width is .006" (at 1:1) and minimum space width is .004" (at 1:1).

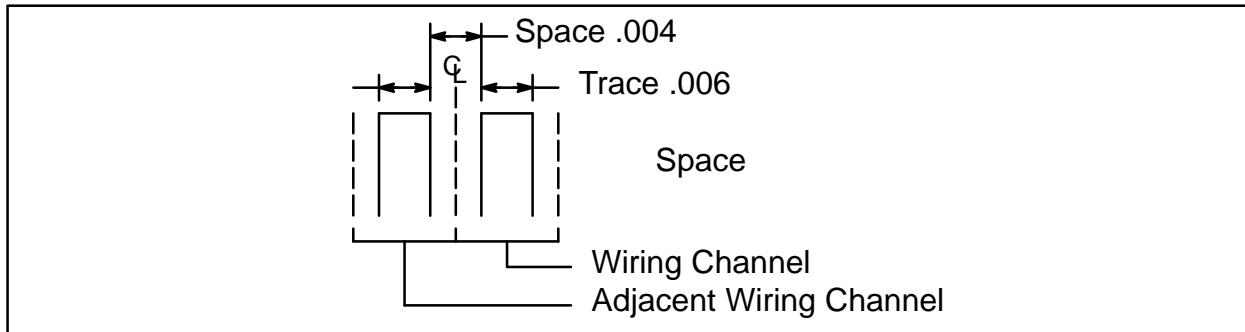
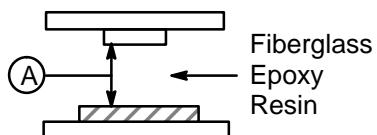


Figure D-2. Typical Wiring Channel Top View

Signal		0.5 oz. Foil
		0.0045 (A)
Power		2 oz. Foil
		0.0062 (A)
Signal		1 oz. Foil
		0.00288 (A)
Signal		1 oz. Foil
		0.0062 (A)
Ground		2 oz. Foil
		0.0045 (A)
Signal		0.5 oz. Foil

Notes:

- (B) All dimensions in inches
- (C) Trace to ground impedance $70 \Omega \pm 10\%$
- (D) Finished signal trace width: 0.00425
- (E) Overall thickness: 0.060
after plating: 0.062
- (F) Plating 7070

Figure D-3. PowerPC 603/604 Board Fabrication

D.2.2 General Wiring Guidelines

- 1 A power (ground or voltage) plane split occurs where there is a discontinuity in the plane. As shown in Figure D-4, route wires across splits in a power plane in the most perpendicular manner possible. Do not run wires parallel to the split in close proximity to the split.

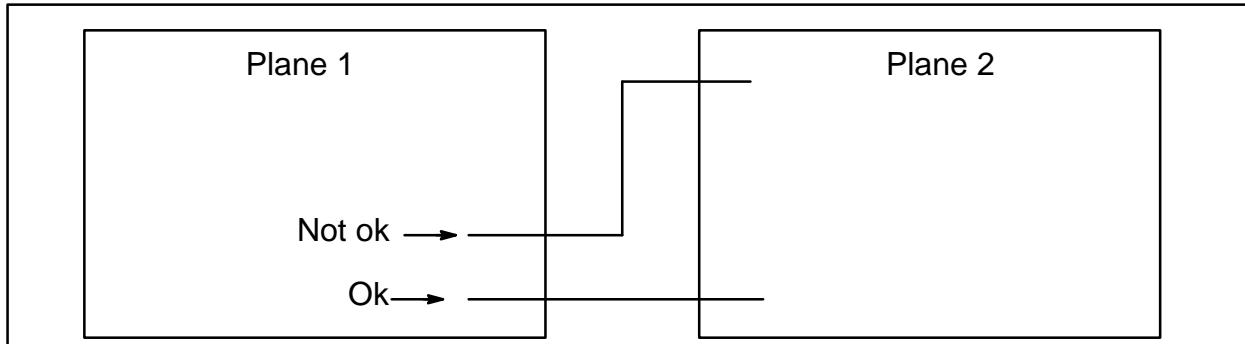


Figure D-4. Power Plane Split

- 2 Minimize the number of wires on the top surface of the board that cross under the clock generator. Do not run any wires close to the crystal connections.
- 3 If a wire is routed near a via which is part of a clock net, there must be at least one vacant wiring channel between the wire and the clock via.
- 4 Several groups of nets require special attention to ensure correct system operation. They are listed below in order of the importance of meeting the design rules that are suggested for that group. For example, ensuring that the clock nets are routed according to the design rules suggested in Section D.3 is more important than ensuring that the PCI bus nets are routed according to the design rules suggested in Section D.4. The ideal system design will meet all of the suggested design rules, but this information is included to guide the designer in case some tradeoffs have to be evaluated.
 - Clock nets.
 - CPU bus nets and timing critical nets.
 - PCI bus nets.
 - Noise sensitive nets.
 - Other nets.

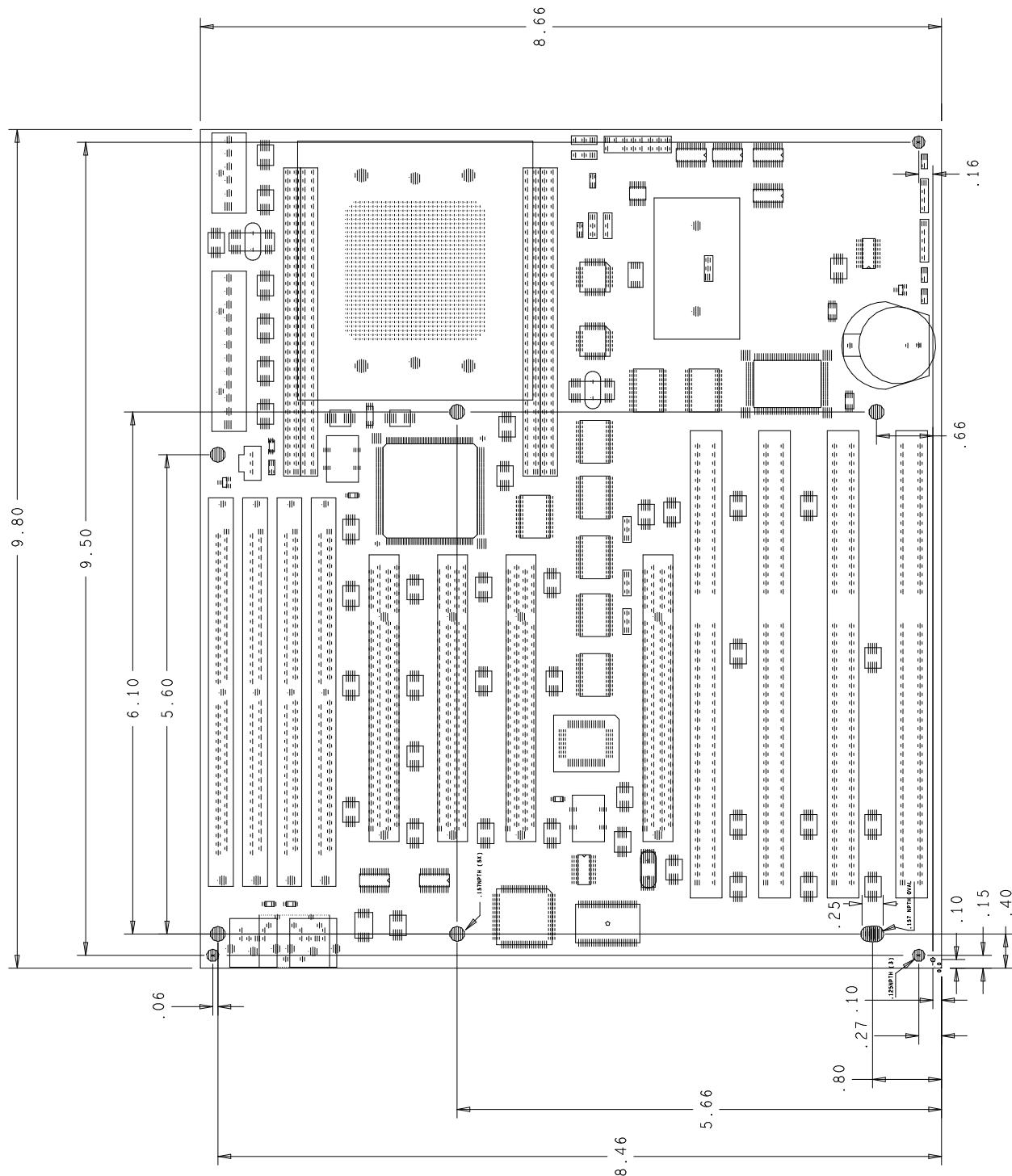
D.3 CPU Bus Nets

- 1 Make the CPU bus nets either 2-node nets or daisy-chained. Stubs and star fanouts are not allowed.
- 2 Route nets so as to minimize noise reception. Make these nets as short as possible.

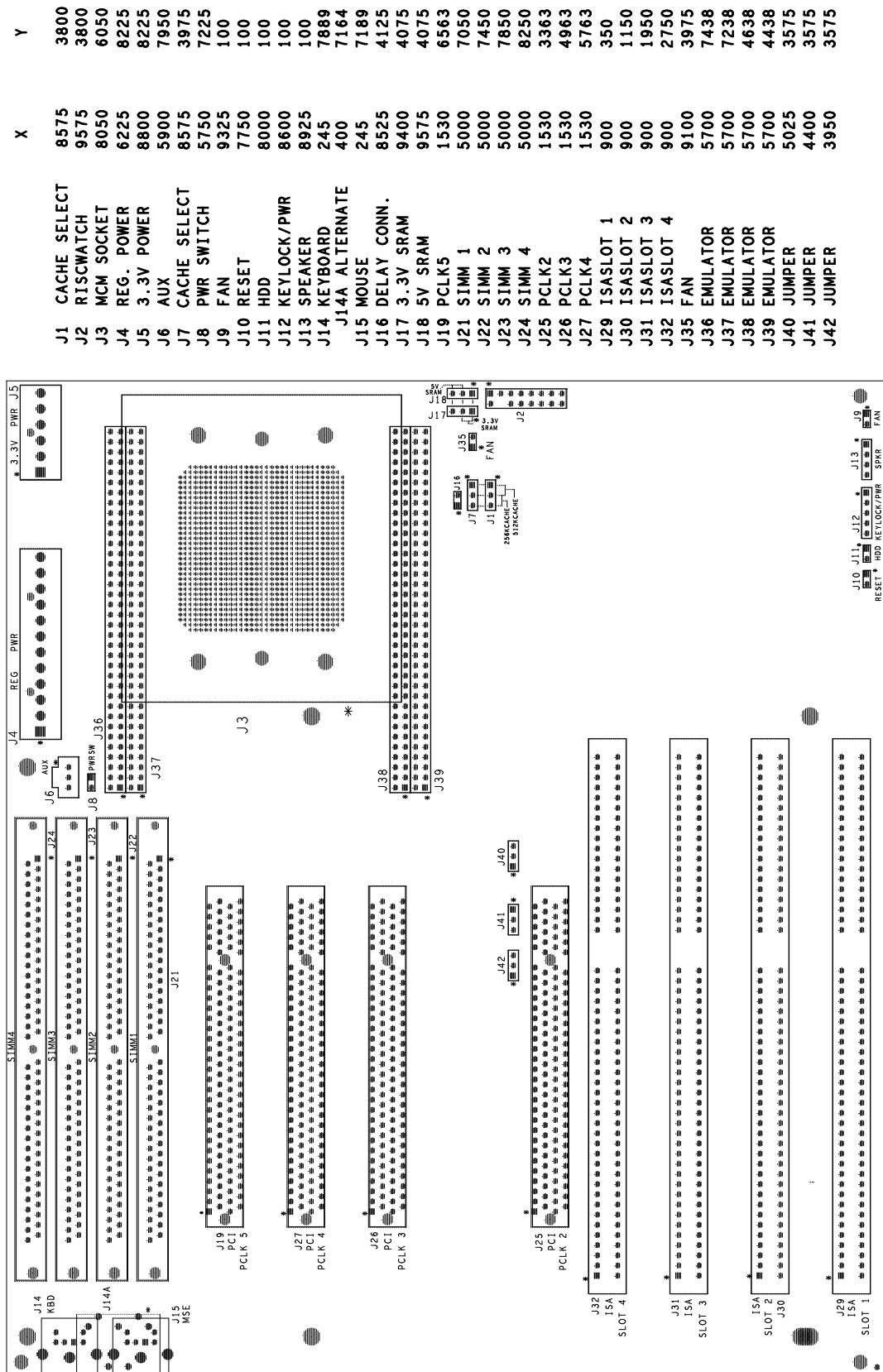
D.4 PCI Bus Nets

- 1 Daisy chain the PCI bus nets. Stubs and star fanouts are not allowed.
- 2 Route these nets so as to minimize noise reception and timing delays. Make these nets as short as possible.

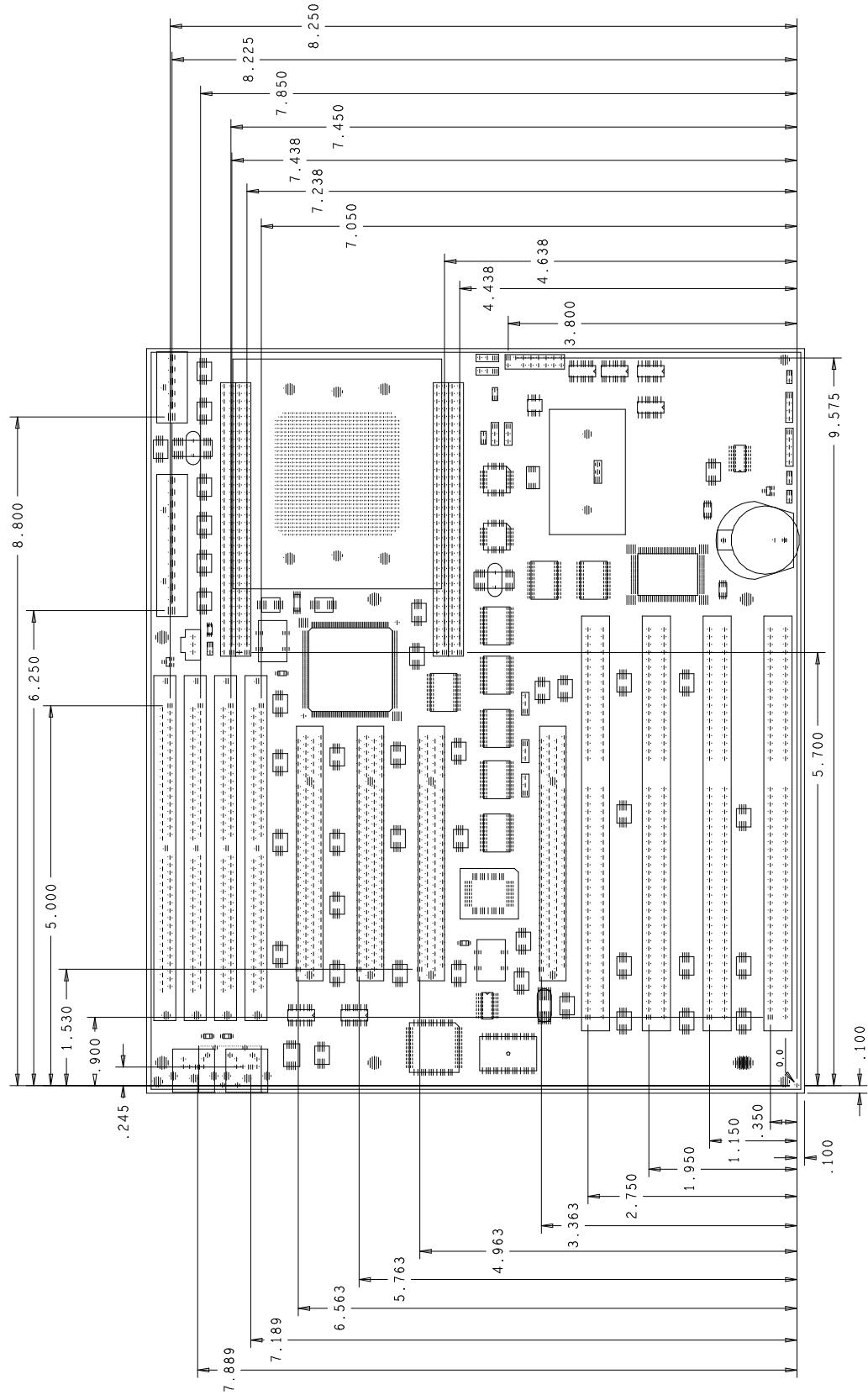
D.5 Example Planar Mechanical



D.5.1 Example Planar Connectors



D.5.2 Example Planar Connector Locations



D.5.3 Keyboard Connector J14

The keyboard connector uses a 6-pin miniature DIN connector (see Figure D-5) Pins are assigned as shown in Table D-4 (as viewed from the back of the machine).

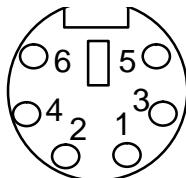


Figure D-5.The Keyboard Connector

Table D-4. Keyboard Connector Pin Assignments

Pin	I/O	Signal Name
1	I/O	DATA
2	NA	n/c
3	NA	GROUND
4	NA	+ 5V DC
5	I/O	CLOCK
6	NA	n/c
M1,2,3	NA	Shield Ground

D.5.4 Alternate Keyboard Connector J14A

The alternate keyboard connector uses a 5-pin miniature DIN connector (see Figure D-6) Pins are assigned as shown in Table D-5 (as viewed from the back of the machine).

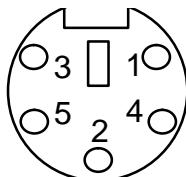


Figure D-6.The Alternate Keyboard Connector

Table D-5. Alternate Keyboard Connector Pin Assignments

Pin	Signal Name
1	CLOCK
2	DATA
3	n/c
4	GROUND
5	+ 5V DC
M1,2,3	Shield Ground

D.5.5 Mouse Connector J15

The mouse connector uses a 6-pin miniature DIN connector (see Figure D-7). Pins are assigned as shown in Table D-6 (as viewed from the back of the machine).

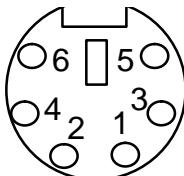


Figure D-7. The Mouse Connector

Table D-6. Mouse Connector Pin Assignments

Pin	I/O	Signal Name
1	I/O	DATA
2	NA	n/c
3	NA	GROUND
4	NA	+ 5V DC
5	I/O	CLOCK
6	NA	n/c
M1,2,3	NA	Shield Ground

D.5.6 Speaker Connector J13

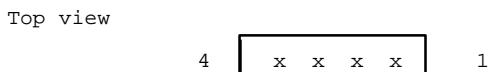
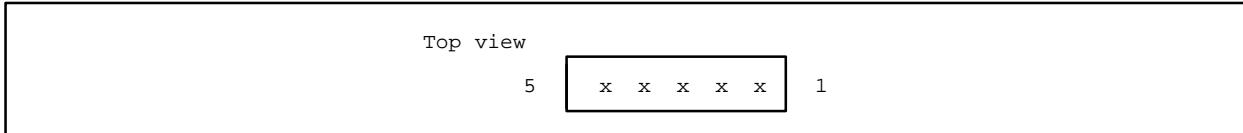


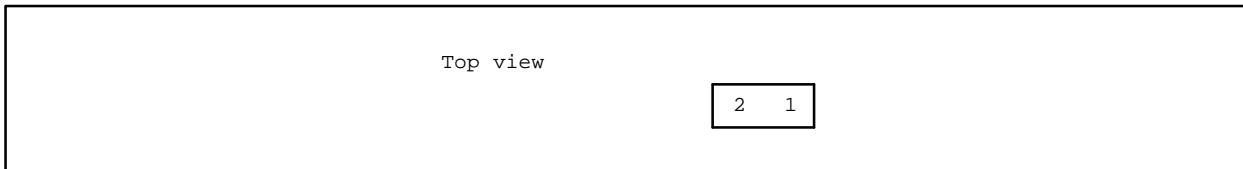
Figure D-8.1x4 Speaker Connector

Table D-7. Speaker Connector Pin Assignments

Pin No.	Signal Name
1	MINUS (-) INPUT TO SPEAKER
2	NO CONNECT
3	NO CONNECT
4	GND

D.5.7 Power Good LED/KEYLOCK# Connector J12**Figure D-9.1x5 Power Good LED Connector****Table D-8. Power Good LED Connector**

Pin No.	Signal Name
1	LED_POWER_GOOD/RESET#
2	NO CONNECT
3	GND
4	KEYLOCK* (Not Used)
5	GND

D.5.8 HDD LED Connector J11 (1 x 2 Berg)**Figure D-10.1x2 HDD LED Connector****Table D-9. HDD LED Connector**

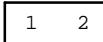
Pin No.	Signal Name
1	+ LED VOLTAGE
2	HDD LED drive signal (low to drive LED)

D.5.9 Reset Switch Connector J10 (1 x 2 Berg)**Figure D-11.1x2 Reset Switch Connector****Table D-10. Reset Switch Connector**

Pin No.	Signal Name
1	RESET N/O CONTACT (close to reset)
2	GROUND

D.5.10 Fan Connector J9, J35

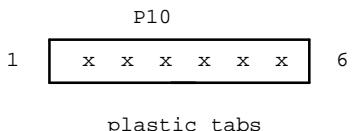
Top view

**Figure D-12.1x2 Fan Connector****Table D-11. Fan Connector Pin Assignments**

Pin No.	Signal Name
1	+12 VOLTS
2	GROUND

D.5.11 3.3V Power Connector J5

Top view

**Figure D-13.1x6 3.3V Power Connector J5****Table D-12. 3.3V Power Connector J5 Pin Assignments**

Pin No.	Signal Name
1	+3.3 V
2	+3.3 V
3	+3.3 V
4	GROUND
5	GROUND
6	GROUND

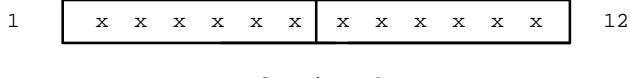
D.5.12 Power Connector J4

Top view

open side

P1

P2

**Figure D-14.1x12 Power Connector****Table D-13. Power Connector J4 Pin Assignments**

Pin No.	Signal Name
1	POWER GOOD
2	+5 VOLTS
3	+12 VOLTS
4	-12 VOLTS
5, 6, 7, 8	GROUND
9, 10, 11, 12	-5 VOLTS

D.5.13 AUX5/ON-OFF Connector J6

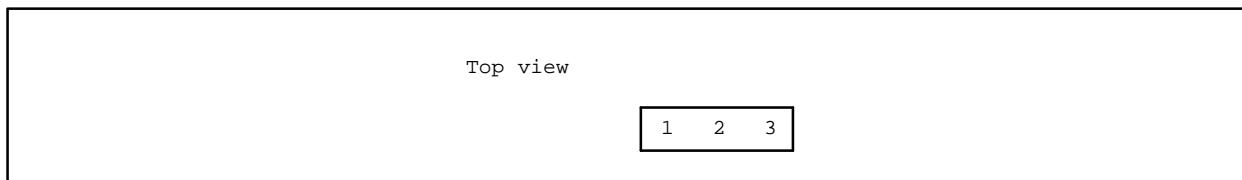


Figure D-15.AUX5/ON-OFF Connector

Table D-14. AUX5/ON-OFF Connector Pin Assignments

Pin No.	Signal Name
1	AUX +5V
2	ON/OFF
3	GROUND

D.5.14 PCI Connectors J19, J25, J26, and J27

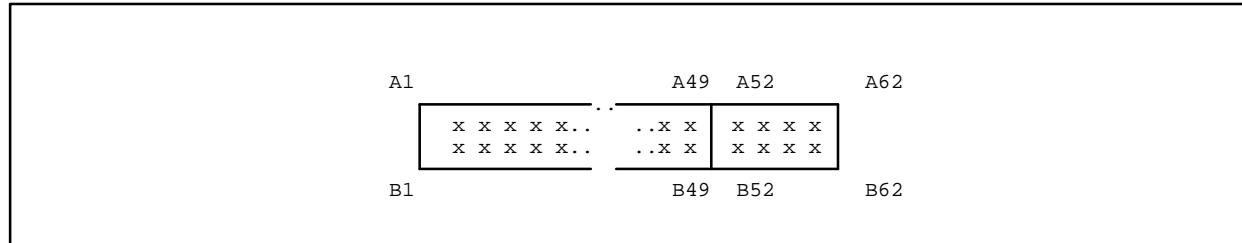


Figure D-16.PCI Connector

Table D-15. PCI Connector Pin Assignments

Pin	Function	Pin	Function
A1	TRST#	B1	-12 VOLTS#
A2	+12 VOLTS	B2	TCK
A3	TMS	B3	GROUND
A4	TDI	B4	TDO
A5	+5 VOLTS	B5	+5 VOLTS
A6	INTA#	B6	+5 VOLTS
A7	INTC#	B7	INTB#
A8	+5 VOLTS	B8	INTD#
A9	RESERVED	B9	PRESENT 1# *
A10	+5 VOLTS	B10	RESERVED
A11	RESERVED	B11	PRESENT 2# *
A12	GROUND	B12	GROUND
A13	GROUND	B13	GROUND
A14	RESERVED	B14	RESERVED
A15	RESET#	B15	GROUND
A16	+5 VOLTS	B16	CLK
A17	GNT#	B17	GROUND
A18	GROUND	B18	REQ#
A19	RESERVED	B19	+5 VOLTS
A20	A/D(30)	B20	A/D(31)

Table D-15. PCI Connector Pin Assignments (Continued)

Pin	Function	Pin	Function
A21	+3.3V	B21	A/D(29)
A22	A/D(28)	B22	GROUND
A23	A/D(26)	B23	A/D(27)
A24	GROUND	B24	A/D(25)
A25	A/D(24)	B25	+3.3V
A26	IDSEL	B26	C/BE#(3)
A27	+3.3V	B27	A/D(23)
A28	A/D(22)	B28	GROUND
A29	A/D(20)	B29	A/D(21)
A30	GROUND	B30	A/D(19)
A31	A/D(18)	B31	+3.3V
A32	A/D(16)	B32	A/D(17)
A33	+3.3V	B33	C/BE#(2)
A34	FRAME#	B34	GROUND
A35	GROUND	B35	IRDY#
A36	TRDY#	B36	+3.3V
A37	GROUND	B37	DEVSEL#
A38	STOP#	B38	GROUND
A39	+3.3V	B39	LOCK#
A40	SDONE	B40	PERR#
A41	SBO#	B41	+3.3V
A42	GROUND	B42	SERR#
A43	PAR	B43	+3.3V
A44	A/D(15)	B44	C/BE#(1)
A45	+3.3V	B45	A/D(14)
A46	A/D(13)	B46	GROUND
A47	A/D(11)	B47	A/D(12)
A48	GROUND	B48	A/D(10)
A49	A/D(9)	B49	GROUND
A50	<Key>	B50	<Key>
A51	<Key>	B51	<Key>
A52	C/BE#(0)	B52	A/D(8)
A53	+3.3V	B53	A/D(7)
A54	A/D(6)	B54	+3.3V
A55	A/D(4)	B55	A/D(5)
A56	GROUND	B56	A/D(3)
A57	A/D(2)	B57	GROUND
A58	A/D(0)	B58	A/D(1)
A59	+5 VOLTS	B59	+5 VOLTS
A60	REQ64#	B60	ACK64#
A61	+5 VOLTS	B61	+5 VOLTS
A62	+5 VOLTS	B62	+5 VOLTS

Notes:

* The two card type bits, B9 and B11, are connected on the riser.

** Int A-D, A6, A7, B7, and B8 are connected together on the riser.

D.5.15 ISA Connectors J29, J30, J31, J32

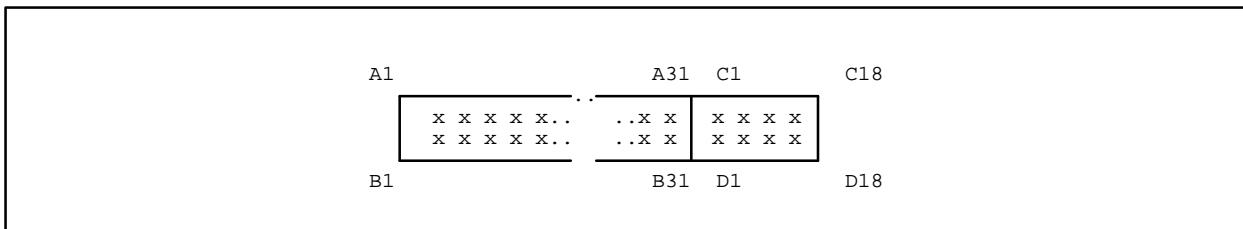


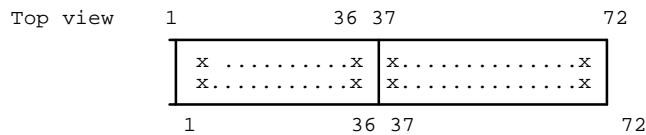
Figure D-17.ISA Connector

Table D-16. ISA Connector Pin Assignments

Pin	Function	Pin	Function
A1	IO CHCK#	B1	GROUND
A2	SD(7)	B2	RESET_DRV
A3	SD(6)	B3	+5 VOLTS
A4	SD(5)	B4	IRQ9
A5	SD(4)	B5	-5 VOLTS
A6	SD(3)	B6	DRQ2
A7	SD(2)	B7	-12 VOLTS
A8	SD(1)	B8	ZERO WS#
A9	SD(0)	B9	+12 VOLTS
A10	IO CHRDY	B10	GROUND
A11	AEN	B11	SMEMW#
A12	SA(19)	B12	SMEMR#
A13	SA(18)	B13	IOW#
A14	SA(17)	B14	IOR#
A15	SA(16)	B15	DACK3#
A16	SA(15)	B16	DRQ3
A17	SA(14)	B17	DACK1#
A18	SA(13)	B18	DRQ1
A19	SA(12)	B19	REFRESH
A20	SA(11)	B20	CLK
A21	SA(10)	B21	IRQ7
A22	SA(9)	B22	IRQ6
A23	SA(8)	B23	IRQ5
A24	SA(7)	B24	IRQ4
A25	SA(6)	B25	IRQ3
A26	SA(5)	B26	DACK2#
A27	SA(4)	B27	T/C
A28	SA(3)	B28	BALE
A29	SA(2)	B29	+5 VOLTS
A30	SA(1)	B30	Oscillator
A31	SA(0)	B31	GROUND
C1	SBHE#	D1	MEMCS16#
C2	LA(23)	D2	IO CS16#
C3	LA(22)	D3	IRQ10
C4	LA(21)	D4	IRQ11
C5	LA(20)	D5	IRQ12

Table D-16. ISA Connector Pin Assignments (Continued)

Pin	Function	Pin	Function
C6	LA(19)	D6	IRQ15
C7	LA(18)	D7	IRQ14
C8	LA(17)	D8	DACK0#
C9	MEMR#	D9	DRQ0
C10	MEMW#	D10	DACK5#
C11	SD(8)	D11	DRQ5
C12	SD(9)	D12	DACK6#
C13	SD(10)	D13	DRQ6
C14	SD(11)	D14	DACK7#
C15	SD(12)	D15	DRQ7
C16	SD(13)	D16	+5 VOLTS
C17	SD(14)	D17	MASTER#
C18	SD(15)	D18	GROUND

D.5.16 DRAM SIMM Connectors J21, J22, J23, and J24**Figure D-18.DRAM SIMM Connector****Table D-17. DRAM SIMM Connector Pin Assignments**

SIMM Pin	Motherboard Pin	Function
1	1	GROUND
2	2	DQ0
3	3	DQ18
4	4	DQ1
5	5	DQ19
6	6	DQ2
7	7	DQ20
8	8	DQ3
9	9	DQ21
10	10	+5 V
11	11	CASP
12	12	A0
13	13	A1
14	14	A2
15	15	A3
16	16	A4
17	17	A5
18	18	A6
19	19	A10

Table D-17. DRAM SIMM Connector Pin Assignments (Continued)

SIMM Pin	Motherboard Pin	Function
20	20	DQ4
21	21	DQ22
22	22	DQ5
23	23	DQ23
24	24	DQ6
25	25	DQ24
26	26	DQ7
27	27	DQ25
28	28	A7
29	29	BS0/A11
30	30	+5V
31	31	A8
32	32	A9
33	33	RAS3#
34	34	RAS2#
35	35	DQ26
36	36	DQ8
37	37	DQ17
38	38	DQ35
39	39	GROUND
40	40	CAS0#
41	41	CAS2#
42	42	CAS3#
43	43	CAS1#
44	44	RAS0#
45	45	RAS1#
46	46	BS1
47	47	WE#
48	48	RESD1
49	49	DQ9
50	50	DQ27
51	51	DQ10
52	52	DQ28
53	53	DQ11
54	54	DQ29
55	55	DQ12
56	56	DQ30
57	57	DQ13
58	58	DQ31
59	59	+5 V
60	60	DQ32
61	61	DQ14
62	62	DQ33
63	63	DQ15
64	64	DQ34

Table D-17. DRAM SIMM Connector Pin Assignments (Continued)

SIMM Pin	Motherboard Pin	Function
65	65	DQ16
66	66	BS2
67	67	PD1
68	68	PD2
69	69	PD3
70	70	PD4
71	71	BS3
72	72	GROUND

D.5.17 Power Switch Connector J8

Top view

**Figure D-19.1x2 Power Switch Connector****Table D-18. Power Switch Connector Pin Assignments**

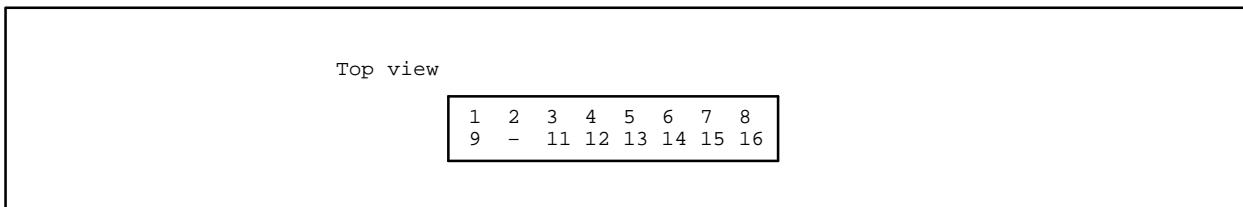
Pin No.	Signal Name
1	SWITCH_P1
2	SWITCH_GD

D.5.18 Power Up Configuration Connector J7

Top view

**Figure D-20.1x2 Power Up Configuration Connector****Table D-19. Power Up Configuration Connector Pin Assignments**

Pin No.	Signal Name
1	PWR_CFG1
2	PWR_CFG2

D.5.19 RISCWatch Connector J2**Figure D-21.2x8 RISCWatch Connector****Table D-20. RISCWatch Connector Pin Assignments**

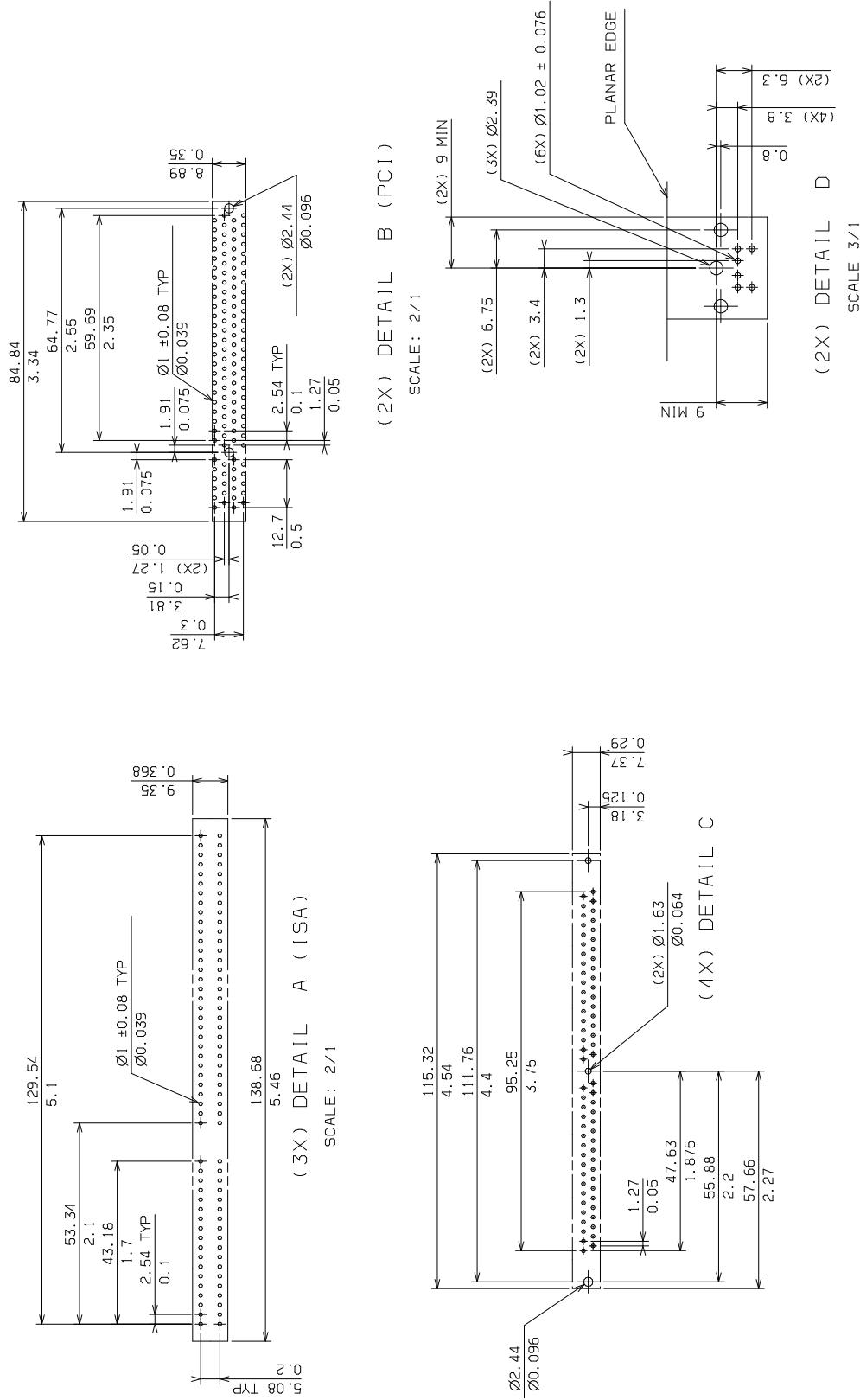
Pin No.	Signal Name
1	CHECK_STOP#
2	HDWR_RESET#
3	RESET_INTERRUPT#
4	CNTL/SCAN_DATA
5	SHIFT_CLK
6	+RUN/-BREAKPOINT
7	SCAN_IN
8	SCAN_OUT
9	GROUND
10	<Key>
11	GROUND
12	RESERVED
13	RESERVED
14	+5 VOLTS
15	OCS_OVERRIDE
16	RESERVED

D.5.20 Battery Connector BT2

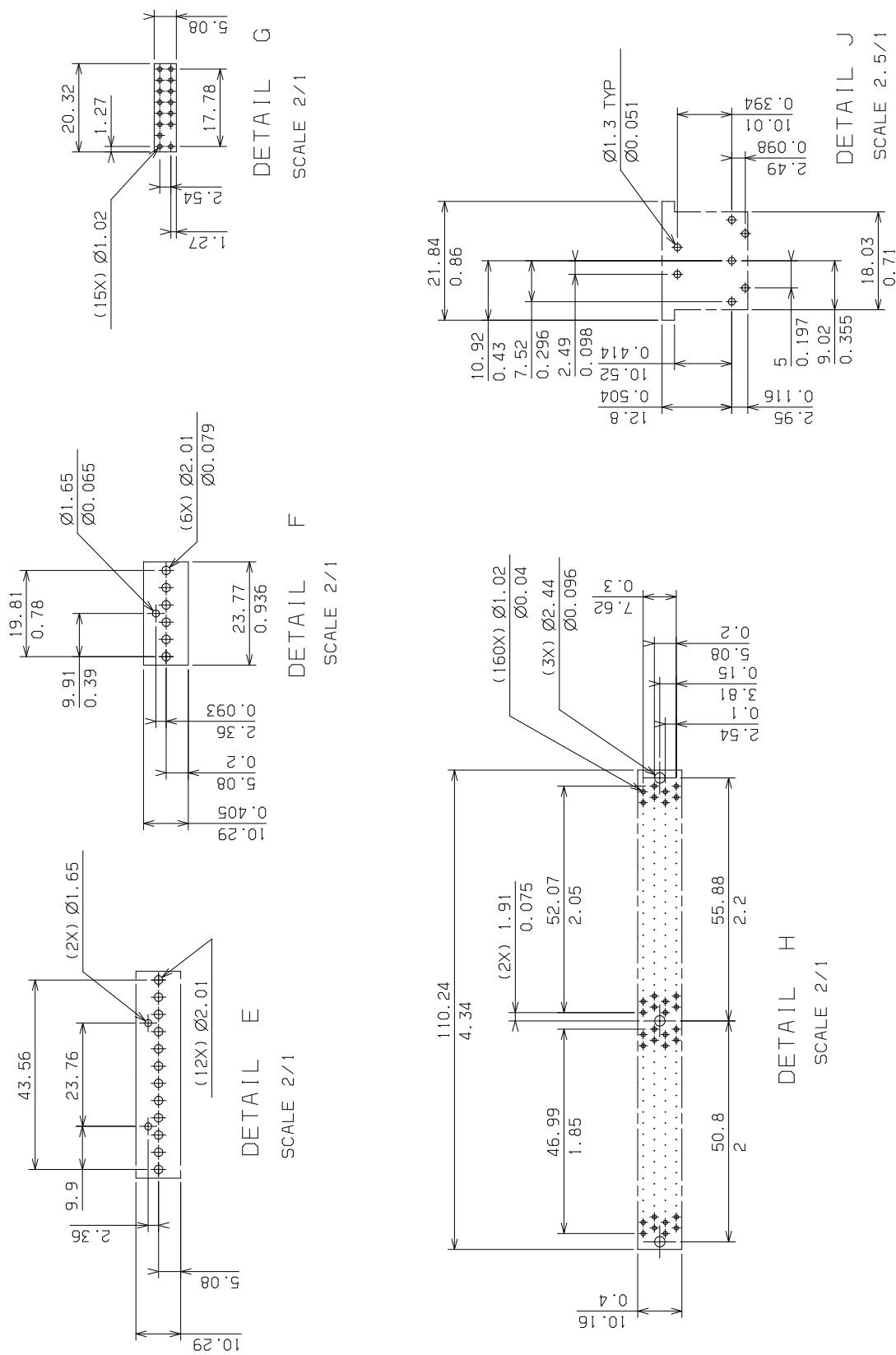
The battery type is CR2032 3 Volt.

Insert the battery with + side up.

D.5.21 MCM Board Connector Footprint #1



D.5.22 MCM Board Connector Footprint #2



D.6 Enclosure

The example planar is mechanically very similar to the Harley board. Enclosures for Harley should work for the example planar, but this has not been verified. The information in this section should be considered as tentative. The experienced designer will verify the appropriateness of an enclosure before specifying it.

The following vendors supply enclosures:

Olson Metal Products Company Inc.

Attn: Michelle Seay

1903 N. Austin Street

Seguin, Texas 78155

1-800-951-9517 or (210) 379-7000

AT Desktop model number: CC300249

Medium Tower model number: CC400000

Altex Electronics

11342 IH-35 North

San Antonio, Texas

1-800-531-5367 or FAX (210) 637-3264

Mini Tower model number STC-05

Medium Tower model number STC-08

Full Tower model number STC-16

Mega-Tech Marketing Inc.

3900-D Drossett Dr.

Austin, Texas

Mini Tower model number A6601

Full Tower model number A5561

Note: IBM makes no recommendations regarding vendors of any components.

In planning the layout of an enclosure, the height of the items in Table D-21 should be considered.

Table D-21. Height Considerations

Item	Approximate Height
3.3V regulator heat sink	26.0 mm
SIMMs (seated)	28.4 mm
L2 SRAM SIMM (seated)	28.4 mm
MCM and mounting	

Appendix E Bill of Materials

E.1 Planar Bill of Materials

Table E-1. Planar Bill of Materials

No	Part Name	Part Number	Reference Designator	Qty	Manufacturer Name	Manufacturer Part Number	JEDEC TYPE	Part Description
1	10MQ040_SOIC-BASE	87F4917	CR1,CR7	2	SANKEN	SFBP-54YL	D64	40V 1A SCHOTTKY REC
2	29F040ROM_MECH-MECH-BASE	82G6496	X3	1	AMD	AM29F040-120JC	PART	512KX8 FLASH ROM
3	2N7002_SMD-BASE	31F2311	Q1	1	MOTOROLA	2N7002	SOT23	3 PIN FET TRANSISTOR
4	7406_SMD-BASE	17F7776	U3,U28	2	TI	7406	SO14	HEX INVERTER OC
5	74F08_SMD-BASE	61X9236	U1,U29	2	NATL	F08	SO14	QUAD 2 INPUT & GATE
6	74F11_SMD-BASE	17F7745	U13	1	NATL	F11	SO14	TRI 3 INPUT AND GATE
7	74F125_SMD-BASE	68X2888	U16	1	NATL	74F125SCX	SO14	74F125 BUFFERS 3 STATE
8	74F244_SMD-BASE	6480438	U10,U11,U17-U20, U40	7	PHILIPS	74F244	SO20W	OCTAL BUFF/DRIVER
9	74F245_SMD-BASE	55X8091	U12	1	PHILIPS	74F245	SO20W	OCTAL TRANSCEIVER
10	74F74_SMD-BASE	61X9238	U21	1	NATL	F74	SO14	DUAL "D" FF W/CLEAR & PRESET
11	74HCT14_SMD-BASE	37F9034	U2	1	PHILIPS	74HCT14	SO14	74HCT14 SCHMIDT TRIGGER INV
12	8042H_SMD-BASE	1054195	U27	1	INTEL	8042H	PLCC44	KEYBOARD CONTROLLER
13	8G4756_SMD-BASE	08G4756	CR6	1	MOTOROLA	MMBD914 LT1	SOT23	70V 200MA SWC DIODE
14	BATTERY_MECH-BASE	15F8409	B1	1	SONY	CR2032	PART	BATTERY 3.3V
15	BERG1X2_DIP-BASE	6181127	J8-J11,J16,J35	6	BERG	69190-502	BERG1X2	1X2 100MIL HEADER VERTICAL
16	BERG1X4_DIP-BASE	6359315	J13	1	BERG	69190-504	BERG1X4	1X4 100MIL HEADER VERTICAL
17	BERG1X5_DIP-BASE	88G4908	J12	1	AMP	104345-3	BERG1X5	BERGSTICK 1X5
18	CAPACITOR-0.001UF,20%	98F1292	C129-C148	20	KYOCERA	08055C102 KAT2A	SMC0805	CAPACITOR

Table E-1. Planar Bill of Materials (Continued)

No	Part Name	Part Number	Reference Designator	Qty	Manufacturer Name	Manufacturer Part Number	JEDEC TYPE	Part Description
19	CAPACITOR-0.01UF,20%	41F0313	C3-C5,C28-C31, C33-C37,C39,C40, C46-C50,C52,C53, C57,C58,C61,C80- C82,C84,C86,C93, C95,C117-C120, C124,C127,C151, C157-C164,C167- C169,C172-C174, C203-C206,C208- C210,C212-C215, C217-C220,C222- C225,C255-C257, C259,C269-C274	81	KYOCERA	0805X103 M2B05	SMC0805	CAPACITOR
20	CAPACITOR-0.1UF,20%	41F0316	C9,C43,C91,C92, C94	5	KYOCERA	0805Y104 Z1B05	SMC0805	CAPACITOR
21	CAPACITOR-10UF,20%	71F7911	C8	1	NEC	TESVEC 1C106M12R	71F7911	CAPACITOR
22	CAPACITOR-2200PF,20%	42G3220	C32,C83,C85,C123, C125	5	MURATA	GRM40X7R 222J050AD	SMC0805	CAPACITOR
23	CAPACITOR-33UF/20%/16V	57G9281	C202,C207,C211, C216,C221	5	KEMET	T491D336 M016AS	CAP33UF3SMT	CAPACITOR
24	CAPACITOR-68PF,20%	62G4724	C87-C90,	4	KYOCERA	08055A680 KAT2	SMC0805	CAPACITOR
25	CAPACITOR-DO_NOT_POP, 20%	-NONE-	C1,C2,C6,C7,C10- C21,C70,C71,C98, C99,C149,C150, C152-C154,C156, C226	27	NONE	NONE	SMC0805	CAPACITOR
26	CHANDRA_SMD-BASE	SKT_100	M1	1	ALTERA	100 QFP SOCKET	100QFP	GTP W/128 MACROCELLS
27	CONN BATT_KELPIE_CONN-BASE	33F8027	M5	1	KELPIE	K17-BH-001	CONN_33F8027	BATTERY HOLDER
28	CONNBERG2X36_DIP-BASE	N_MOLEX	J36-J39	4	MOLEX	15-44-4536	BERG2X36	2X36 MOLEX PCB CONN
29	CONN DIN5-DO_NOT_POP	-NONE-	J14A	1	NONE	NONE	CONN_5DIN	AT KEYBOARD CONNECTOR
30	CONN MDIN6-CONN MDIN6	15F6890	J14,J15	2	AMP	749180-1	CONN_15F6890	6 POS CIRCULAR MINI DIN
31	CONNPOWER2_DIP-BASE	N_BURND	J5	1	BURNDY	GTC6R-1	CONN_1X6	1X6 0 156 CL FRET LCK HDR
32	CONNPOWER_DIP-BASE	55X8085	J4	1	MOLEX	15-48-0212	CONN_55X8085	1X12 0 156 CL FRET LCK HDR
33	CONN_BATTERY-SONY_NO_POP	19G2441	M3	1	SONY		CONN_19G2441	BATTERY HOLDER
34	CONN_BERG1X3_DIP-BASE	1501831	J1,J7,J17,J18,J40-J42	7	BERG	69190-503	BERG1X3	1X3 100MIL HEADER VERTICAL
35	CONN_ISA_DIP-BASE	6137473	J29-J32	4	AMP	645169-4	ISA2X49	2X49 ISA CONNECTOR
36	CONN_PCI_DIP-BASE	72G0316	J19,J25-J27	4	AMP	646255-1	PCI2X605V	2X60 32 BIT PCI CONNECTOR
37	CONN_SIMM72_DIP-BASE	64F5806	J21-J24	4	AMP	822032-4	SIMM72	72 PIN SIMM CONNECTOR
38	CRYSTAL_2PIN-32.768KHZ	03G9527	Y3	1	EPSON	MC405-32.768KHZ-6PF	XTALSMT1	CRYSTAL
39	CRYSTL-DO_NOT_POP	-NONE-	Y2,Y4	2	NONE	NONE	XTALSMT2	CRYSTAL
40	DO-NOT-POP_PLCC20_PLCC20-BASE	-NONE-	S3	1	NONE	NONE	PLCC20	DO-NOT-POP
41	DS1385_SMD-BASE	70G6764	U5	1	DALLAS	DS1385S	SO28W	RTC+ 4KX8SRAM
42	ELCAP-100UF, 10%/16V	75G8253	C110	1	SPRAGUE	293D107X 96R3D2T	SMC2816	CAPACITOR

Table E-1. Planar Bill of Materials (Continued)

No	Part Name	Part Number	Reference Designator	Qty	Manufacturer Name	Manufacturer Part Number	JEDEC TYPE	Part Description
43	ELCAP-33UF, 20%/16V	57G9281	C27,C38,C112–C116,C121,C122, C165,C166,C170, C171,C175–C183, C187–C198,C261–C263	37	KEMET	T491D336 MO16AS	CAP33UF3SMT	CAPACITOR
44	EPM5130FP_SMD–BASE	05H0966	X2	1	ALTERA	EPM5130–QC–100–1	PART	GTP W/128 MACROCELLS
45	ESP_CONN_CONN–BASE	42F6867	J2	1	BERG	79282–516	CONN_42F6867	2X8 PIN HEADER CONN
46	F_BEAD_BEAD–BASE	26F4865	FB1,FB4–FB6	4	TDK	HF50ACB–321611T	SMC1206	FERRITE BEAD
47	IRFZ44_SMD–BASE	03G9500	Q6	1	IR	IRFZ44	TO220X	TRANSISTOR 50V 35A
48	LT1431REG_DIP–BASE	31F2428	U4	1	LINEARTECH	LT1431CS8	SO8	VOLT REGULATOR
49	MTGHOLE–DO_NOT_POP	–NONE–	MH1,MH3–MH5, MH7,MH8	6			157TOOL	
50	OSCLR-14.3181MHZ	87F5263	Y1	1	EPSON	SG–615P–14.31818 MC	OSCSMT4	OSCLR
51	OSCLR-24.0MHZ	87F5265	Y5	1	EPSON	SG–615P–24.0000MC	OSCSMT4	OSCLR
52	PAL_20–SKT_19G5840	19G5840	S1	1	AMP	3-822269-1	PLCC20	SOCKET
53	PAL_MECH_MECH–BASE	05H0958	X5	1	AMD	PALCE16V8 H–5JC–5	PART	PALCE16V8H–5JC–5
54	PART–JUMPER1X2	92F1420	X1,X4	2	AMP	530153–2	PART	JUMPER 1X2
55	PIN1089_DIP–BASE	60G8498	J3	1	IBM	CUSTOM	MCM33X33SKT	1089 PIN CGB SOCKET
56	PLCC32SKT_SMD–BASE	10G7624	M2	1	AMP	821977–1	PLCC32SKT	32 PIN PLCC SOCKET
57	POLYSWITCH_SMD–BASE	34G3113	F1	1	RAYCHEM	SMD100	POLYSWITCH	TAPE RESISTOR
58	POWER1X3_DIP–BASE	65G3724	J6	1	MOLEX	705–43–0037	POWER1X3	AUX POWER CONNECTOR
59	REGHSINK_DIP–BASE	8185352	M4	1	AAVID	533722 B22552	REGHSINK	HEATSINK VOLTREG
60	RESISTOR-0,5%	98F1665	R14,R15,R17–R19, R69,R70,R72,R95, R129,R130,R135, R470,R471,R473	15	PANASONIC	ERJ8GVJ 0R	SMC0805	RESISTOR
61	RESISTOR-1.5K,5%	98F1741	R436	1	ROHM	MCR10EZH LJW152	SMC0805	RESISTOR
62	RESISTOR-10,5%	58F1831	R40,R82,R85,R87, R89,R96,R97	7	PANASONIC	ERJ6GVYJ 100S	SMC0805	RESISTOR
63	RESISTOR-100,5%	41F0328	R38,R39,R101, R107,R108,R153, R154,R176,R190, R192,R194,R238, R240,R365,R366, R485	16	PANASONIC	ERJ6VJ 101S	SMC0805	RESISTOR
64	RESISTOR-10K,5%	41F0337	R1,R5–R8,R10,R13, R43,R74–R76,R78, R80,R81,R90–R93, R99,R102,R111– R113,R119,R138– R145,R147,R148, R155–R166,R174, R237,R253–R260, R287–R351,R364, R443,R488–R495, R497,R510	133	PANASONIC	ERJ6GVJ 103S	SMC0805	RESISTOR
65	RESISTOR-12.7K,1%	40G7066	R512	1	PANASONIC	ERJ–6VNF 1272S	SMC0805	RESISTOR

Table E-1. Planar Bill of Materials (Continued)

No	Part Name	Part Number	Reference Designator	Qty	Manufacturer Name	Manufacturer Part Number	JEDEC TYPE	Part Description
66	RESISTOR-150,5%	98F1671	R123	1	ROHM	MCR10EZH MJW151	SMC0805	RESISTOR
67	RESISTOR-1K,5%	41F0333	R12,R21,R28,R29, R31-R36,R41,R44, R47,R55,R77,R100, R230,R239,R280, R505,R513	21	PANASONIC	ERJ6GVJ 102S	SMC0805	RESISTOR
68	RESISTOR-2.7K,5%	09G9748	R261-R272	12	PANASONIC	ERJ-6GVYJ 272S	SMC0805	RESISTOR
69	RESISTOR-200,1%	40G6920	R4	1	PANASONIC	ERJ-6VNF 2000S	SMC0805	RESISTOR
70	RESISTOR-200,5%	98F1424	R73	1	PANASONIC	ERJ-6GEYJ 201V	SMC0805	RESISTOR
71	RESISTOR-22,5%	98F1736	R56,R88,R177, R178	4	PANASONIC	ERJ6GVYJ 220S	SMC0805	RESISTOR
72	RESISTOR-220K,5%	61F2952	R122	1	PANASONIC	ERJ6GVJ 224S	SMC0805	RESISTOR
73	RESISTOR-249,1%	40G7233	R437	1	PANASONIC	ERJ6VNF 2490S	SMC0805	RESISTOR
74	RESISTOR-3.9K,5%	42G3067	R120	1	PANASONIC	ERJ6GVYJ 392S	SMC0805	RESISTOR
75	RESISTOR-300,5%	98F1674	R103,R106,R225, R236,R241,R251, R281-R286	12	ROHM	MCR10EZH MJW301	SMC0805	RESISTOR
76	RESISTOR-33,5%	41F0327	R22-R27,R30,R65, R67,R68,R71,R83, R84,R86,R104, R105,R226-R229, R243,R250	22	PANASONIC	ERJ6GVYJ 330S	SMC0805	RESISTOR
77	RESISTOR-4.7K,5%	41F0336	R231-R235,R273- R279	12	PANASONIC	ERJ-6GVJ 472S	SMC0805	RESISTOR
78	RESISTOR-4.99K,1%	40G7034	R511	1	PANASONIC	ERJ6VNF 4991S	SMC0805	RESISTOR
79	RESISTOR-470K,5%	03G9709	R124	1	ROHM	MCR10EZH LJW474	SMC0805	RESISTOR
80	RESISTOR-5.6K,5%	98F1737	R352-R358	7	PANASONIC	ERJ6GVJ 562S	SMC0805	RESISTOR
81	RESISTOR-510,5%	41F0331	R118,R245	2	PANASONIC	ERJ6GVYJ 511S	SMC0805	RESISTOR
82	RESISTOR-620,5%	61F2960	R20	1	PANASONIC	ERJ6GVJ 621S	SMC0805	RESISTOR
83	RESISTOR-75,5%	61F2961	R79,R359-R362	5	PANASONIC	ERJ6GVJ 750S	SMC0805	RESISTOR
84	RESISTOR-80.6,1%	40G6887	R9	1	PANASONIC	ERJ-6VNF 80R6S	SMC0805	RESISTOR
85	RESISTOR-DO_NOT_POP,5%	-NONE-	R2,R3,R11,R16, R37,R42,R45,R46, R54,R57,R58,R62, R66,R94,R98,R121, R146,R149-R152, R189,R191,R193, R440,R441,R451, R472	28	NONE	NONE	SMC0805	RESISTOR
86	RLS4148_SOIC-BASE	87F4920	CR2	1	ROHM	RLS4148	MELF	40V 1A SCHOTTKY REC
87	S82378ZB_SMD-BASE	82G6542	U7	1	INTEL	S82378ZB	QFP208_5MM	PCI TO ISA BRIDGE CONT
88	XTAL-16.50MHZ	89G3833	Y2A,Y4A	2	ECLIPTEK	ECX-2900-16.50MHZ	CLP_XTAL	16.5 MHZ CRYSTAL
90	XTAL-DO_NOT_POP	-NONE-	Y3A	1	NONE	NONE	CLP_XTAL	NONE
Total				591				

Appendix F Planar Schematics

This section contains the schematics of the Odyssey MCM planar. For schematics of the MCM, see Section 13.

The schematic pages are numbered separately from the rest of the MCM technical specification.

Page 50 of the planar schematics shows a PAL that is used only for development and factory testing of parts of the MCM. This device is not required in user systems, and can be completely removed from the circuit. Note that in normal operation, the PAL is not populated, and the nets that are connected to the PAL outputs are driven by other devices.

F.1 Component Placement

ODYSSEY PLANTAR

FOR MCM

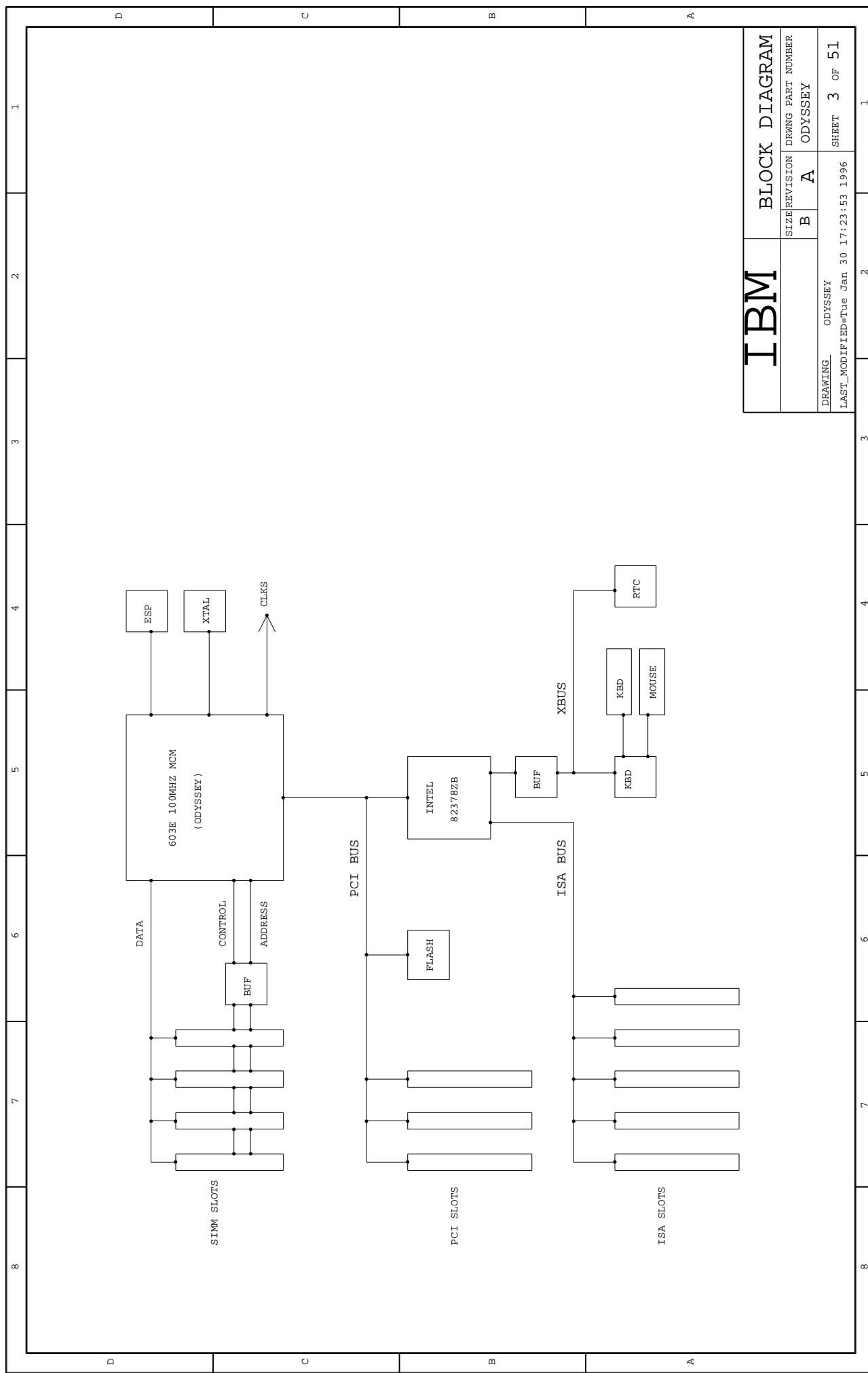
LAST_MODIFIED=Tue Apr 2 17:28:21 1996

EC LEVEL: XXXXXXXXX
RELEASE DATE: XXXXXXXXX

This Document contains preliminary information and is subject to change without notice. IBM assumes no responsibility of liability for any use of the information contained herein. Nothing in this document shall operate as an express or implied license or indemnity under the intellectual property rights of IBM or third parties. No warranty or guarantee is given concerning any information contained in this document. To ensure that you have the latest version of this document, please contact your marketing representative or IBM MICROELECTRONICS tactical marketing 1-800-IBM-0181.

IBM		COVER SHEET	
SIZE	REVISION	DRAWING PART NUMBER	
B	A	ODYSSEY	SHEET 1 OF 51
8	7	6	5
4	3	2	1

CONTENTS						1	
2		3		4		5	
D	1 FRONT PAGE	1					
	2 CONTENTS	2					
	3 BLOCK DIAGRAM	3					
	4 CRITICAL NETS & CHANGE HISTORY	4					
	5 SCHEMATICS	5-51					
C	- MCM INTERFACE	5					
	- MCM INTERFACE	6					
	- MCM INTERFACE	7					
	- EMULATOR CARD INTERFACE	8					
	- EMULATOR CARD INTERFACE	9					
	- SIGNAL SERIES RESISTORS	10					
	- BLANK	11					
	- BLANK	12					
	- RISWATCH	13					
	- BLANK	14					
	- ISA BRIDGE	15					
	- ISA BRIDGE	16					
	- ISA BRIDGE	17					
	- X BUS BUFFERS	18					
	- I/O EPID	19					
	- BLANK	20					
	- POWER SUPPLY MISC	21					
	- DRAM PD REDRIVE BUFFERS	22					
	- PLANAR ID	23					
	- L2 & PCI PRESENT DETECT	24					
	- FLASH ROM	25					
	- REAL TIME CLOCK	26					
	- KEYBOARD & MOUSE CONTROL	27					
	- FRONT PANEL MISC	28					
B	- CLOCK OSCILLATOR & CRYSTALS	29					
	- VOLTAGE REGULATOR	30					
	- POWER & FRONT PANEL HEADERS	31					
	- KEYBOARD & MOUSE CONNECTORS	32					
	- BLANK	33					
	- PCI PULLUPS	34					
	- ISA PULLUPS AND PULLDOWNS	35					
	- EMC CAPS	36					
	- MEMORY SIMMS	37					
	- MEMORY SIMMs	38					
	- PCI CONNECTOR	39					
	- PCI CONNECTOR	40					
	- PCI CONNECTOR	41					
	- PCI CONNECTOR	42					
	- PCI CONNECTOR	43					
	- ISA CONNECTOR	44					
	- ISA CONNECTOR	45					
	- ISA CONNECTOR	46					
	- BLANK	47					
	- BLANK	48					
	- CONFIG RESISTORS	49					
	- TAG_MATCH CONTROL	50					
	- SPARE PAL SITE	51					
A	TBM						1
	CONTENTS						
	DRAWING	ODYSSEY	SIZE	REVISION	DRWNG PART NUMBER		
		B	A		ODYSSEY		
	LAST_MODIFIED=Thu Feb 8 14:59:44 1996				SHEET 2 OF 51		



D									
8	7	6	5	4	3	2	1		
C									
B									
A									

CHANGE HISTORY

V0.1A 07/28/95 DRAFT

CRITICAL NET NAME	NET LENGTH	INSTRUCTIONS
OTHER CONSIDERATIONS		
-sysclk_pclk_en	10.0	yes
-l2_pclk<4..0>	10.0	yes
-639_pclk	10.0	yes
-624_pclk	10.0	yes
-pc1_0lk	15.0	yes
-pc1_clk4..1>	8.5	yes
-isa_clk	7.3	yes
-butf_ps_power_good	asap	yes
-manureset_x	asap	no
-power_nen	asap	no
-sysreset	asap	no
-pc1_reset	asap	no
-hreset_60x	asap	no
-reset_60x	asap	no
-esp_lrc	asap	no
-hdwrst	asap	no
-res0_EF16x	asap	no
-res0_EF16x	asap	no
-txd6014	asap	no
-ts60x	asap	no
-xats60x	asap	no
-txr10>	asap	no
-ta60x	asap	no
-okstop_out_60x	asap	no
-int60x	asap	no
-ope60x	asap	no
-ope60x	asap	no
-epd_tms60x	asap	no
-drt60x	asap	no
-ta60x	asap	no
-tea60x	asap	no
-okstop_out_60x	asap	no
-ram_asr	asap	no
-tag_match	asap	no
-tag_we_id	asap	no
-adq310a	asap	no
-pc1_iframe	asap	no
-pc1_frdy	asap	no
-pc1_irdy	asap	no
-pc1_stop	asap	no
-pc1_devel	asap	no
-pc1_derr	asap	no
-pc1_coe<3..0>	asap	no
-pc1_lock	asap	no
-pc1_bar	asap	no

NOTE : - All nets should be in the inner layer & daisy chained wherever

(1) All the nets should be daisy chained.

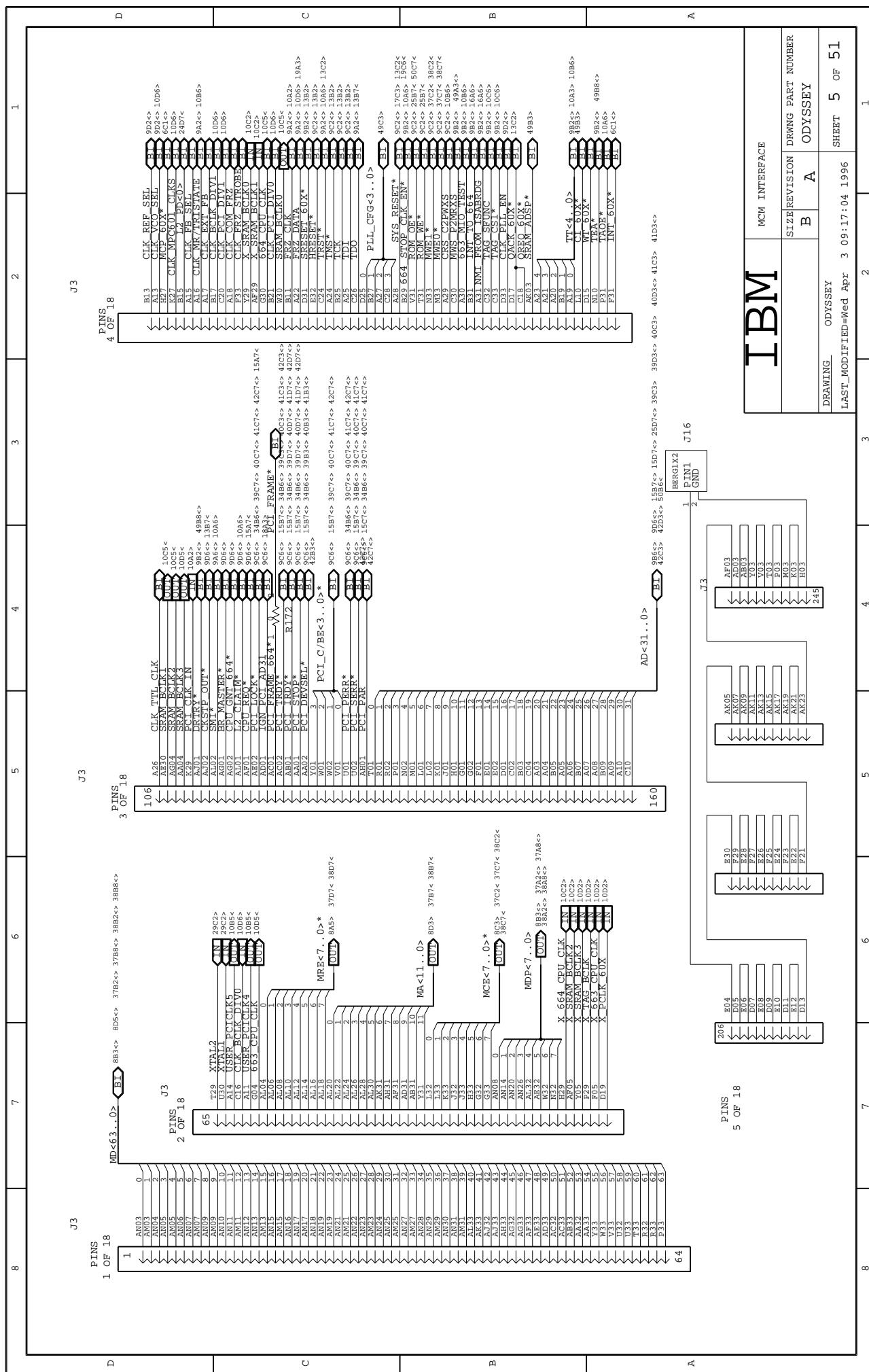
(2) All the nets should be daisy chained.
The wiring order is : U4 - U33 - U32 - U5 - U37 - U38 - J3.

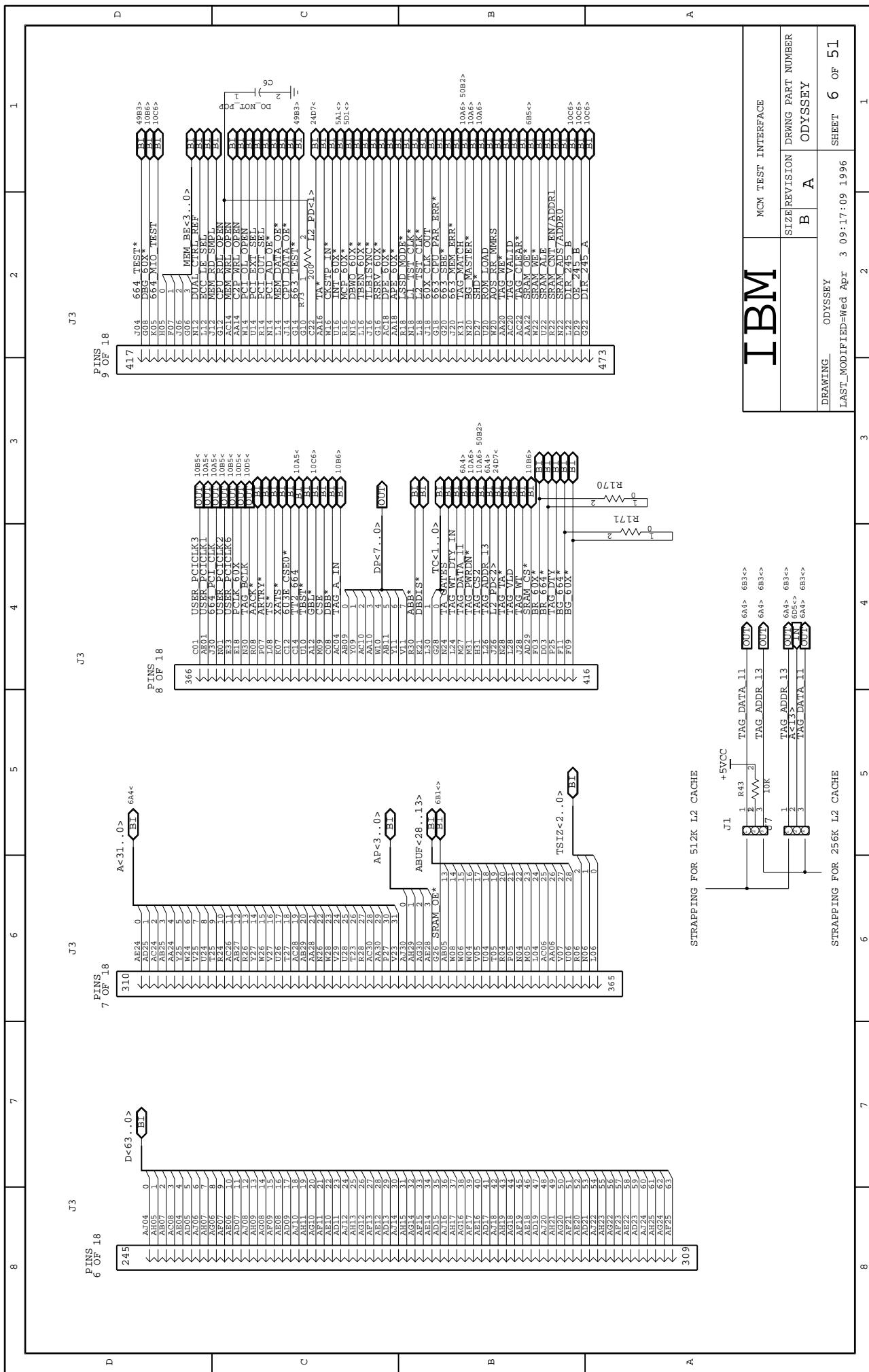
(3) All the nets should be daisy chained.
The wiring order is : M2 - M3 - U5 - U7 - J25 - J26 - J27 - J28 .

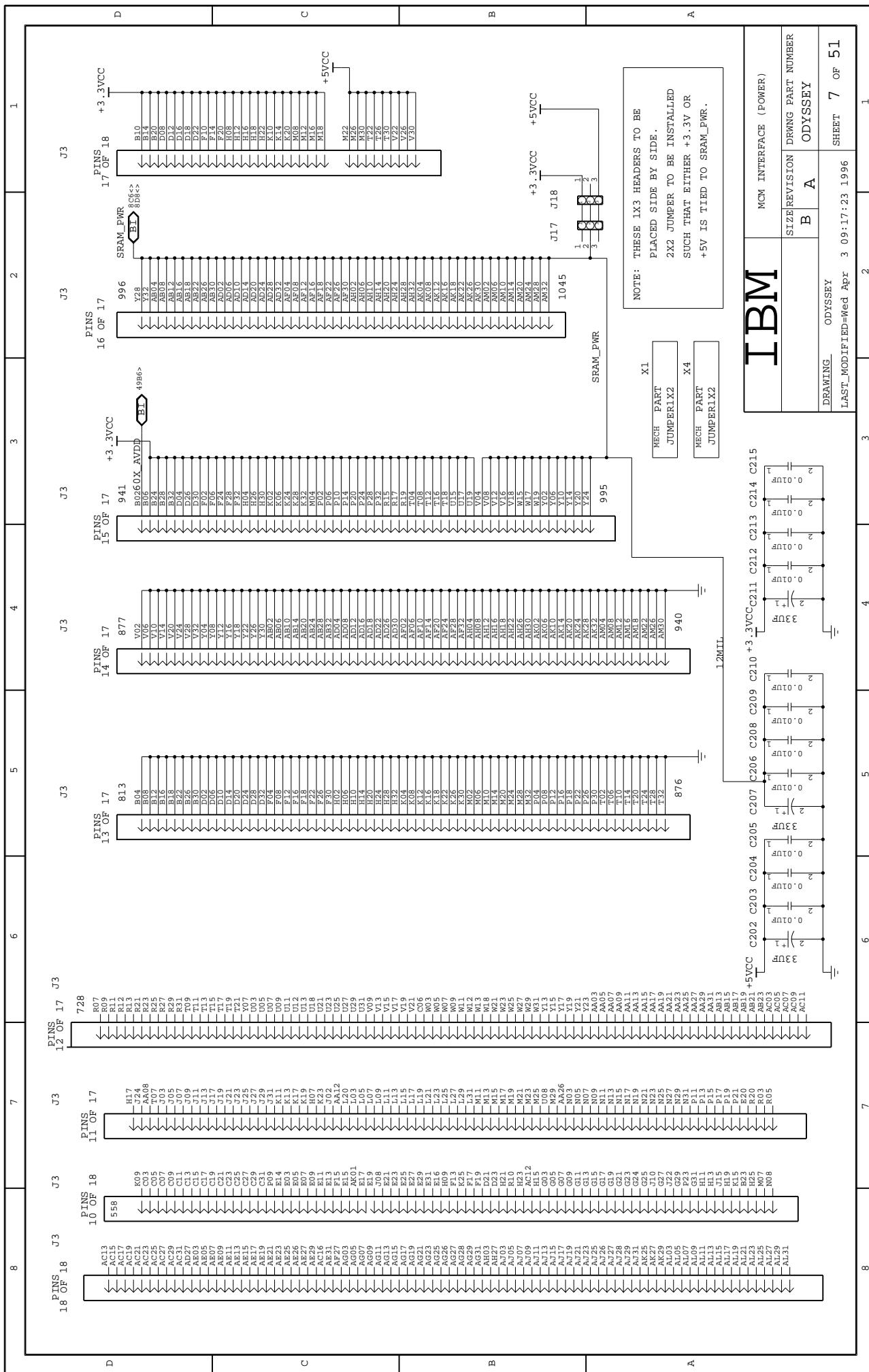
(4) Matched length group name. Max Net length.

(5) Max Net length.

IBM	CRITICAL NETS CHANGE HISTORY		
DRAWING	SIZE	REVISION	DRAWING PART NUMBER
ODYSSEY	B	A	ODYSSEY
LAST_MODIFIED=Mon Jan 29 11:20:08 1996	2	3	SHEET 4 OF 51
	2	3	1

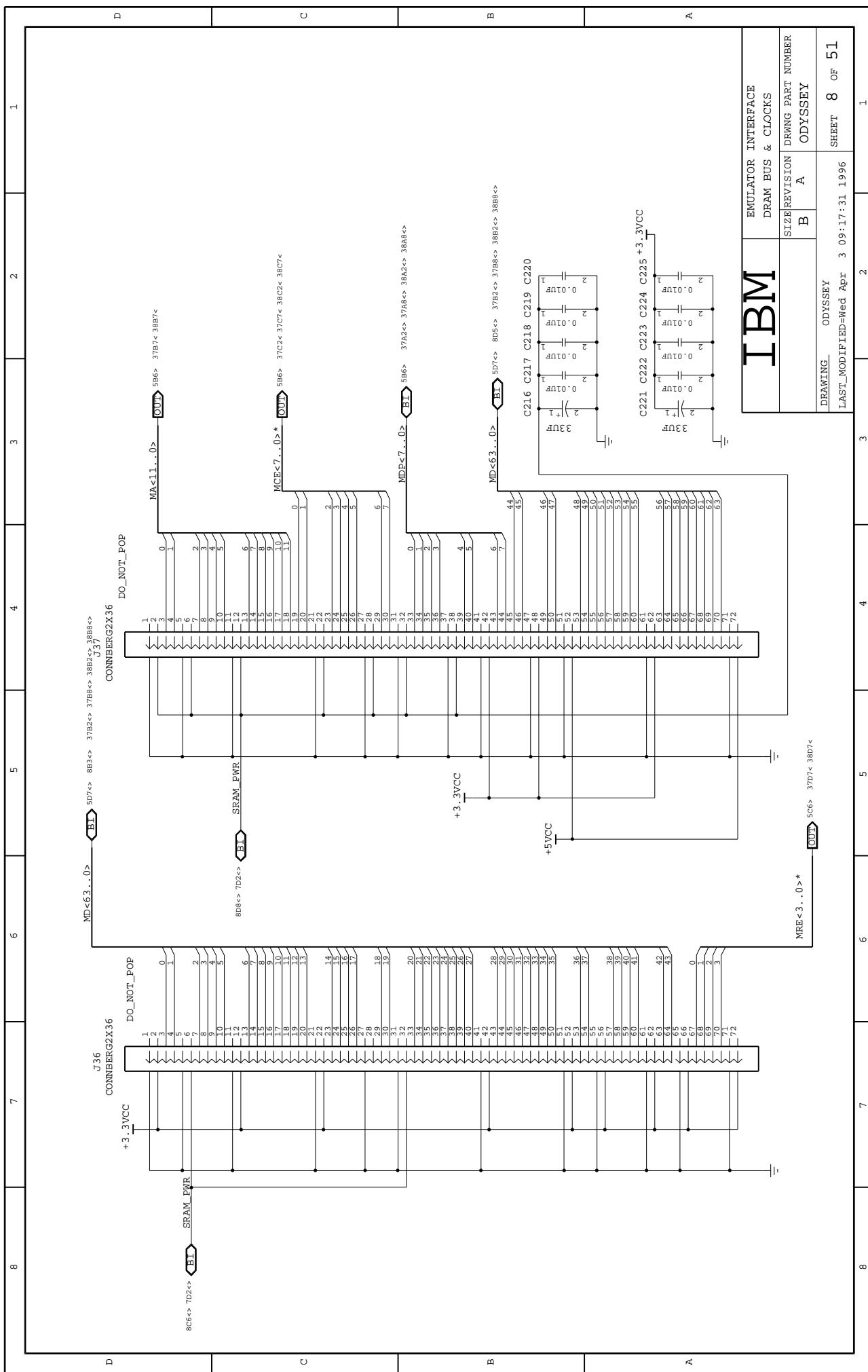


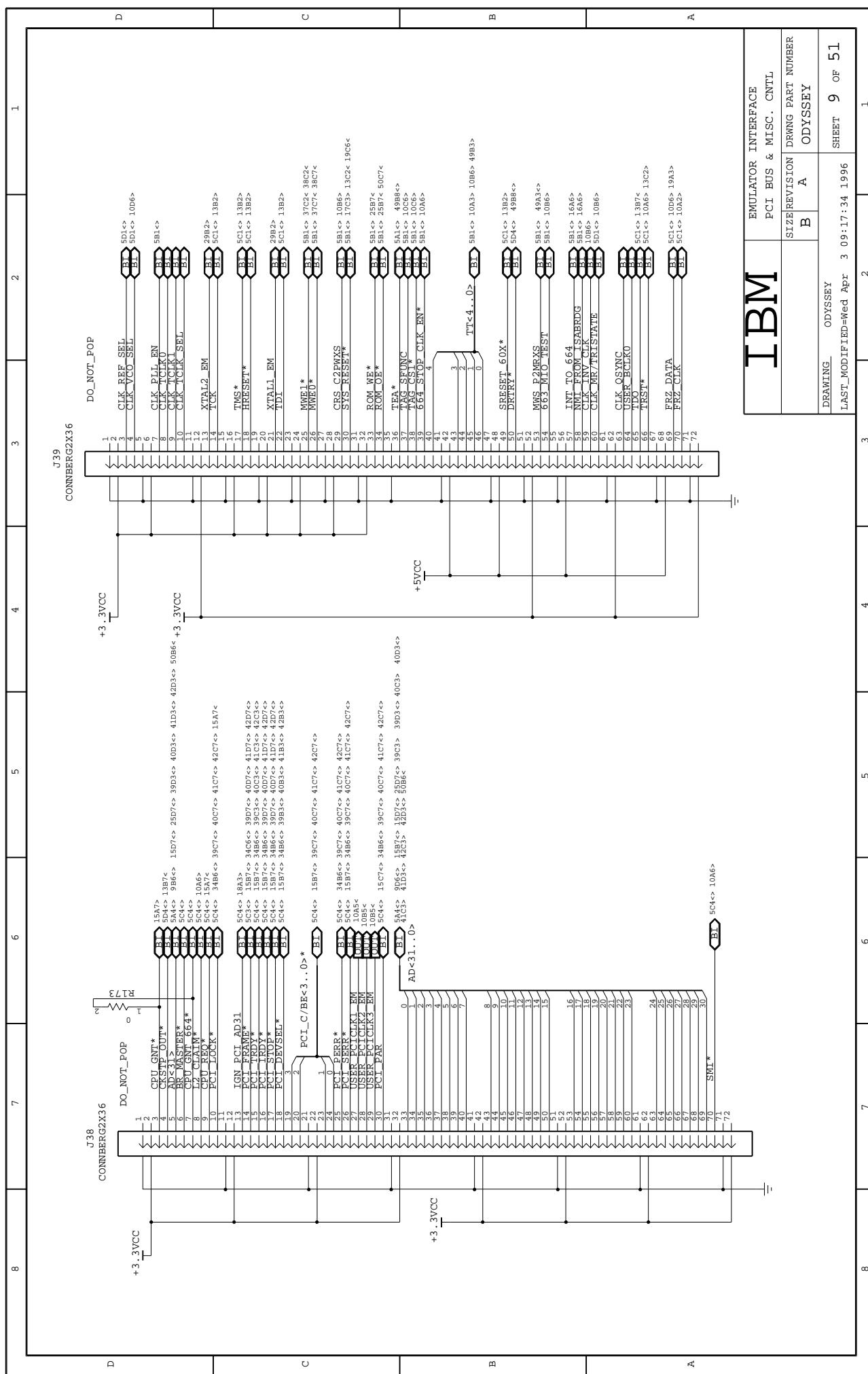


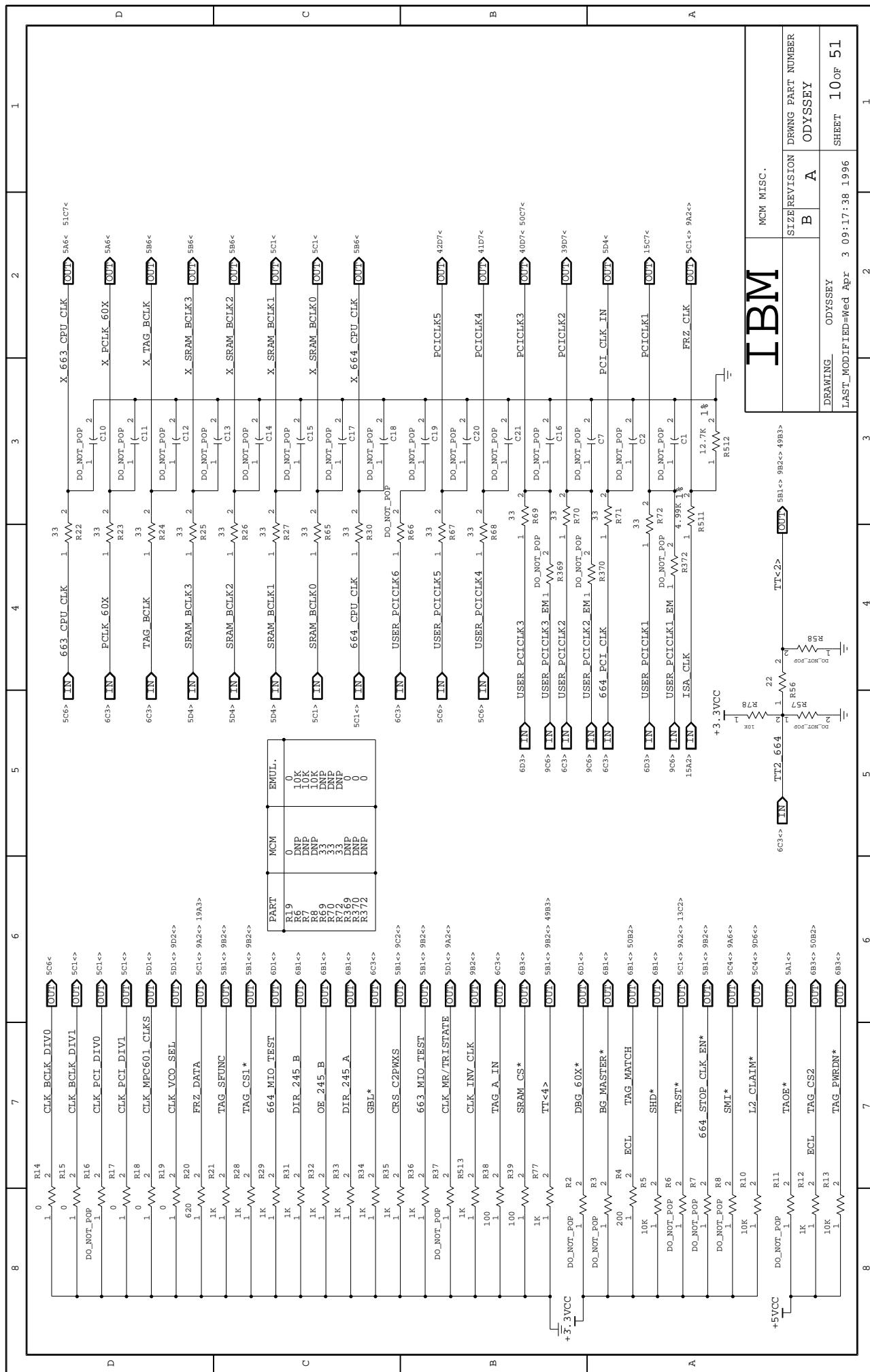


IBM MCM INTERFACE (POWER)		
DRAWING	DRAWING PART NUMBER	SHEET
LAST_MODIFIED=Wed Apr 3 09:17:23 1996	ODYSSEY	7 OF 51

B	A	SIZE
2	2	2





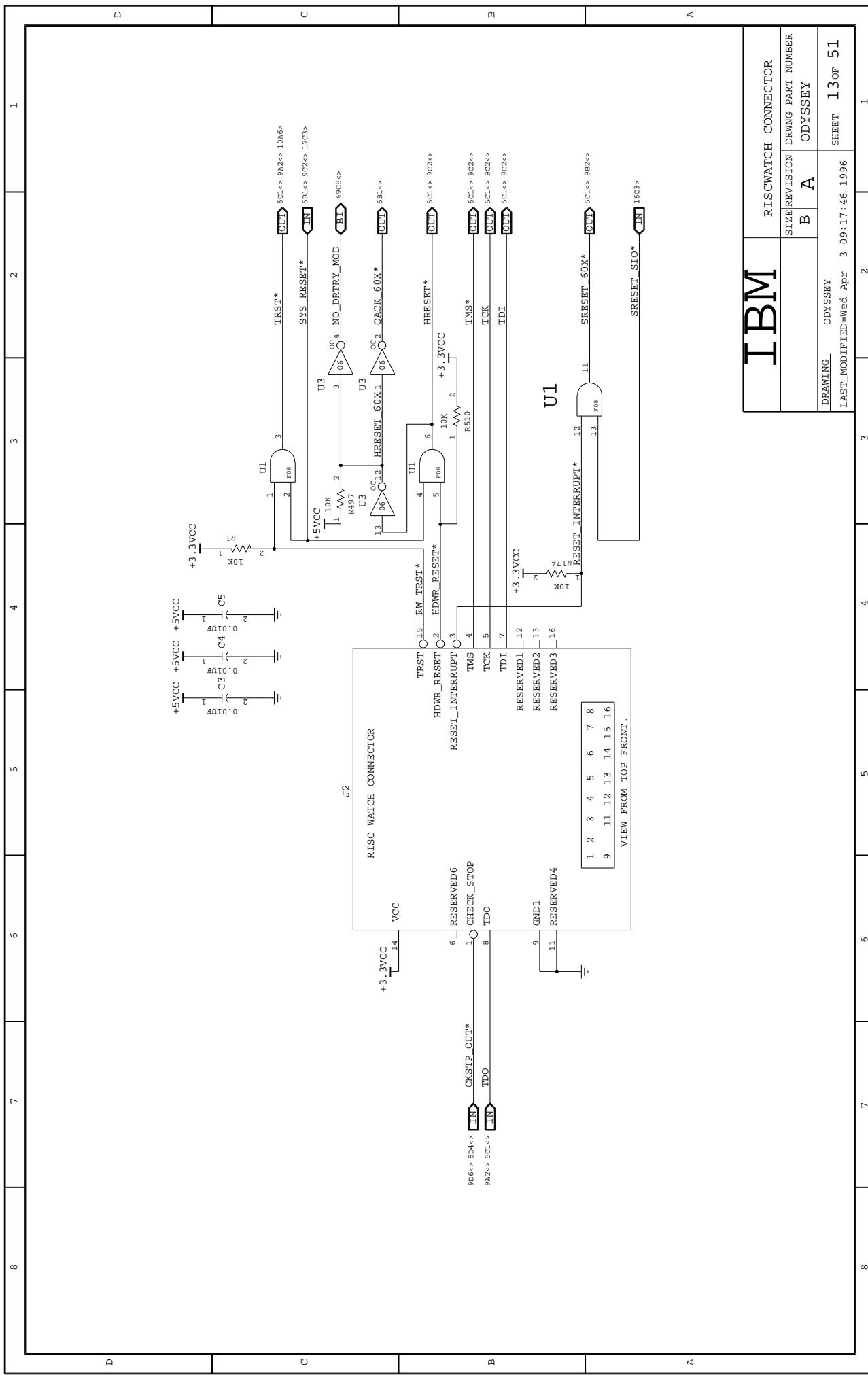


D	C	B	A
8 7 6 5 4 3 2 1			
BLANK			

IBM		BLANK	
DRAWING	LAST_MODIFIED=Mon Jan 29 16:02:20 1996	SIZE	REVISION
		B	A
		2	1
		3	
		4	
		5	
		6	
		7	
8			

D	C	B	A																																
<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												

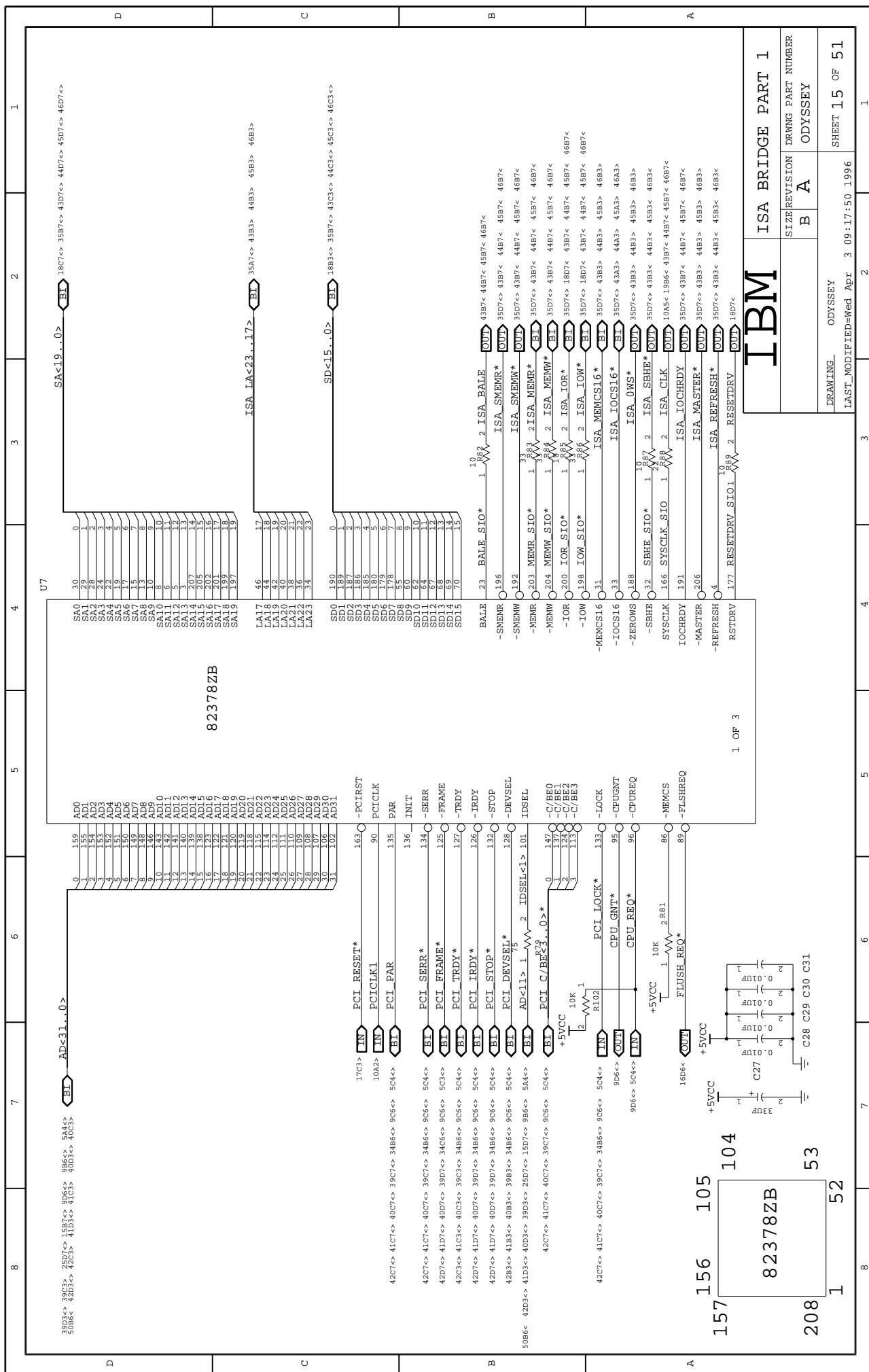
BLANK

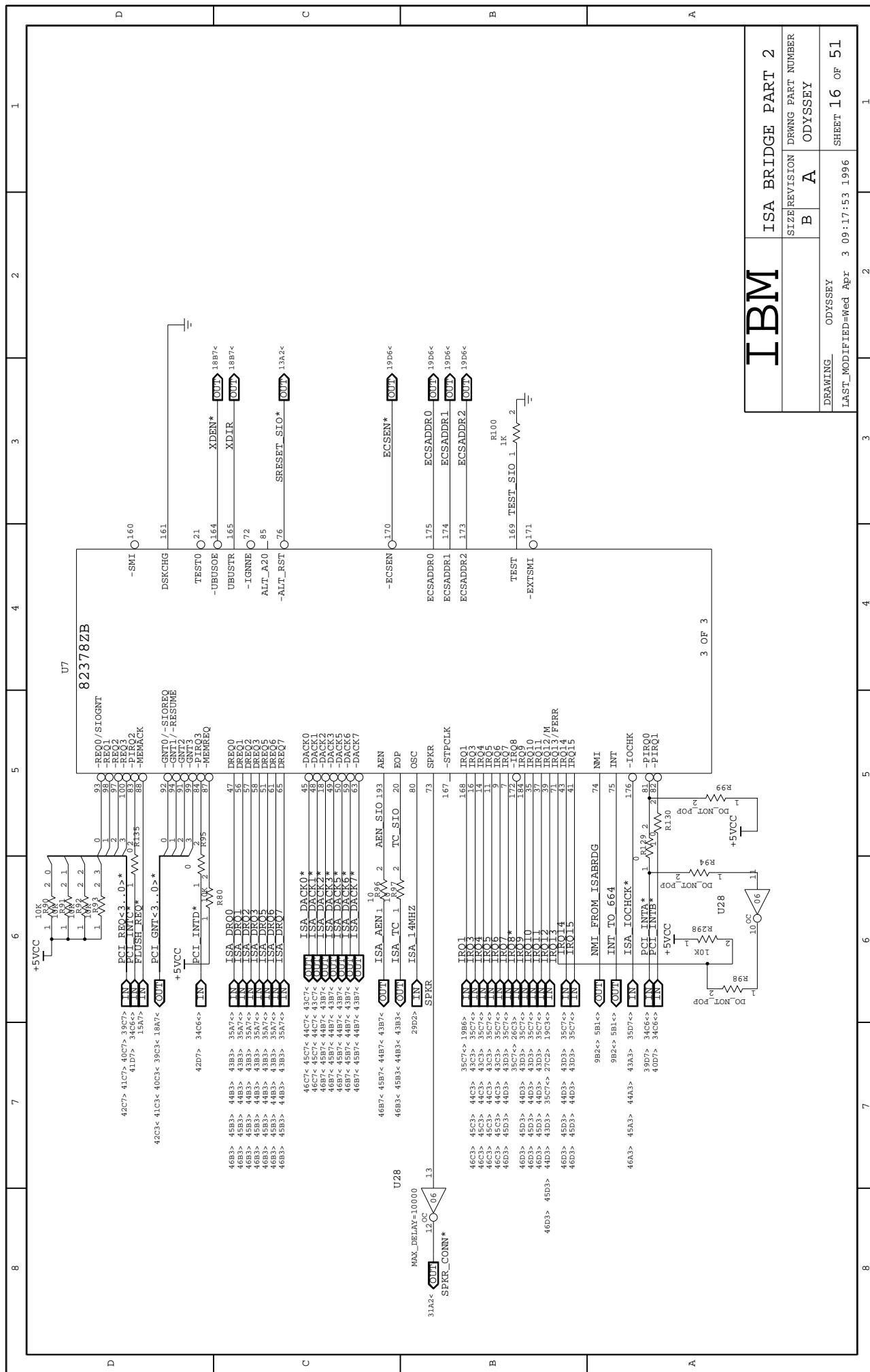


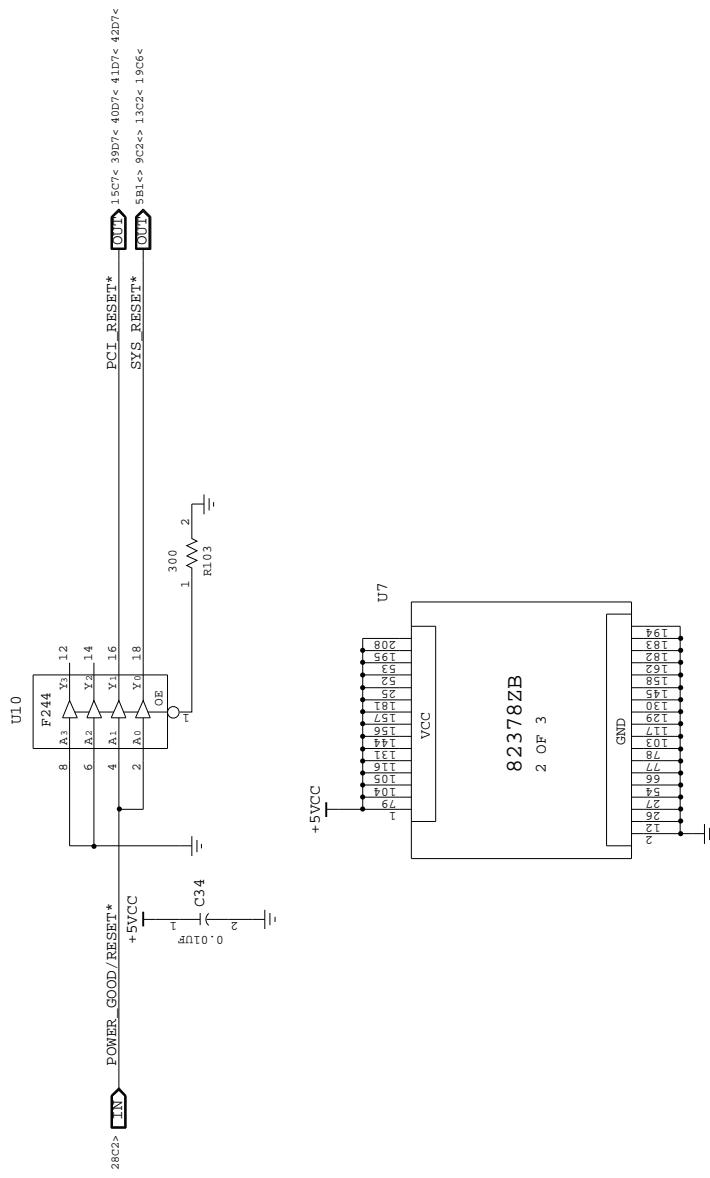
D	C	B	A																																
<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												

BLANK

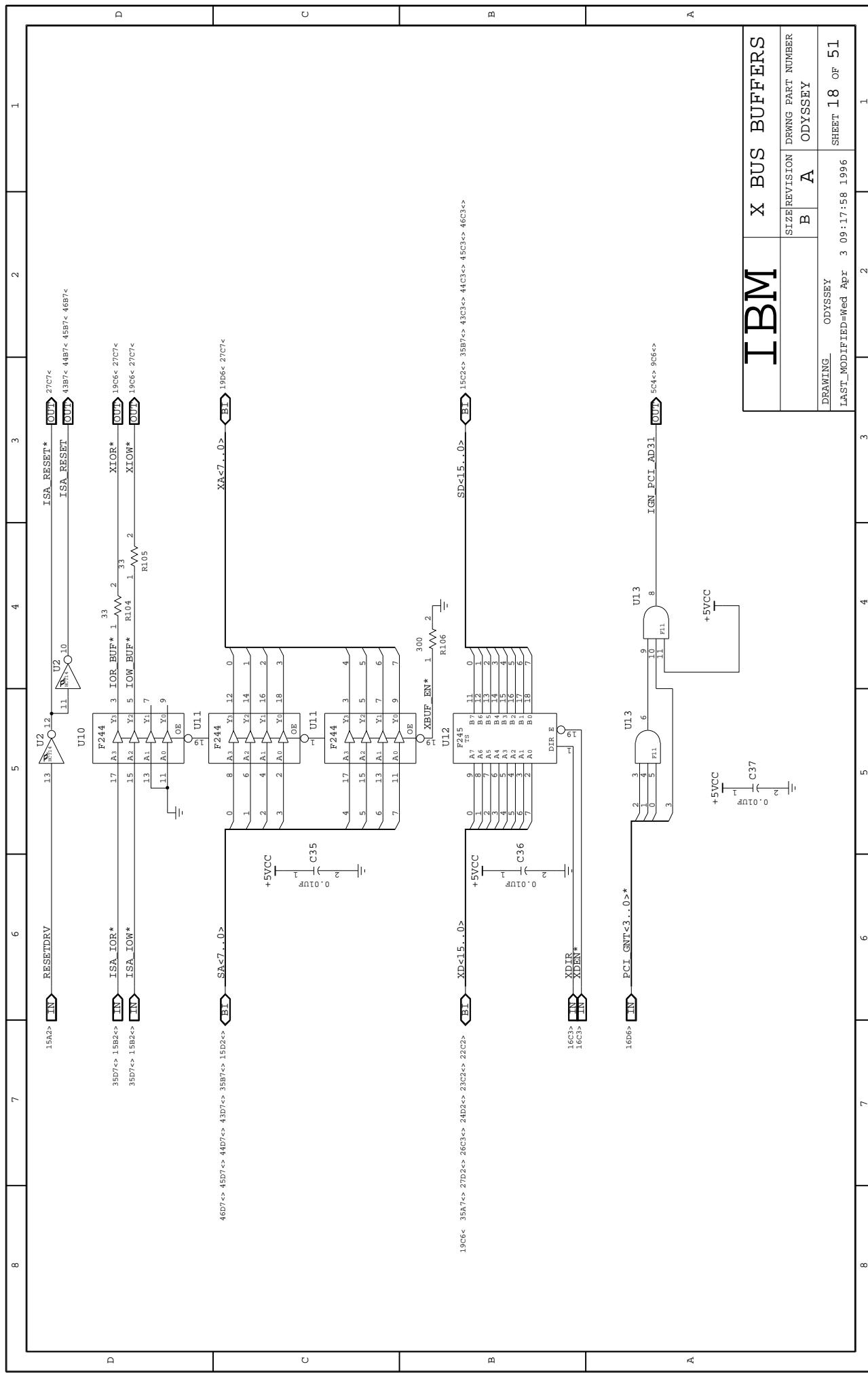
IBM		BLANK	
SIZE	REVISION	DRAWING PART NUMBER	
B	A	ODYSSEY	
DRAWING	ODYSSEY	LAST_MODIFIED=Mon Jan 29 16:02:54 1996	SHEET 14 OF 51

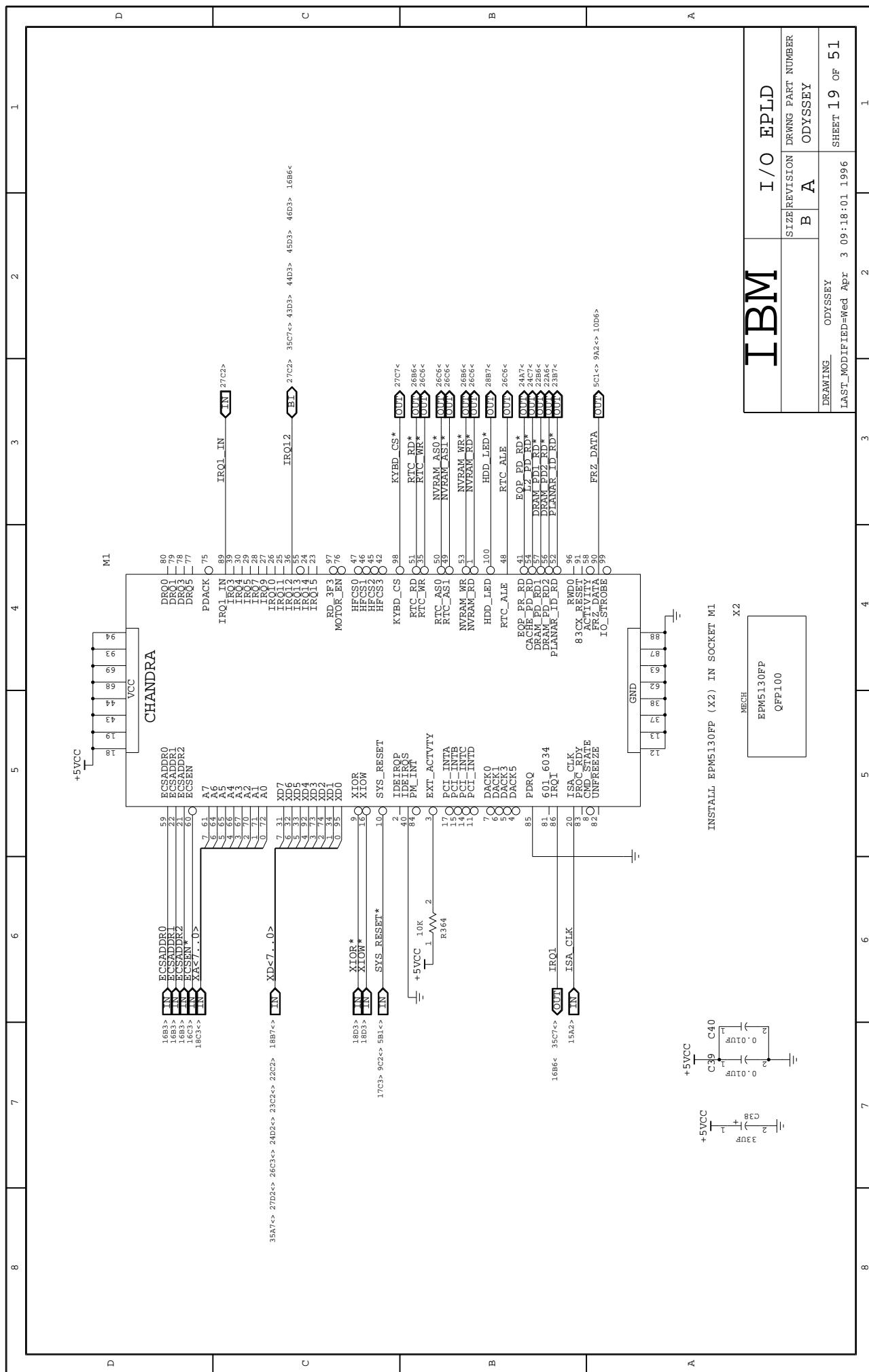






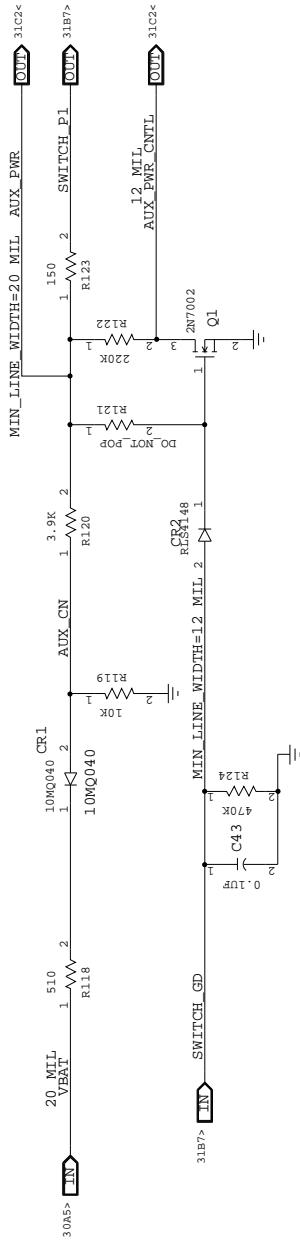
IBM		ISA BRIDGE PART 3	
DRAWING	ODYSSEY	SIZE	REVISION
LAST_MODIFIED-Wed Apr 3 09:17:56 1996		B	A
			DRWNG PART NUMBER ODYSSEY
			SHEET 17 OF 51





D	C	B	A																																
<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												

BLANK



IBM POWER SUPPLY MIS.C.			
	SIZE B	REVISION A	DRAWING PART NUMBER ODYSSEY
DRAWING	ODYSSEY	ODY	
LAST_MODIFIED	Med. Apr	3	09:18:03 1996
			SHEET 21 OF 51

JOURNAL OF POLYMER SCIENCE: PART A: POLYMERS

8

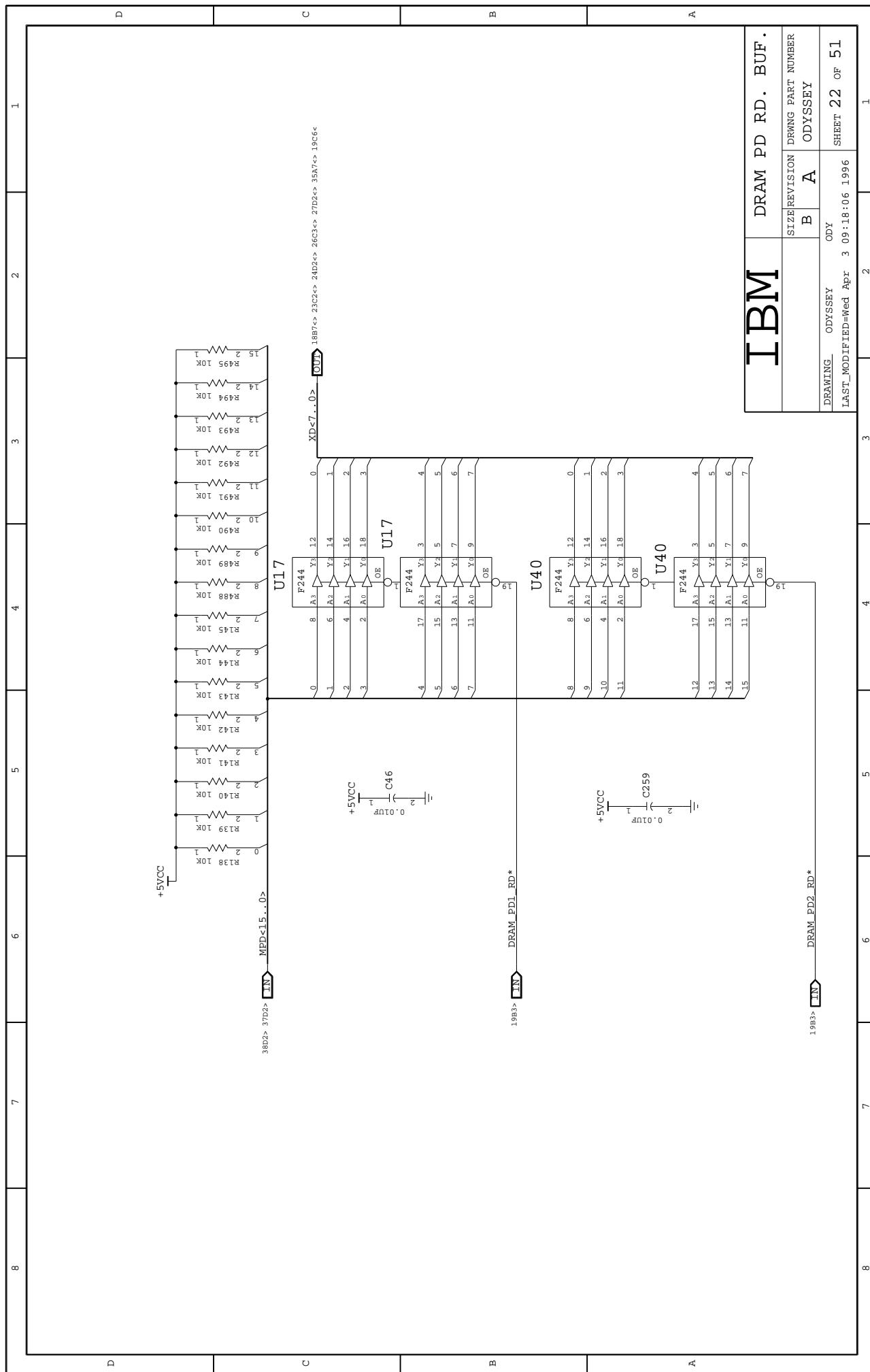
A

5

B

A

U



IBM		DRAM PD RD. BUF.	
SIZE	REVISION	DRAWING PART NUMBER	DRWG PART NUMBER
B	A	ODY	ODYSSEY

DRAWING : ODYSSEY
LAST MODIFIED=Wed Apr 3 09:18:06 1996
SHEET 22 OF 51

19B3> DRAM_FDL RD*

8

7

6

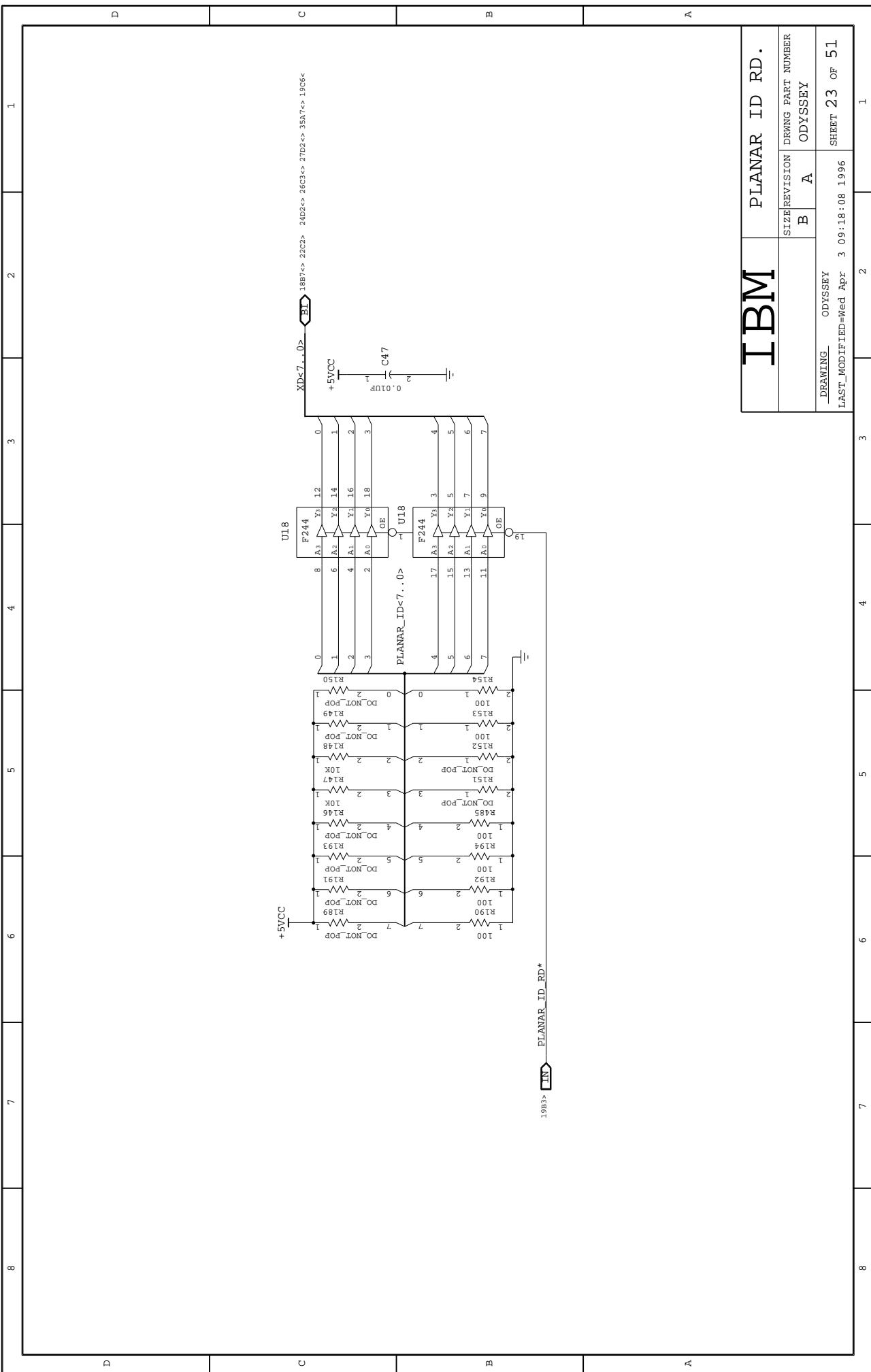
5

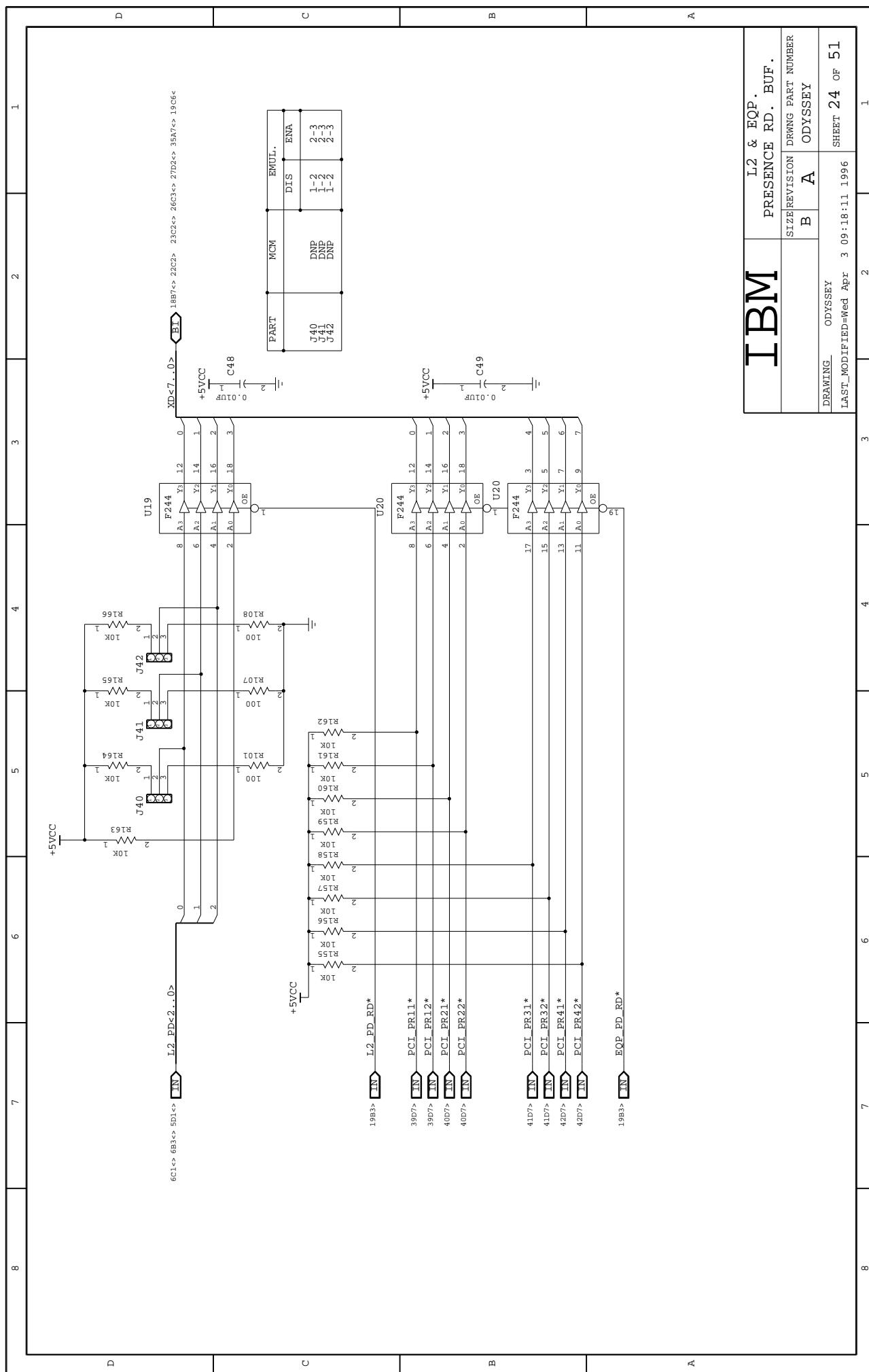
4

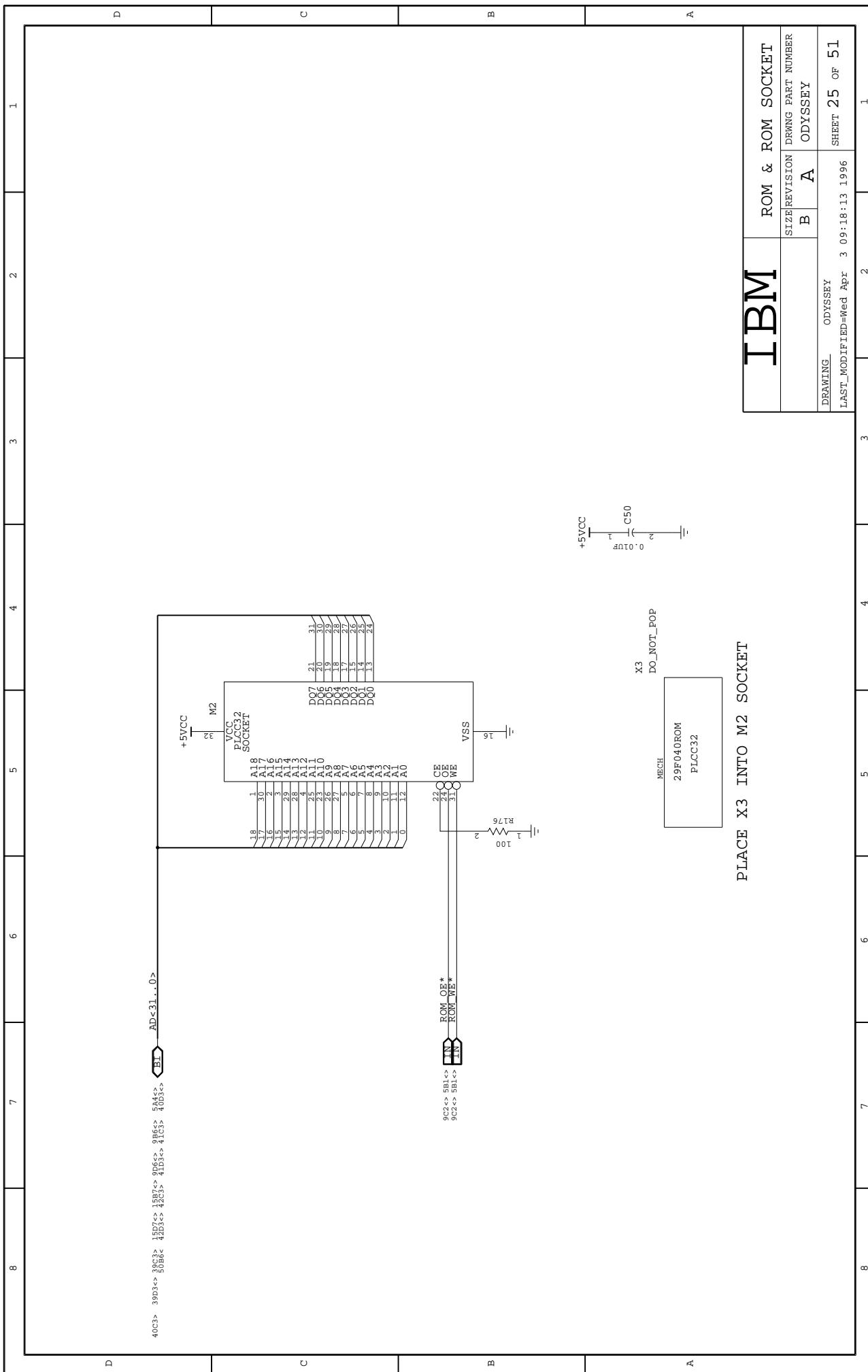
3

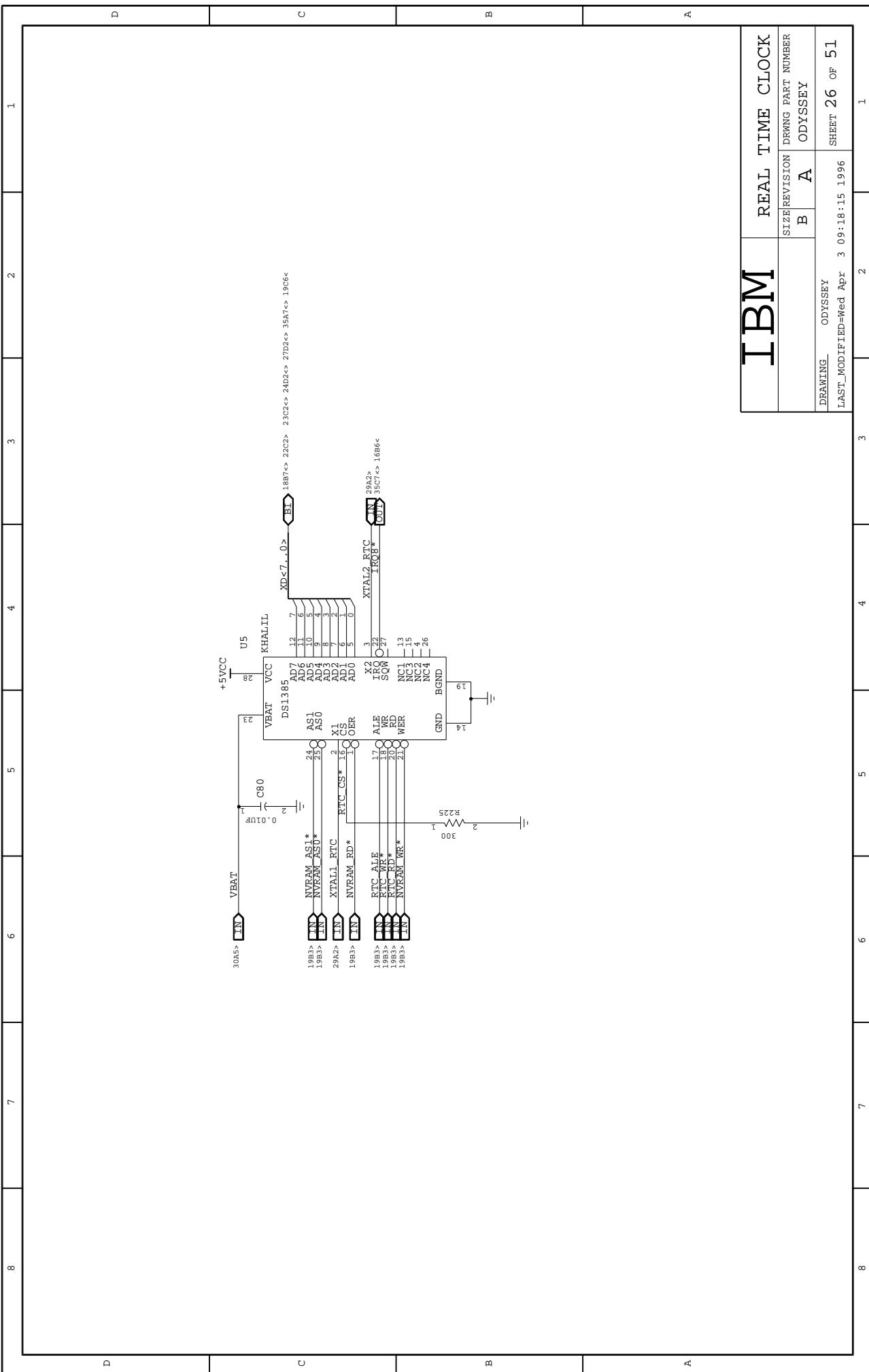
2

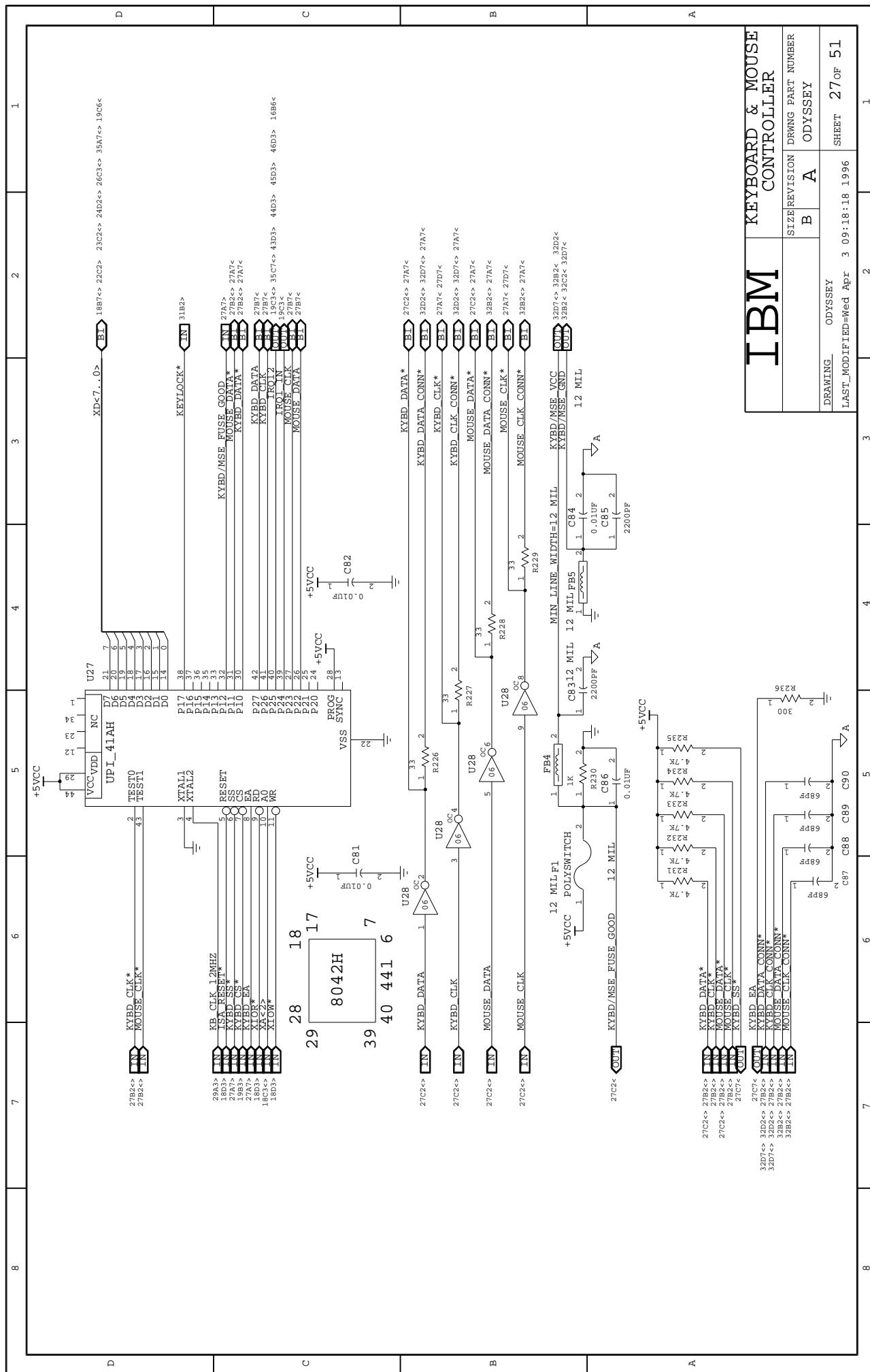
1

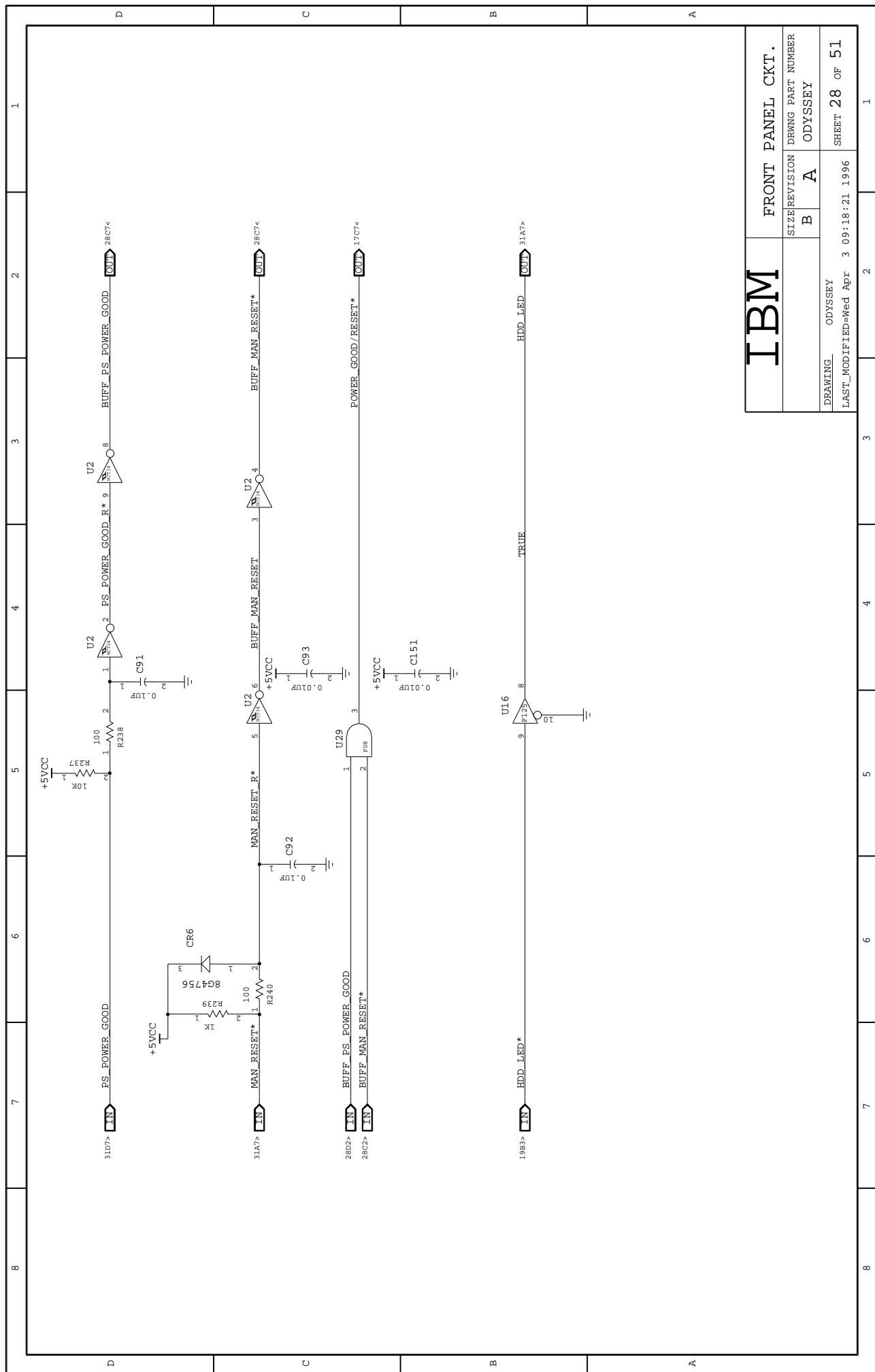


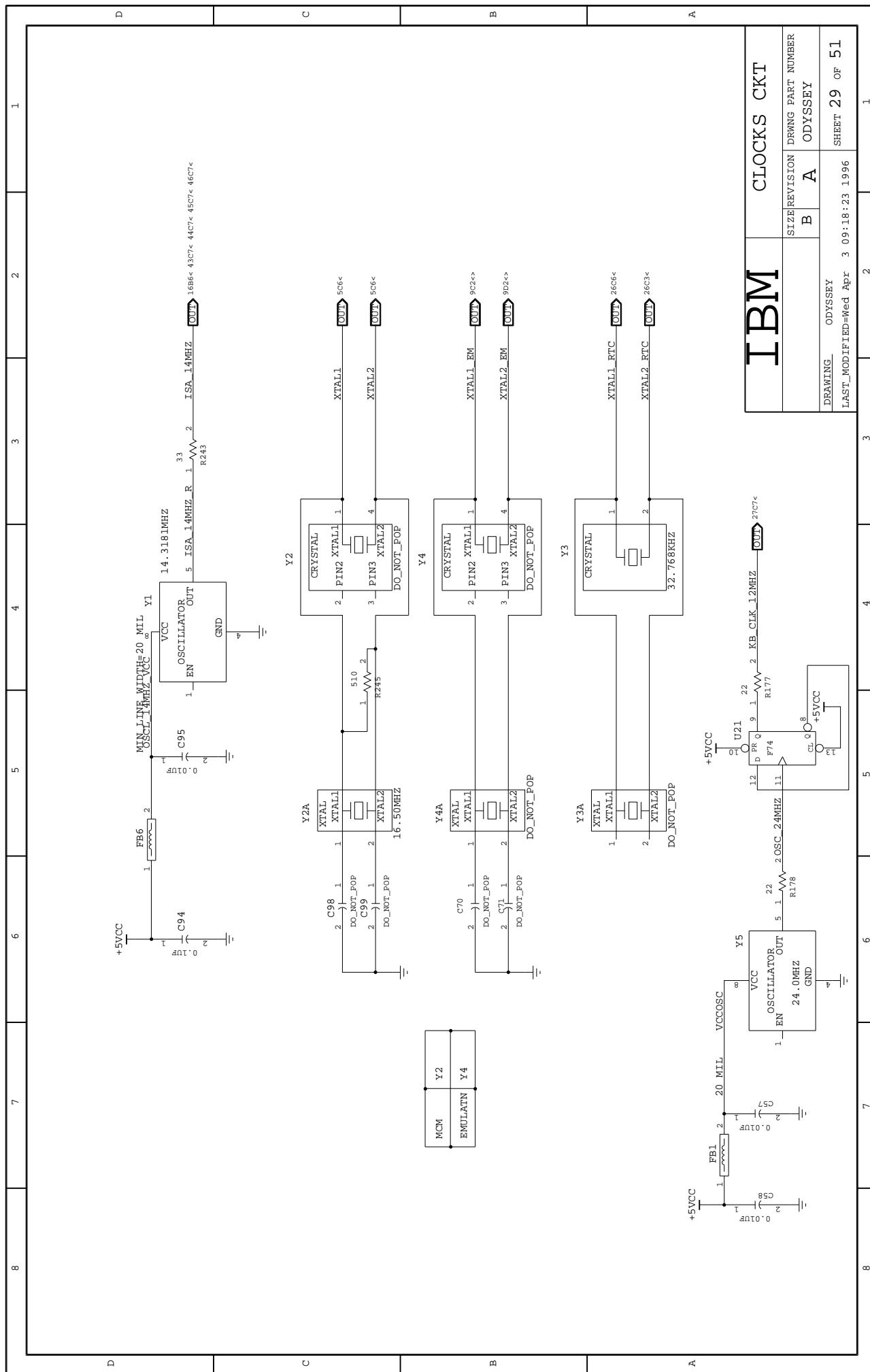










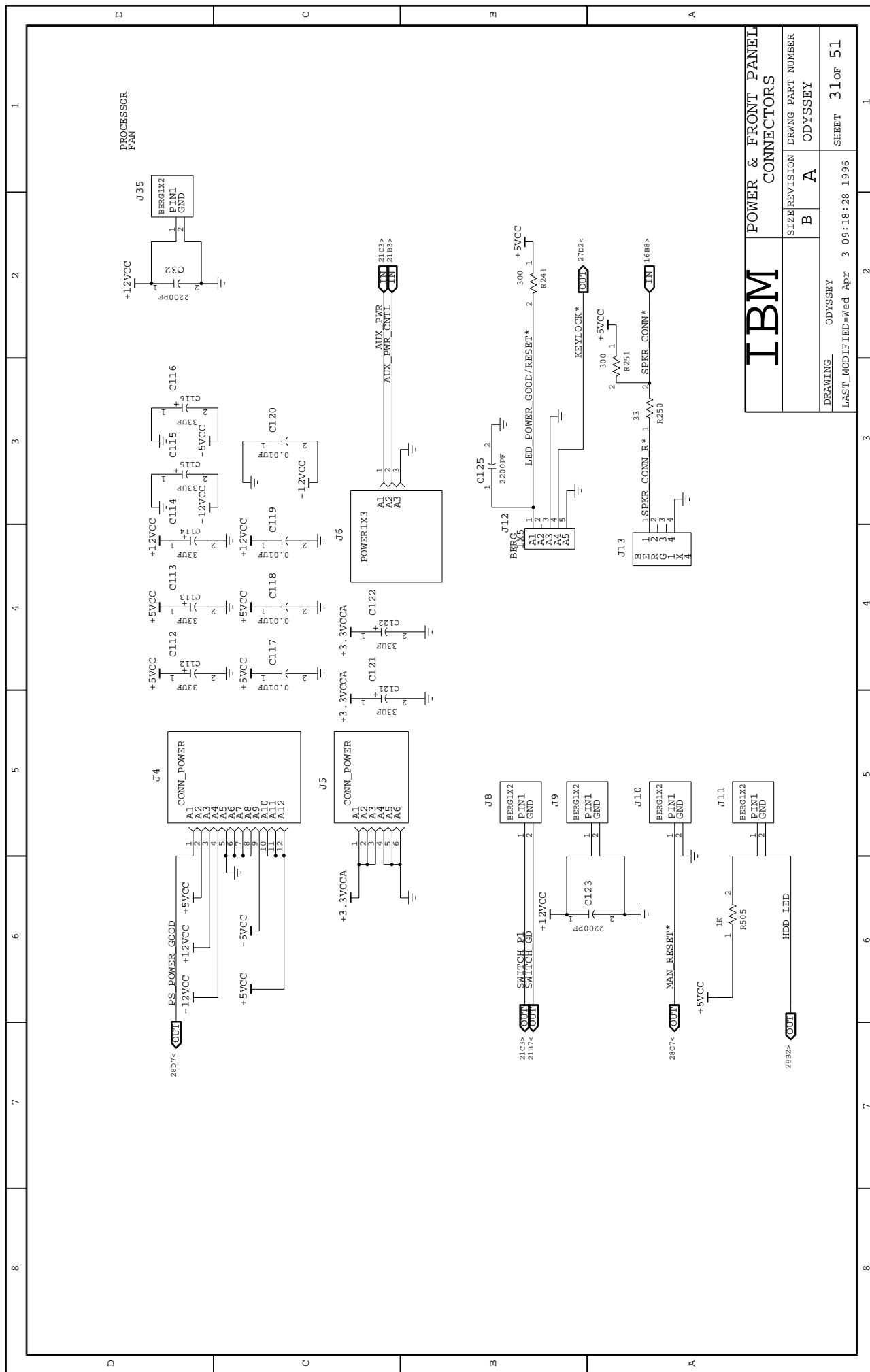


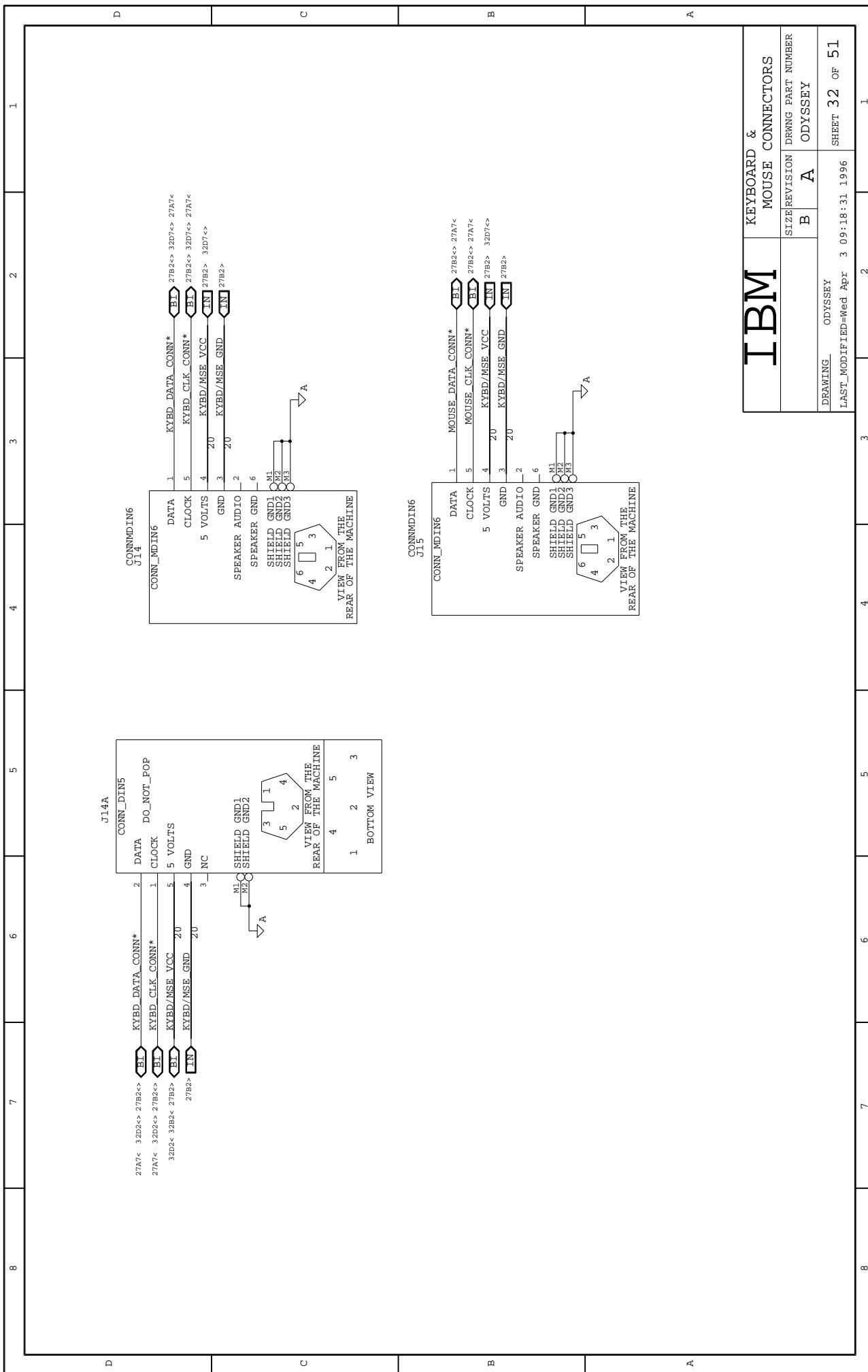
D	C	B	A
8	7	6	5
			4
			3
			2
			1

NOTE: VOLTAGE REGULATOR OUTPUT CAN BE VARIED BY CHANGING R9.

VOLATGE	R9
DO NOT POP	80 ₁₆ 10 ₈
2.5V	3.3V
3.6V	

TBM		VOLTAGE REGULATORS	
DRAWING	ODYSSEY	SIZE/REVISION	DRWNG PART NUMBER
B	A	ODY	ODYSSEY
LAST MODIFIED=Rev. Apr 3 09:18:25 1996		SHEET 30 OF 51	

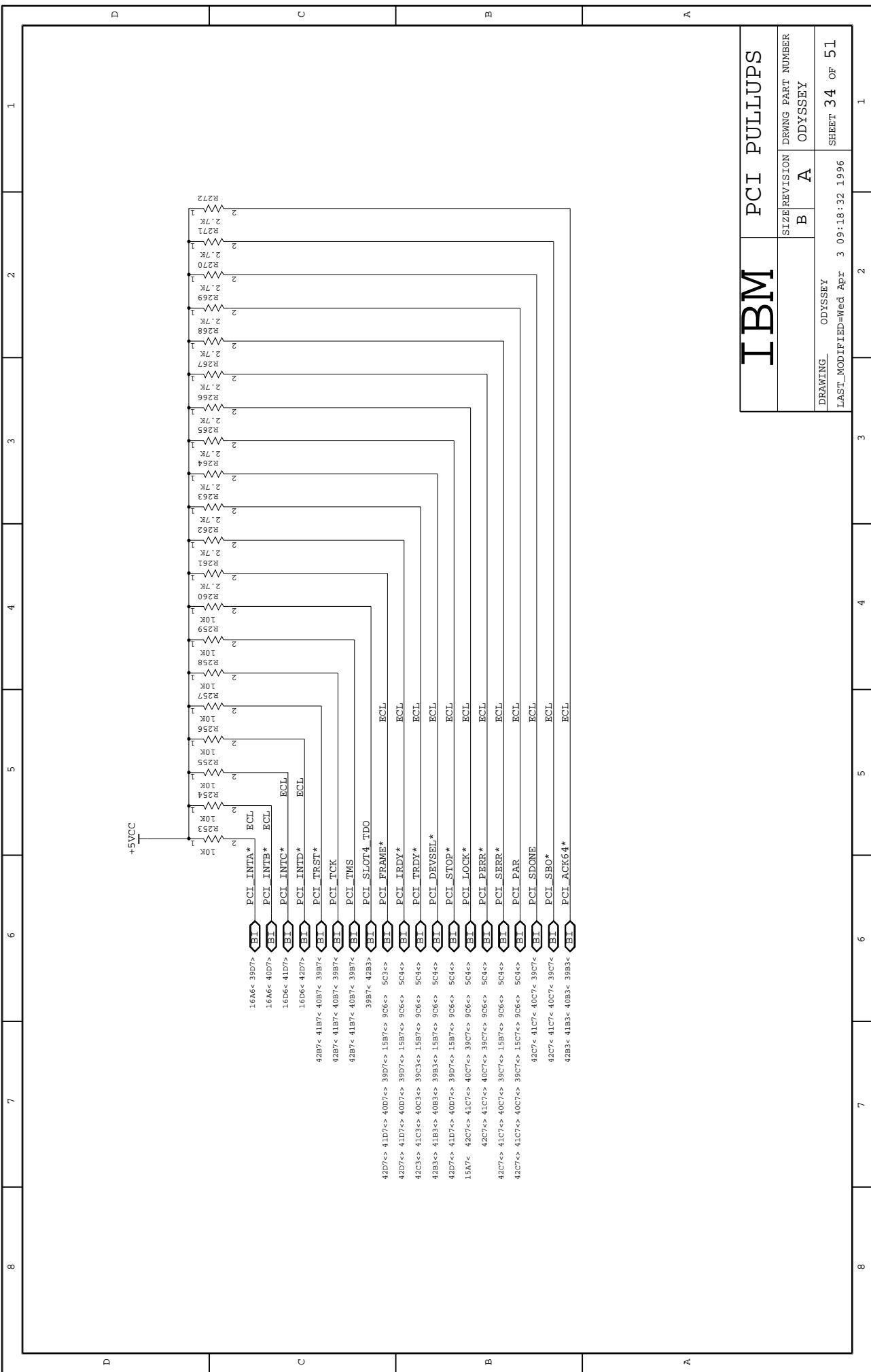




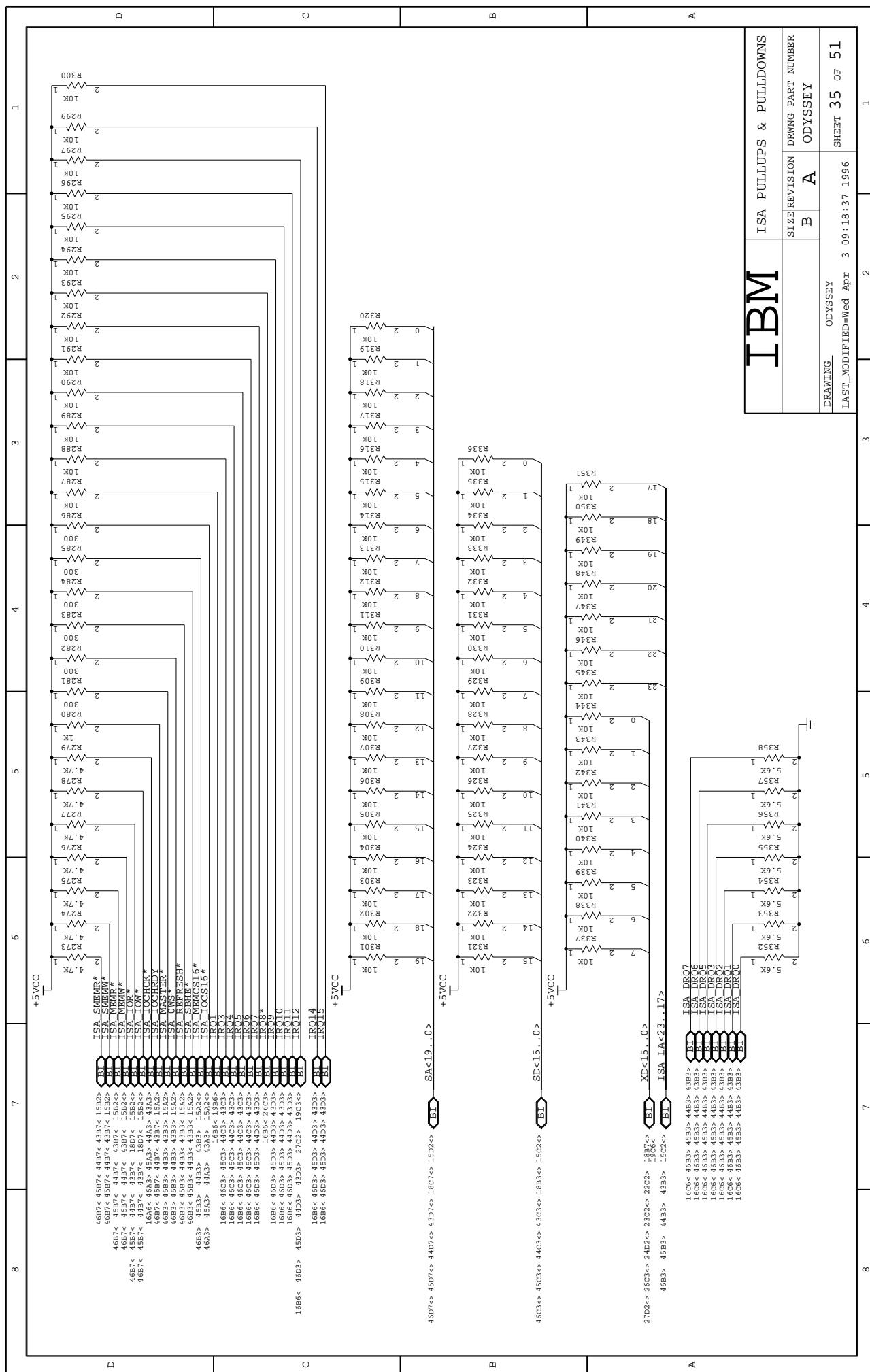
D	C	B	A																																
<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												

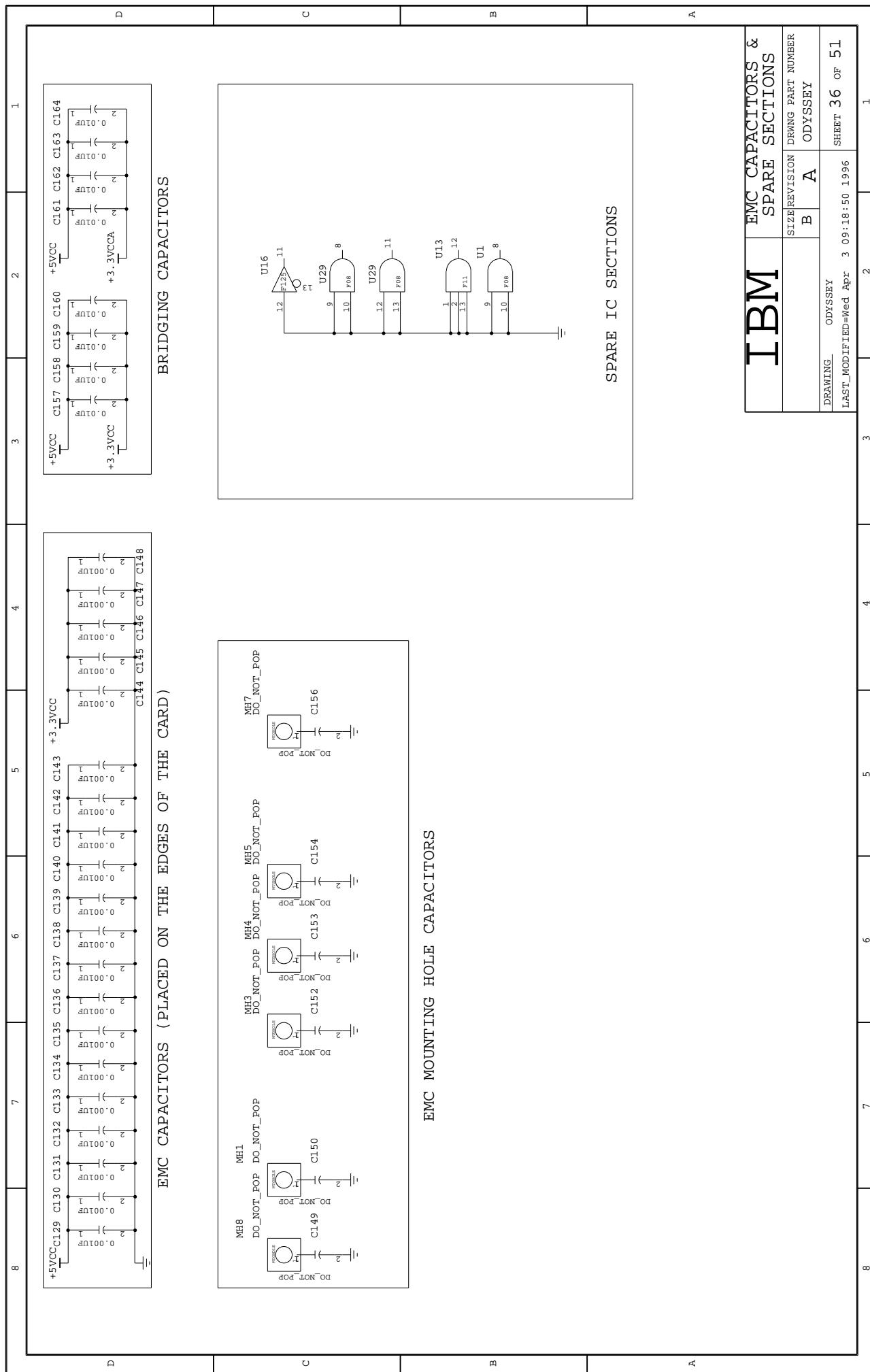
BLANK

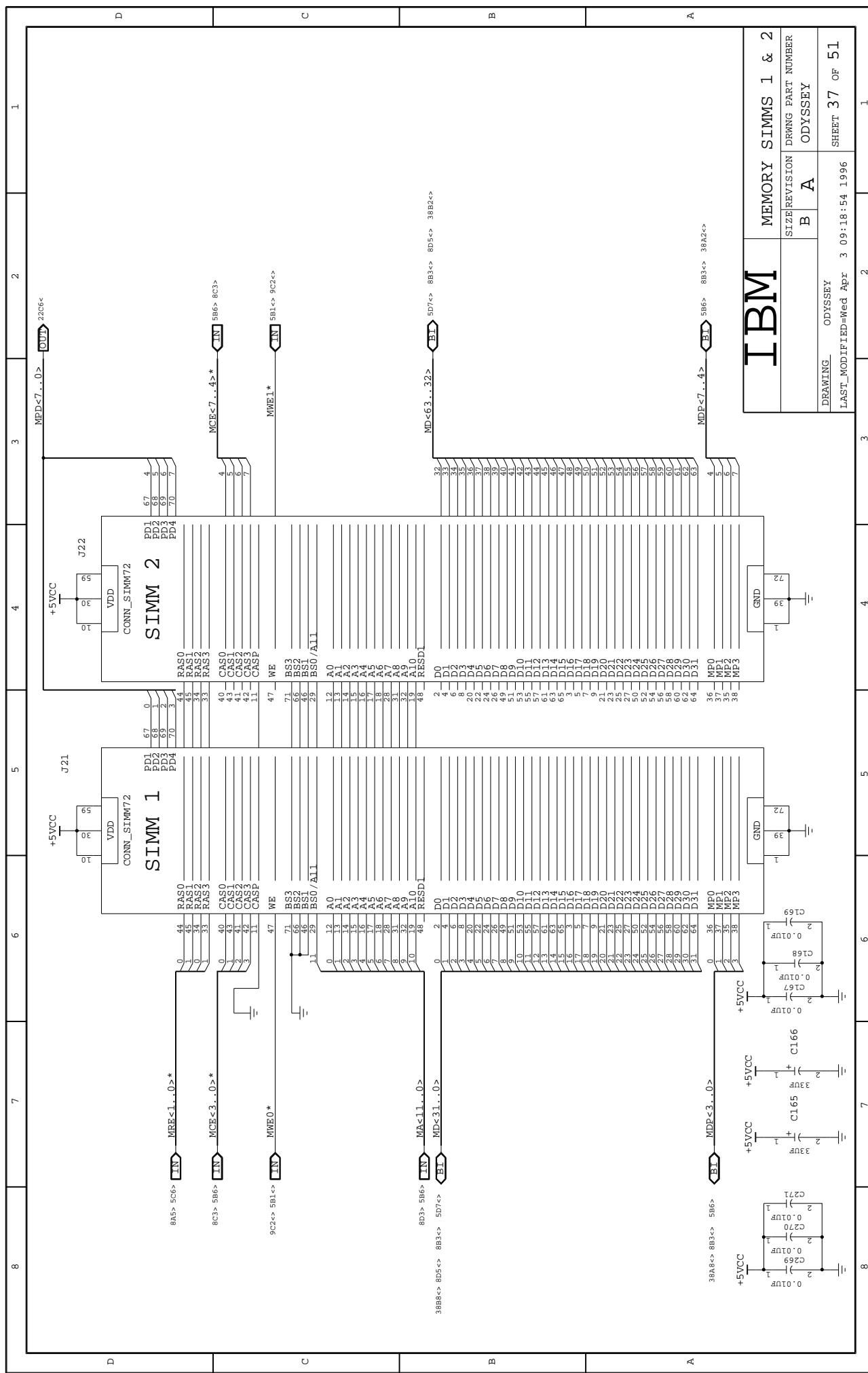
IBM		BLANK	
SIZE	REVISION	DRAWING PART NUMBER	
B	A	ODYSSEY	
DRAWING	ODYSSEY	LAST_MODIFIED=Mon Jan 29 16:04:21 1996	SHEET 33 OF 51

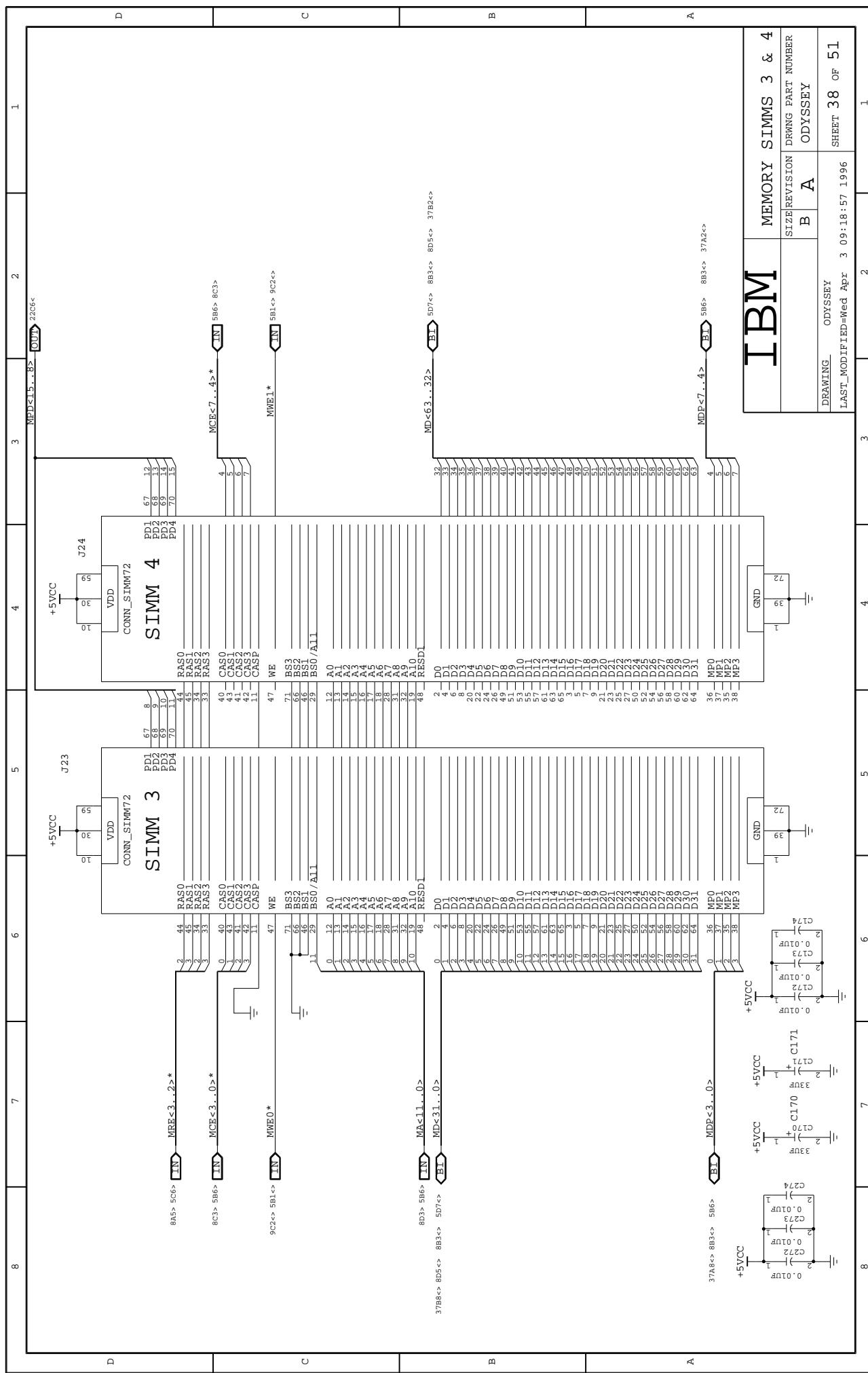


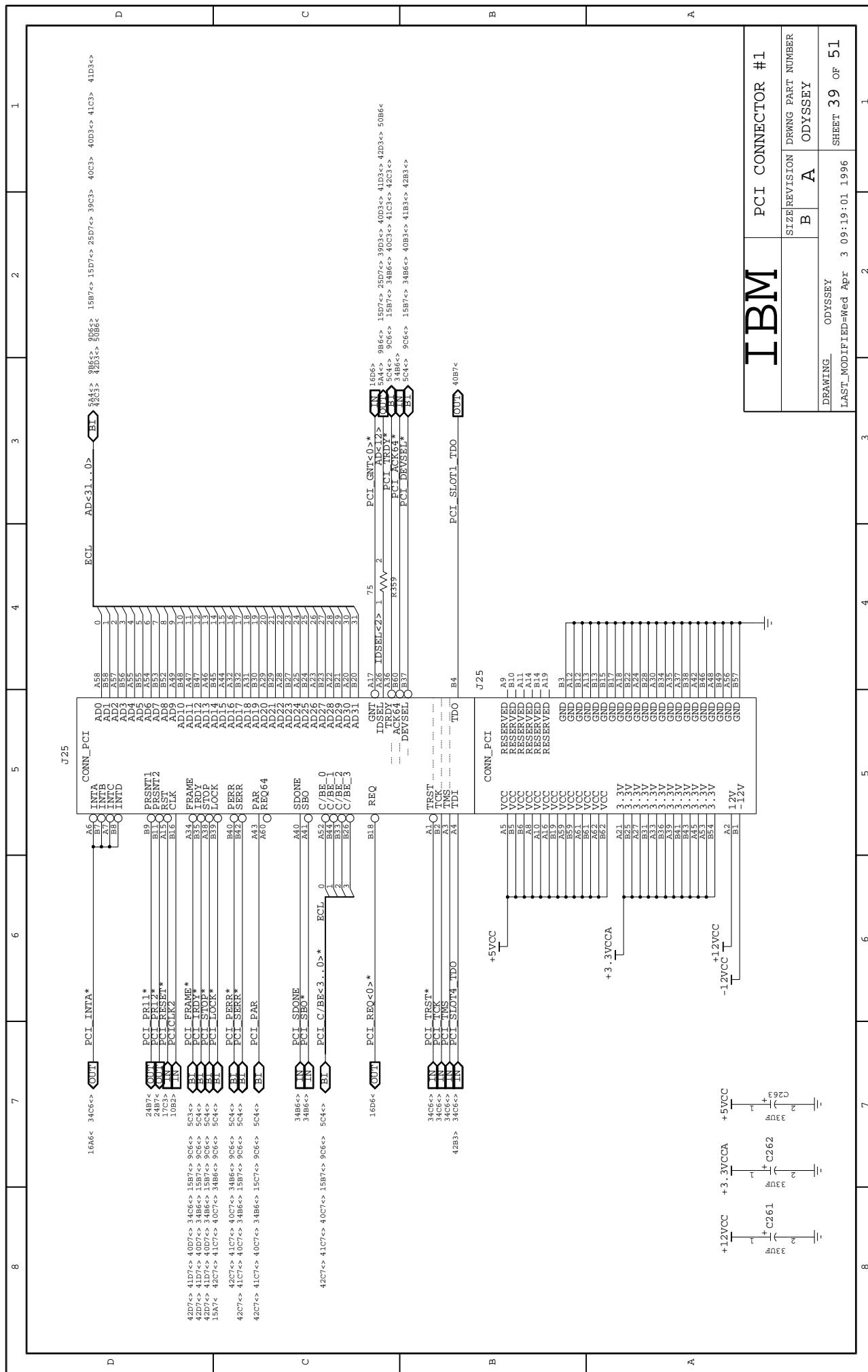
IBM PCI PULLUPS		DRAWING PART NUMBER		
SIZE	REVISION	B	A	ODYSSEY
LAST_MODIFIED=Wed Apr 3 09:18:32 1996	SHEET 34 OF 51	2	1	

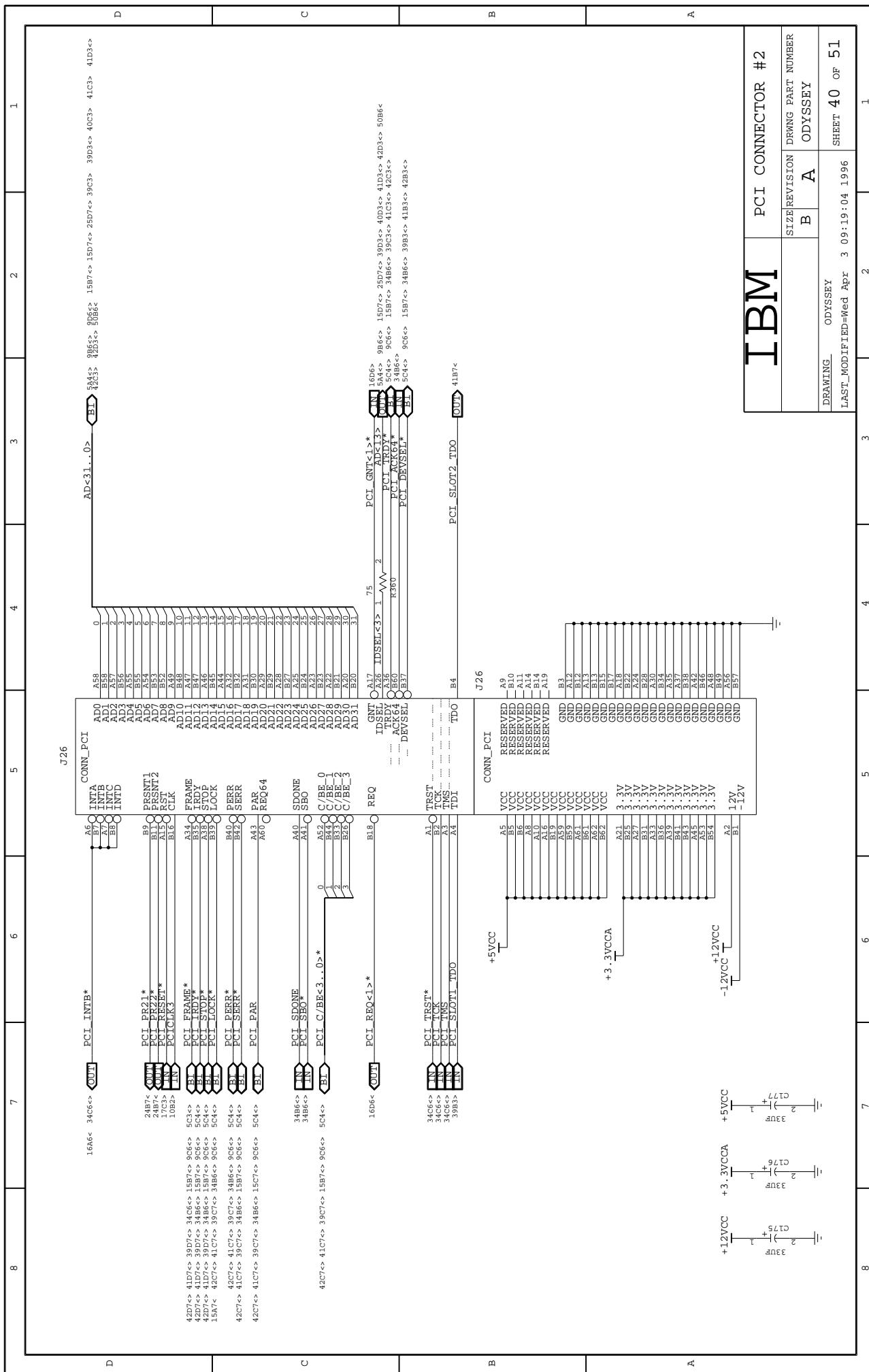


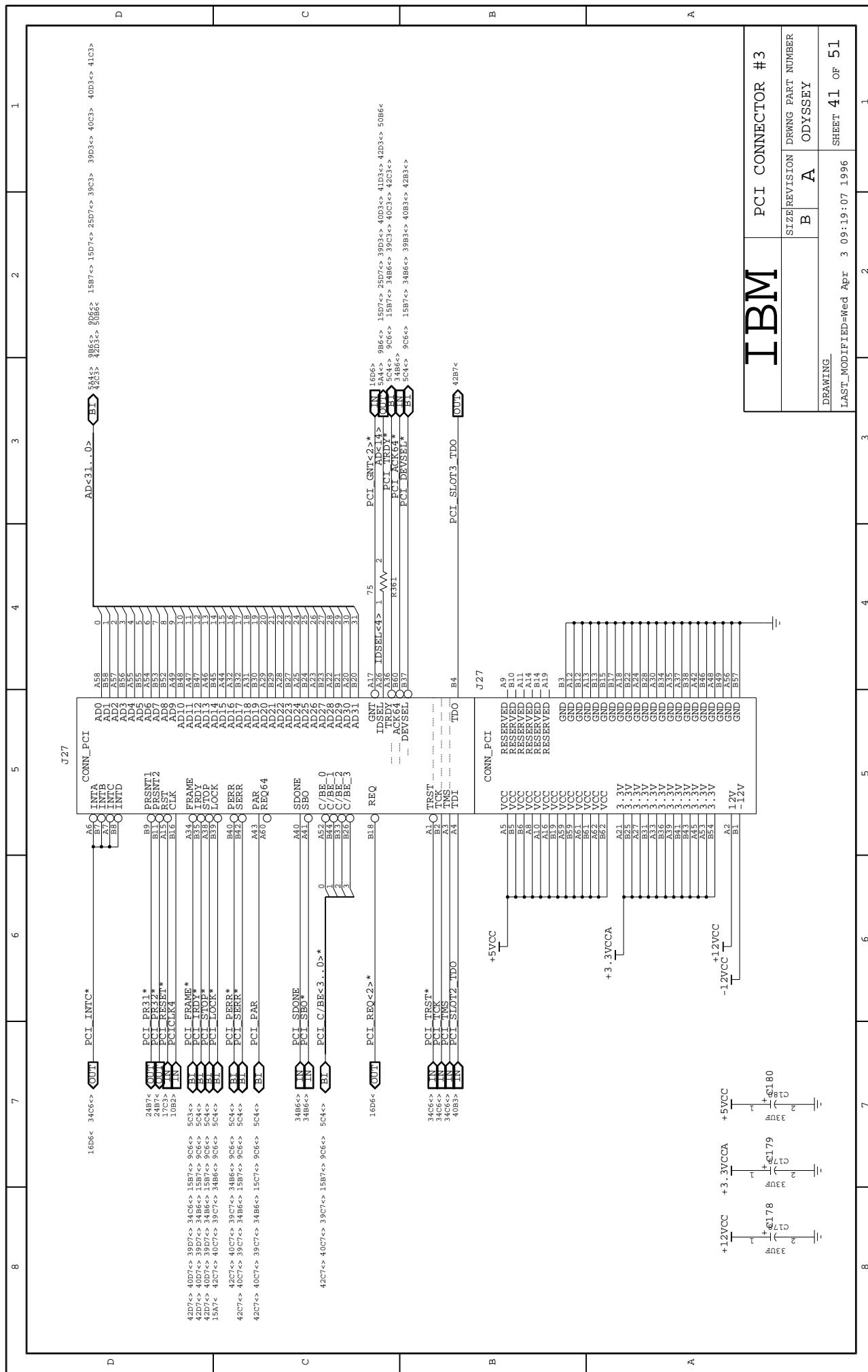


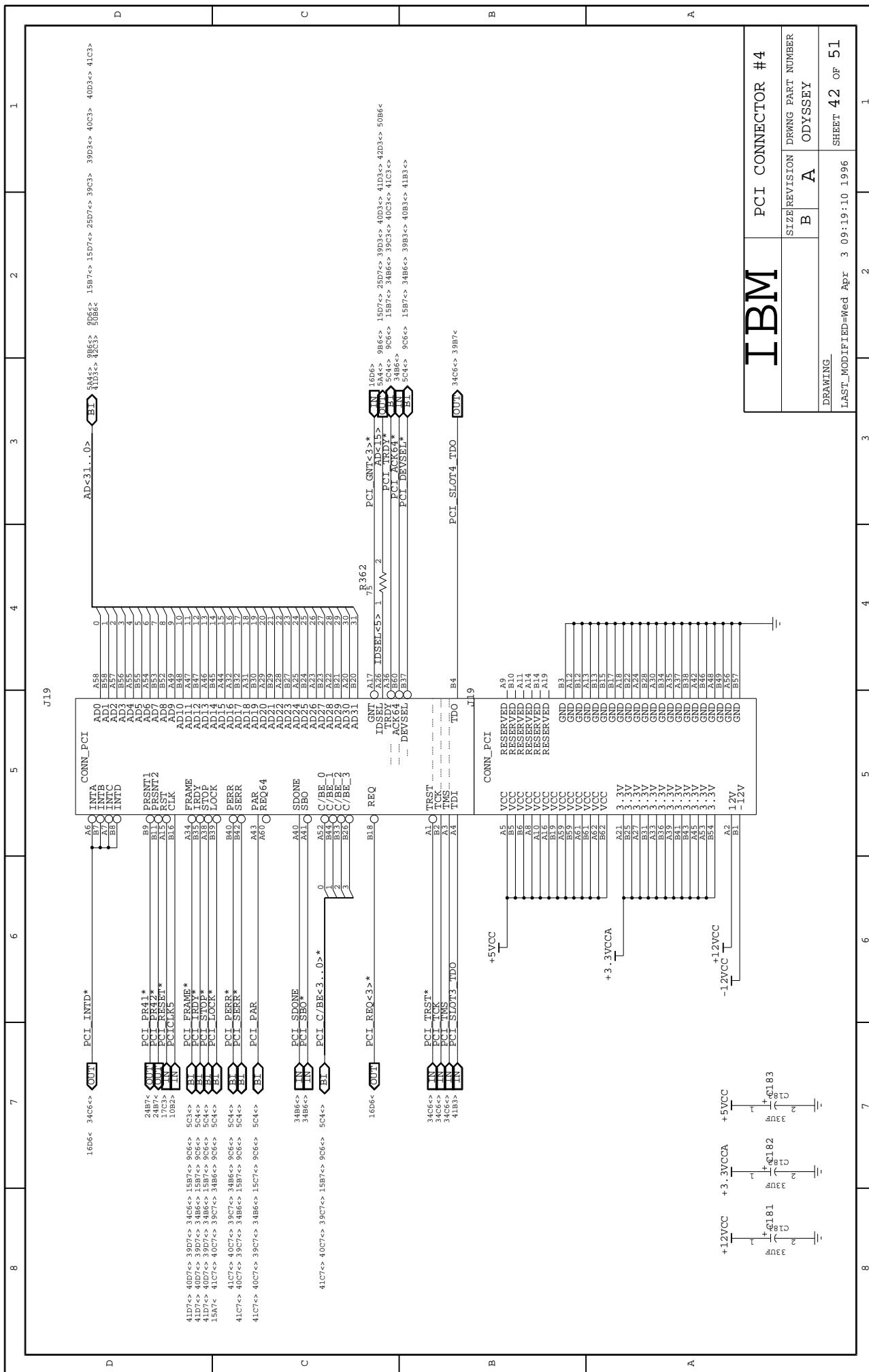




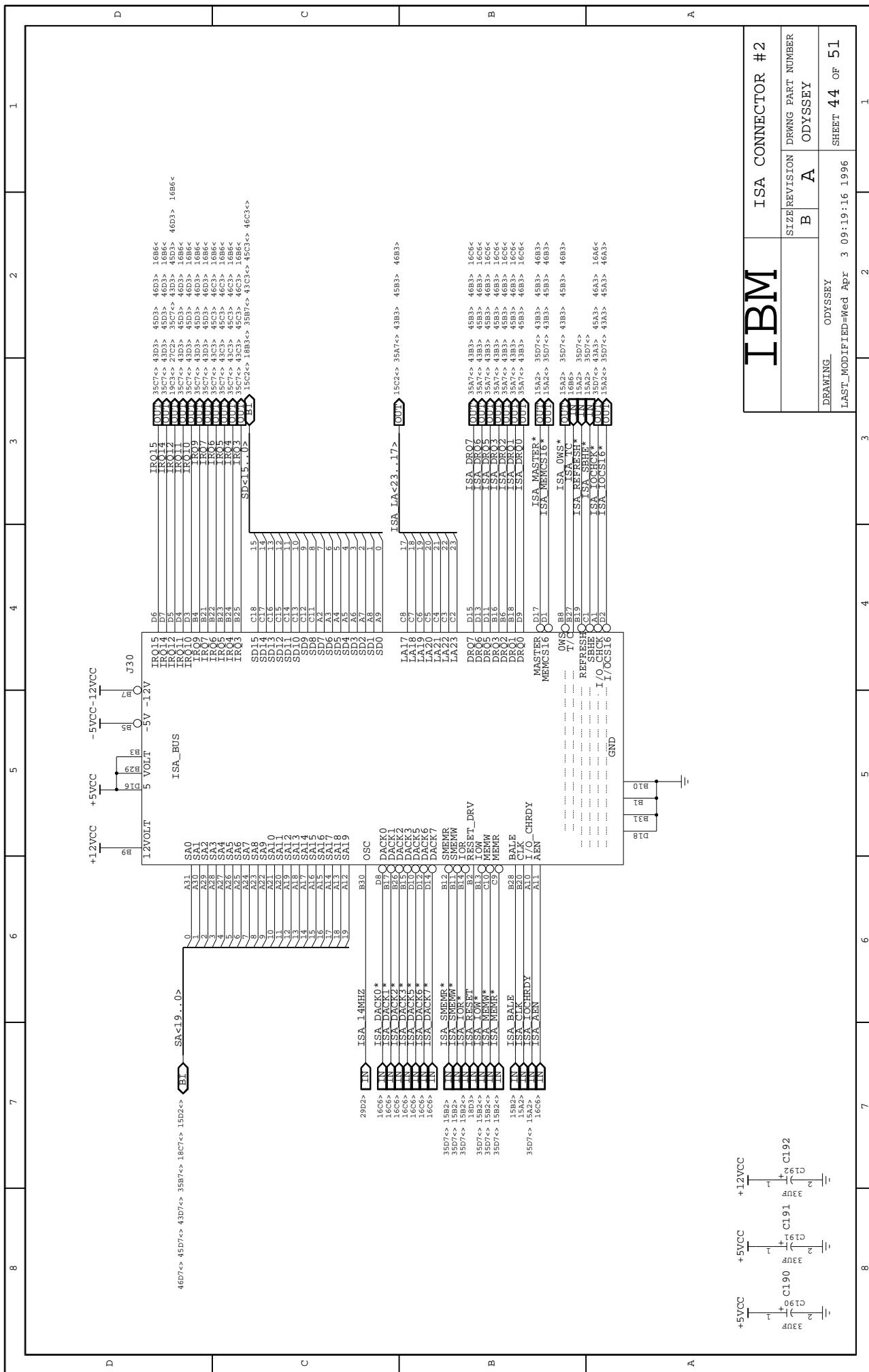




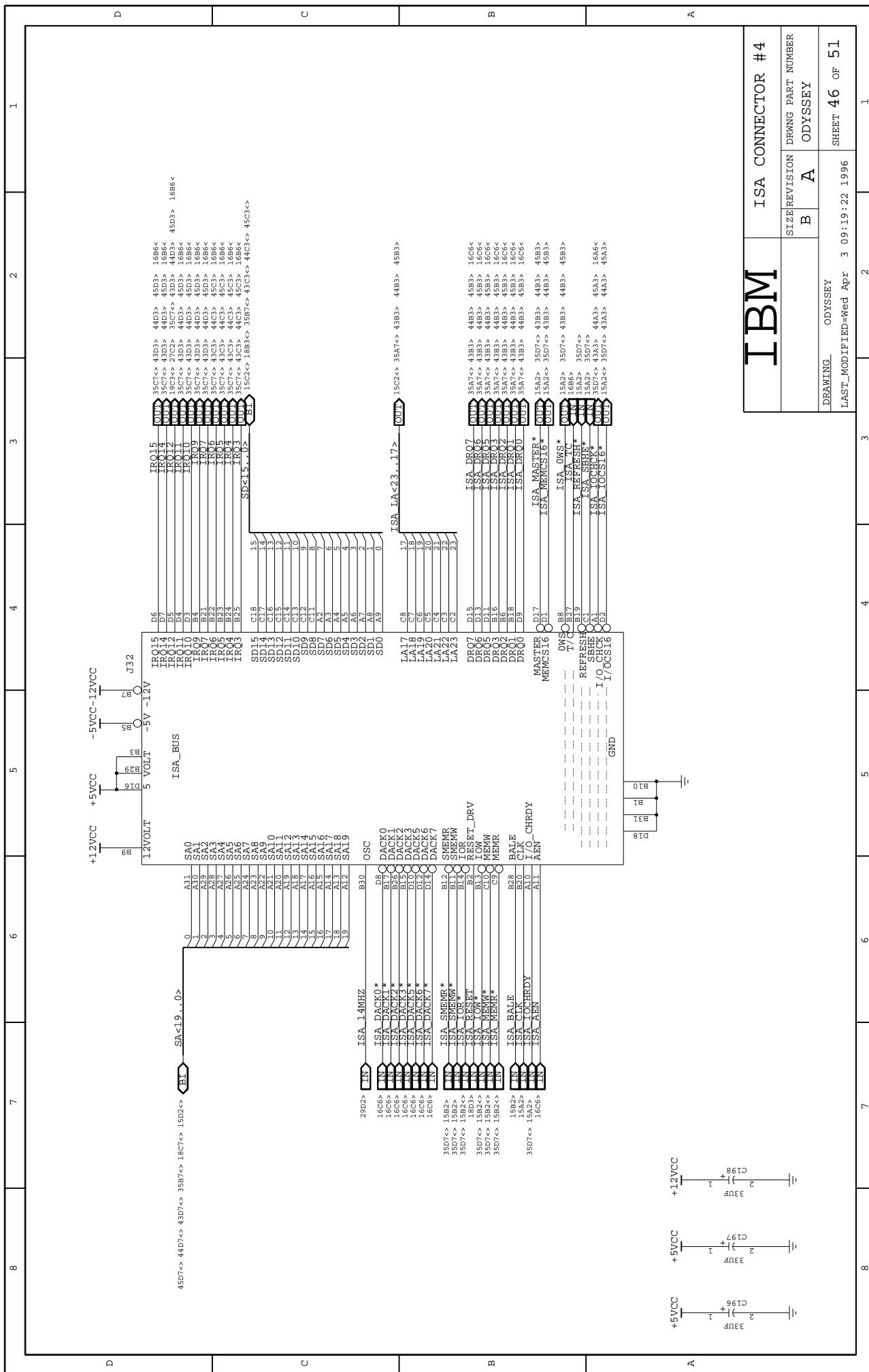












D		C		B		A	
8	7	6	5	4	3	2	1

BLANK

IBM		BLANK			
DRAWING LAST_MODIFIED=Thu Feb 8 15:00:41 1996	ODYSSEY	SIZE B	REVISION A	DRAWING PART NUMBER ODYSSEY	
				SHEET 47 OF 51	
8	7	6	5	4	3
					2
					1

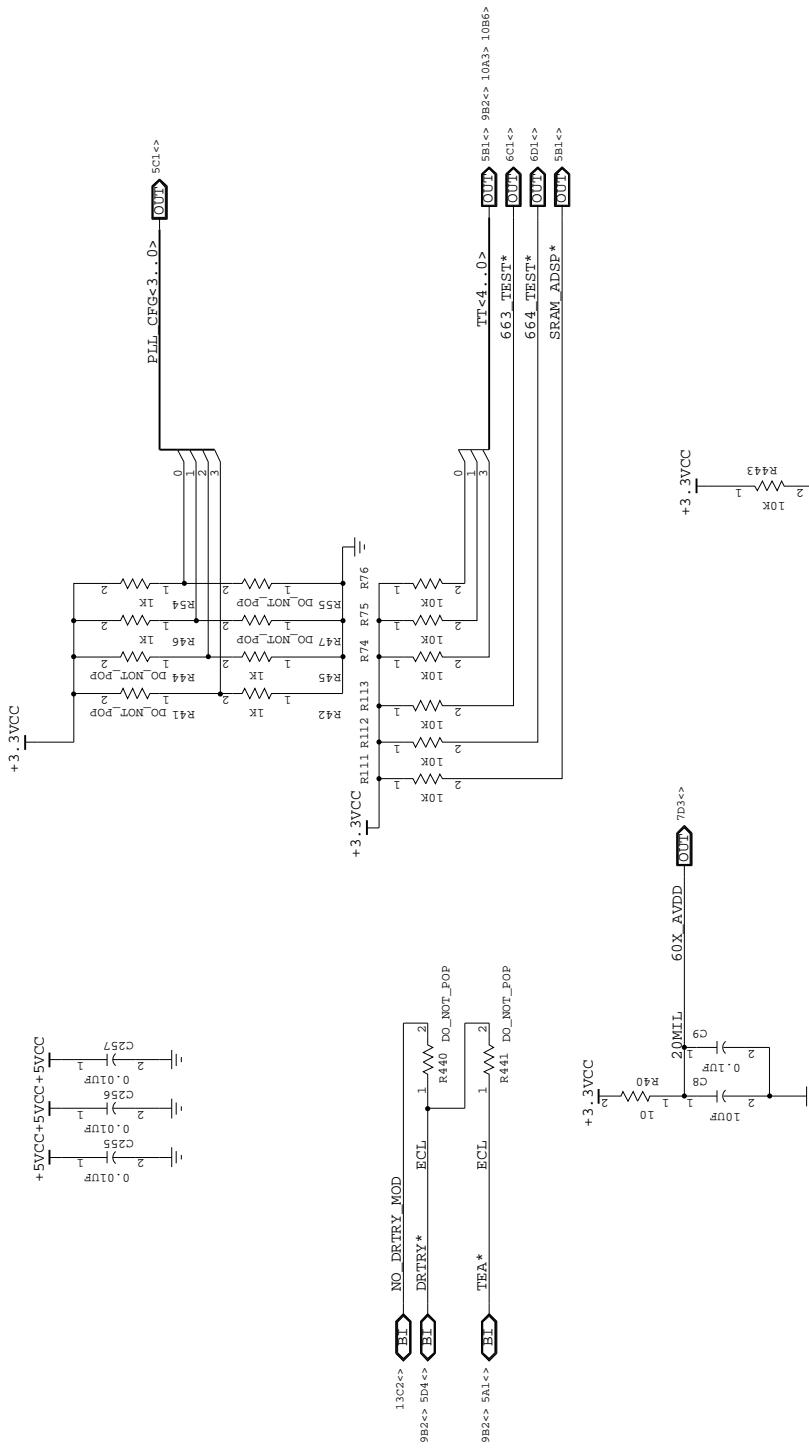
D	C	B	A																																
<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1	<table border="1" style="width: 100%; height: 100%;"> <tr><td>8</td><td>7</td><td>6</td><td>5</td><td>4</td><td>3</td><td>2</td><td>1</td></tr> </table>	8	7	6	5	4	3	2	1
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												
8	7	6	5	4	3	2	1																												

BLANK

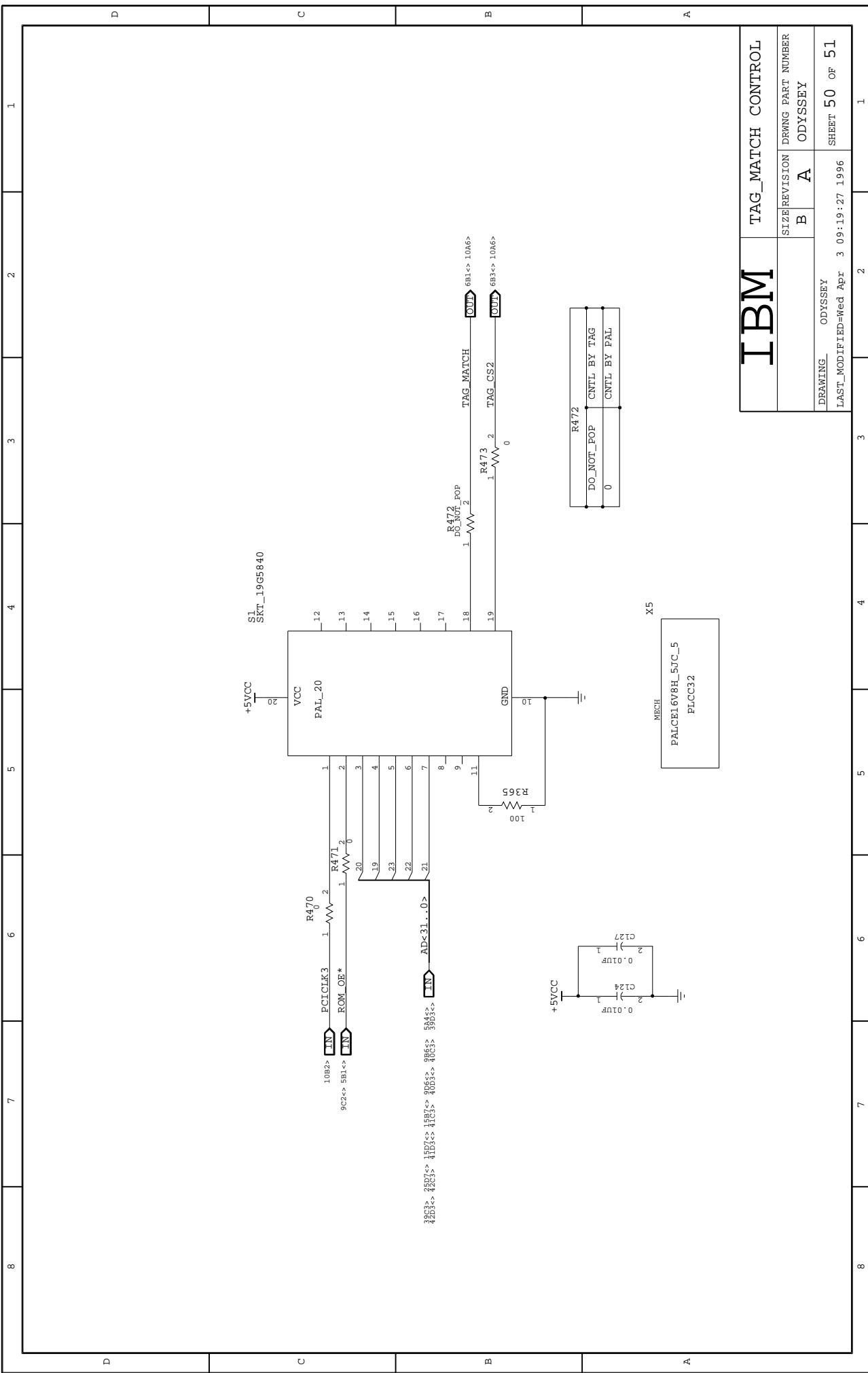
TBM		BLANK	
DRAWING	LAST_MODIFIED=Mon Jan 29 16:05:09 1996	SIZE	REVISION
		B	A

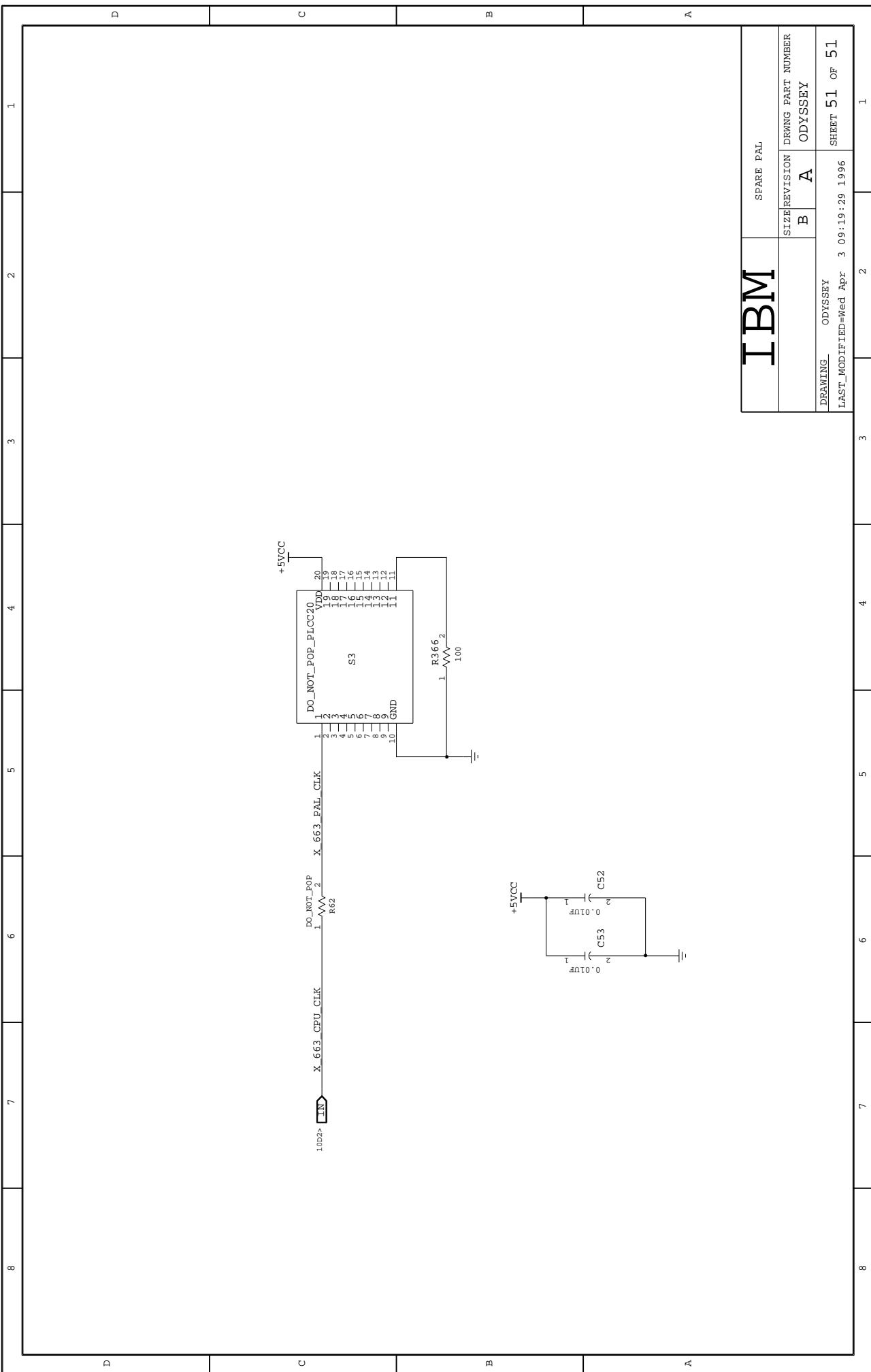
DRAWING : ODYSSEY SHEET 48 OF 51

LAST_MODIFIED=Mon Jan 29 16:05:09 1996



IBM		CONFIGURATION RESISTORS		
DRAWING	LAST_MODIFIED=Wed Apr 3 09:19:24 1996	SIZE B	REVISION A	DRIVING PART NUMBER ODYSSEY
				SHEET 49 OF 51





D		C		B		A	
8	7	6	5	4	3	2	1
PCL_DEVSEL *	5C4<> 906<> 15B7<> 34B6<> 39B3<>	RTC_WK *	1B9> 26G6<	USBR_BCIK0	942<		
PCL_EXT_SPL	40B3<> 41B3<> 42B3<>	RW_TINST *	1B94	USR_PCTCL1	6D3>	10A5<	
PCL_FRAME *	5C3<> 906<> 15B7<> 34C6<> 39D7<>	SA <1..>	15D2<> 18C7<> 35B7<> 43D7<> 44D7<>	USR_PCTCL1_EP	9C0>	10A5<	
PCL_FRAME_604 *	5C5 <3..0>	SBRH_S10 *	15D5<> 16D7<> 14D7<>	USR_PCTCL2	6C3>	10B5<	
PCL_GINT <3..0>	416D> 18A7<> 39E3<> 41C3<>	SD <1..>	15C2<> 1833<> 35B7<> 43C3<> 44C3<>	USR_PCTCL2_EP	9C0>	10B5<	
PCL_INTA *	42C3<>	SHD *	6B1<> 10A6<>	USR_PCTCL3	6D3>	10B5<	
PCL_INTB *	34C5<> 39D7> 16A6<	SMI *	5C4<> 946<> 10A6>	USR_PCTCL4	5C6>	10B5<	
PCL_INTC *	34C5<> 40D7> 16A6<	SPKR *	16B6	USR_PCTCL5	5C6>	10B5<	
PCL_INTD *	34C5<> 41D7> 16D6<	SPRS_CCONN_R *	31A2<-	USR_PCTCL6	6C3>	21C7< 26C6<	
PCL_INDE *	5C4<> 90E6<> 15B7<> 34B6<> 39D7<>	SPRAM_ADSR /ADRO	31A3	VCCOSC	29A7		
PCL_IIRDY *	40D7<> 41D7<> 42D7<>	SPRAM_ADSP *	5B1<> 49B3>	WBAT	5B1<>		
PCL_LOCK *	5C4<> 90E6<> 34B6<> 39C7<> 40C7<>	SPRAM_ALE	5C1> 10C5<	WT_60X *	18C3> 19D6<	27C7<	
PCL_OSL_OPEN	41C7<> 42C7<> 15A7<	SPRAM_BCIK0	5D1> 10C5<	XAI *	1A7..> 0>		
PCL_OUTI_SPL	6C1<>	SPRAM_BCIK1	5D4> 10C5<	XBFIF_EN *	1B85		
PCL_PAR	6C1<>	SPRAM_BCIK2	5D4> 10C5<	XD <1..1..0>	1B87<> 22C2<> 23C2<>	24D2<> 26C3<>	
PCL_PERR *	40C7<> 90E6<> 15C7<> 34B6<> 39C7<>	SPRAM_BCIK3	5D4> 10C5<	XD7D <2..2..0>	27D2<> 35A7<> 19C6<		
PCL_PERR1 *	41C7<> 90E6<> 34B6<> 39C7<> 40C7<>	SPRAM_CNT_EN/ADDR1	6B1<>	XDRN *	16C3.. 18B7..		
PCL_PERR2 *	41C7<> 90E6<> 34B6<> 39C7<> 40C7<>	SPRAM_CS *	6B3<> 10B6>	XDIR *	16C3.. 18B7..		
PCL_PERR3 *	41C7<> 90E6<> 34B6<> 39C7<> 40C7<>	SPRAM_CS *	6B1> 6B5<>	YLOW *	1B83.. 19D6.. 27C7..		
PCL_PERR4 *	41C7<> 90E6<> 34B6<> 39C7<> 40C7<>	SPRAM_EPR	7D2> 8C6<> 80B8<>	YTPMLI_EM	29C2.. 29B2..		
PCL_PERR5 *	40D5> 24B7<	SPRAM_EW	6B1<>	XTPMLI_EM	9C2..>		
PCL_PERR6 *	40D5> 24B7<	SPREFL_E0X *	5C1> 9B2<> 1B92>	XTPMLI_FPC	29B2.. 26B6..>		
PCL_PERR7 *	41D7> 24B7<	SPREFL_S10 *	16C3.. 13A2..>	XTPMLI_FPC	29B2.. 26B6..>		
PCL_PERR8 *	41D7> 24B7<	SWITCH_GD	31B7.. 21B7..>	XTPMLI_EM	29B2.. 26C3..>		
PCL_PERR9 *	42D7> 24B7<	SYSTCL_S10	15A4	XTPMLI_FPC	10D2.. 5A8<		
PCL_PERR10 *	42D7> 24B7<	TA *	6C1<> 17C3<> 18C9<	X_65.._PAL_CLK	51C6		
PCL_RSBT *	39C7.. 15C7..>	TAG_ADDR..13	6A4> 6B3..>	X_66.._CPU_CLK	10C2.. 5B6<		
PCL_RSBT *	40D7.. 15C7..>	TAG_A..IN	6C3.. 10B6..>	X_67.._CPU_CLK	10D2.. 5A6<		
PCL_RSBT3_TIO0	39C7.. 15C7..>	TAG_B..IN	6C3.. 10D5..>	X_SRAM_BCIK0	10C2.. 5C1..>		
PCL_RSBT4_TIO0	39C7.. 15C7..>	TAG_B..CLK	6C3..>	X_SRAM_BCIK1	10C2.. 5C1..>		
PCL_STOP *	5C4<> 90E6<> 15B7<> 34B6<> 39D7<>	TAG_CLEAR *	6B1..> 9B2..> 10C6..>	X_SRAM_BCIK2	10C2.. 5B6<		
PCL_STOP *	40D7.. 15C7..>	TAG_CS1 *	6B3..> 10A6..> 5B02..>	X_SRAM_BCIK3	10D2.. 5B6<		
PCL_STOP2_TIO0	39B3.. 40B3..>	TAG_DMTA..11	6B4.. 6B3..>	X_TNG_BCIK			
PCL_STOP3_TIO0	41B3.. 42B7..>	TAG_DTY	6B3..> 10A6..> 5B02..>				
PCL_STOP4_TIO0	34C5..> 39B3..> 39B7..>	TAG_MATCH	6B3..> 10A6..>				
PCL_STOP5_TIO0	34C5..> 39B7..> 41B7..> 42C7..>	TAG_PDNCH *	5B1..> 9B2..> 10C6..>				
PCL_STOP6_TIO0	34C5..> 39B7..> 41B7..> 42C7..>	TAG_SFINC	6B3..>				
PCL_STOP7_TIO0	34C5..> 39B7..> 41B7..> 42C7..>	TAG_TA *	6B3..>				
PCL_STOP8_TIO0	34C5..> 39B7..> 41B7..> 42C7..>	TAG_VALID	6B1..>				
PCL_STOP9_TIO0	34C5..> 39B7..> 41B7..> 42C7..>	TAG_VLD	6B3..>				
PCL_STOP10_TIO0	34C5..> 39B7..> 41B7..> 42C7..>	TAG_WK *	6B1..>				
PCL_STOP11_TIO0	34C5..> 39B7..> 41B7..> 42C7..>	TAG_WT	6B3..>				
PCL_STOP12_TIO0	6C3.. 10B5..>	TAG_WT_DTY..IN	5A1..> 10A6..>				
PCL_STOP13_TIO0	6C3.. 10B5..>	TAG_WT_DTY..OUT	5A1..> 10A6..>				
PCL_STOP14_TIO0	6C3.. 10B5..>	TAG_WT_DTY..S10	5A1..> 9B2..> 49B8..>				
PCL_STOP15_TIO0	6C3.. 10B5..>	TEST..S10	16B3				
PCL_STOP16_TIO0	6C3.. 10B5..>	TLBISYNC *	6C1..>				
PCL_STOP17_TIO0	6C3.. 10B5..>	TEST..*	6C3..>				
PCL_STOP18_TIO0	6C3.. 10B5..>	TC <1..1..0>	5C1..> 9C2..> 13B2..>				
PCL_STOP19_TIO0	6C3.. 10B5..>	TC_S10	16B5				
PCL_STOP20_TIO0	6C3.. 10B5..>	TDL	5C1..> 9C2..> 13B2..>				
PCL_STOP21_TIO0	6C3.. 10B5..>	TDO	5C1..> 9A2..> 13B7..>				
PCL_STOP22_TIO0	6C3.. 10B5..>	TEA *	5A1..> 9B2..> 49B8..>				
PCL_STOP23_TIO0	6C3.. 10B5..>	TEST..S10	16B3				
PCL_STOP24_TIO0	6C3.. 10B5..>	TLBISYNC *	6C1..>				
PCL_STOP25_TIO0	6C3.. 10B5..>	TEST..*	6C3..>				
PCL_STOP26_TIO0	6C3.. 10B5..>	TCR *	5C1..> 9A2..> 10A6..> 13C2..>				
PCL_STOP27_TIO0	6C3.. 10B5..>	TSIZ *	6C3..>				
PCL_STOP28_TIO0	6C3.. 10B5..>	TSIZ..<2..0..>	6B5..>				
PCL_STOP29_TIO0	6C3.. 10B5..>	TT..<4..0..>	5B1..> 9B2..> 10A3..> 10B6..> 49B3..>				
PCL_STOP30_TIO0	6C3.. 10B5..>	TT2..664	6C3..> 10A5..>				
PCL_STOP31_TIO0	6C3.. 10B5..>						
PCL_STOP32_TIO0	6C3.. 10B5..>						
PCL_STOP33_TIO0	6C3.. 10B5..>						
PCL_STOP34_TIO0	6C3.. 10B5..>						
PCL_STOP35_TIO0	6C3.. 10B5..>						
PCL_STOP36_TIO0	6C3.. 10B5..>						
PCL_STOP37_TIO0	6C3.. 10B5..>						
PCL_STOP38_TIO0	6C3.. 10B5..>						
PCL_STOP39_TIO0	6C3.. 10B5..>						
PCL_STOP40_TIO0	6C3.. 10B5..>						
PCL_STOP41_TIO0	6C3.. 10B5..>						
PCL_STOP42_TIO0	6C3.. 10B5..>						
PCL_STOP43_TIO0	6C3.. 10B5..>						
PCL_STOP44_TIO0	6C3.. 10B5..>						
PCL_STOP45_TIO0	6C3.. 10B5..>						
PCL_STOP46_TIO0	6C3.. 10B5..>						
PCL_STOP47_TIO0	6C3.. 10B5..>						
PCL_STOP48_TIO0	6C3.. 10B5..>						
PCL_STOP49_TIO0	6C3.. 10B5..>						
PCL_STOP50_TIO0	6C3.. 10B5..>						
PCL_STOP51_TIO0	6C3.. 10B5..>						
PCL_STOP52_TIO0	6C3.. 10B5..>						
PCL_STOP53_TIO0	6C3.. 10B5..>						
PCL_STOP54_TIO0	6C3.. 10B5..>						
PCL_STOP55_TIO0	6C3.. 10B5..>						
PCL_STOP56_TIO0	6C3.. 10B5..>						
PCL_STOP57_TIO0	6C3.. 10B5..>						
PCL_STOP58_TIO0	6C3.. 10B5..>						
PCL_STOP59_TIO0	6C3.. 10B5..>						
PCL_STOP60_TIO0	6C3.. 10B5..>						
PCL_STOP61_TIO0	6C3.. 10B5..>						
PCL_STOP62_TIO0	6C3.. 10B5..>						
PCL_STOP63_TIO0	6C3.. 10B5..>						
PCL_STOP64_TIO0	6C3.. 10B5..>						
PCL_STOP65_TIO0	6C3.. 10B5..>						
PCL_STOP66_TIO0	6C3.. 10B5..>						
PCL_STOP67_TIO0	6C3.. 10B5..>						
PCL_STOP68_TIO0	6C3.. 10B5..>						
PCL_STOP69_TIO0	6C3.. 10B5..>						
PCL_STOP70_TIO0	6C3.. 10B5..>						
PCL_STOP71_TIO0	6C3.. 10B5..>						
PCL_STOP72_TIO0	6C3.. 10B5..>						
PCL_STOP73_TIO0	6C3.. 10B5..>						
PCL_STOP74_TIO0	6C3.. 10B5..>						
PCL_STOP75_TIO0	6C3.. 10B5..>						
PCL_STOP76_TIO0	6C3.. 10B5..>						
PCL_STOP77_TIO0	6C3.. 10B5..>						
PCL_STOP78_TIO0	6C3.. 10B5..>						
PCL_STOP79_TIO0	6C3.. 10B5..>						
PCL_STOP80_TIO0	6C3.. 10B5..>						
PCL_STOP81_TIO0	6C3.. 10B5..>						
PCL_STOP82_TIO0	6C3.. 10B5..>						
PCL_STOP83_TIO0	6C3.. 10B5..>						
PCL_STOP84_TIO0	6C3.. 10B5..>						
PCL_STOP85_TIO0	6C3.. 10B5..>						
PCL_STOP86_TIO0	6C3.. 10B5..>						
PCL_STOP87_TIO0	6C3.. 10B5..>						
PCL_STOP88_TIO0	6C3.. 10B5..>						
PCL_STOP89_TIO0	6C3.. 10B5..>						
PCL_STOP90_TIO0	6C3.. 10B5..>						
PCL_STOP91_TIO0	6C3.. 10B5..>						
PCL_STOP92_TIO0	6C3.. 10B5..>						
PCL_STOP93_TIO0	6C3.. 10B5..>						
PCL_STOP94_TIO0	6C3.. 10B5..>						
PCL_STOP95_TIO0	6C3.. 10B5..>						
PCL_STOP96_TIO0	6C3.. 10B5..>						
PCL_STOP97_TIO0	6C3.. 10B5..>						
PCL_STOP98_TIO0	6C3.. 10B5..>						
PCL_STOP99_TIO0	6C3.. 10B5..>						
PCL_STOP100_TIO0	6C3.. 10B5..>						
PCL_STOP101_TIO0	6C3.. 10B5..>						
PCL_STOP102_TIO0	6C3.. 10B5..>						
PCL_STOP103_TIO0	6C3.. 10B5..>						
PCL_STOP104_TIO0	6C3.. 10B5..>						
PCL_STOP105_TIO0	6C3.. 10B5..>						
PCL_STOP106_TIO0	6C3.. 10B5..>						
PCL_STOP107_TIO0	6C3.. 10B5..>						
PCL_STOP108_TIO0	6C3.. 10B5..>						
PCL_STOP109_TIO0	6C3.. 10B5..>						
PCL_STOP110_TIO0	6C3.. 10B5..>						
PCL_STOP111_TIO0	6C3.. 10B5..>						
PCL_STOP112_TIO0	6C3.. 10B5..>						
PCL_STOP113_TIO0	6C3.. 10B5..>						
PCL_STOP114_TIO0	6C3.. 10B5..>						
PCL_STOP115_TIO0	6C3.. 10B5..>						
PCL_STOP116_TIO0	6C3.. 10B5..>						
PCL_STOP117_TIO0	6C3.. 10B5..>						
PCL_STOP118_TIO0	6C3.. 10B5..>						
PCL_STOP119_TIO0	6C3.. 10B5..>						
PCL_STOP120_TIO0	6C3.. 10B5..>						
PCL_STOP121_TIO0							

	8	7	6	5	4	3	2	1
D								
C								
B								
A								
*** Unit Cross-Reference ***								
-- -- for the entire design -- --								
B1	BATTERY 30A8							
C1	CAPACITOR 10A3							
C2	CAPACITOR 10A3							
C3	CAPACITOR 13C4 13C5							
C4	CAPACITOR 13C4							
C5	CAPACITOR 2BD6							
C55	CAPACITOR 2BD5							
C58	CAPACITOR 2BC6							
C59	CAPACITOR 2BC5							
C6	CAPACITOR 6C1							
C7	CAPACITOR 10A3							
C8	CAPACITOR 49A7							
C9	CAPACITOR 49A7							
C10	CAPACITOR 10D3							
C11	CAPACITOR 10D3							
C12	CAPACITOR 10D3							
C13	CAPACITOR 10C3							
C14	CAPACITOR 10C3							
C15	CAPACITOR 10B3							
C16	CAPACITOR 10B3							
C17	CAPACITOR 10C3							
C18	CAPACITOR 36D7							
C19	CAPACITOR 10B3							
C20	CAPACITOR 10B3							
C21	CAPACITOR 10B3							
C27	ELCAP							
C28	CAPACITOR 15A7							
C29	CAPACITOR 15A6							
C30	CAPACITOR 15A6							
C31	CAPACITOR 15A6							
C32	CAPACITOR 31D2							
C33	CAPACITOR 30B7							
C34	CAPACITOR 17C6							
C35	CAPACITOR 18C6							
C36	CAPACITOR 18B6							
C37	CAPACITOR 18A5							
C38	ELCAP							
C39	CAPACITOR 19A7							
C40	CAPACITOR 19A6 19A7							
C43	CAPACITOR 21B6							
C46	CAPACITOR 22B5 22C5							
C47	CAPACITOR 23B3							
C48	CAPACITOR 24C3							
C49	CAPACITOR 24B3							
C50	CAPACITOR 25A3 25A4							
C51	CAPACITOR 51A6							
C53	CAPACITOR 51A6							
C57	CAPACITOR 27C6							
C58	CAPACITOR 27A7							
C59	CAPACITOR 28A8							
C61	CAPACITOR 30C7							
C70	CAPACITOR 28B6							
C71	CAPACITOR 27A7							
C77	CAPACITOR 28A3							
C80	CAPACITOR 28C5							
C81	CAPACITOR 27C6							
C82	CAPACITOR 27A7							
C83	CAPACITOR 27B4							
C84	CAPACITOR 27A8							
C85	CAPACITOR 27A3							
C70	CAPACITOR 27A5							
C86	CAPACITOR 27A6							
C87	CAPACITOR 27A6							
C88	CAPACITOR 27A5							
C89	CAPACITOR 27A5							
C90	CAPACITOR 27A5							
C91	CAPACITOR 28D4							
C92	CAPACITOR 28C5							
C93	CAPACITOR 28C4							
C176	ELCAP	40A7						
C177	ELCAP	40A7						
C178	ELCAP	41A8						
C179	ELCAP	41A7						
C180	ELCAP	42A8						
C181	ELCAP	42A8						
C182	ELCAP	42A7						
C183	ELCAP	42A7						
C187	ELCAP	43A8						
C188	ELCAP	43A8						
C189	ELCAP	43A7						
C190	ELCAP	44A8						
C191	ELCAP	44A8						
C192	ELCAP	44A7						
C193	ELCAP	45A8						
C194	ELCAP	45A8						
C195	ELCAP	45A8						
C196	ELCAP	45A8						
C197	ELCAP	46A8						
C198	ELCAP	46A7						
C202	CAPACITOR 7A6							
C203	CAPACITOR 7A6							
C204	CAPACITOR 7A6							
C205	CAPACITOR 7A6							
C206	CAPACITOR 7A6							
C207	CAPACITOR 7A5							
C208	CAPACITOR 7A5							
C210	CAPACITOR 7A4							
C211	CAPACITOR 7A4							
C212	CAPACITOR 7A4							
C213	CAPACITOR 7A4							
C214	CAPACITOR 7A4							
C215	CAPACITOR 7A3							
C216	CAPACITOR BB3							
C217	CAPACITOR BB3							
C218	CAPACITOR BB2							
C219	CAPACITOR BB2							
C220	CAPACITOR BB2							
C221	CAPACITOR 8A3							
C222	CAPACITOR 8A3							
C223	CAPACITOR 8A2							
C224	CAPACITOR 8A2							
C225	CAPACITOR 8A2							
C226	CAPACITOR 8A2							
C227	CAPACITOR 49C7							
C228	CAPACITOR 49C7							
M1	WIRE	30A7						
M2	PLCC28KT	30C5						
M3	COPPER	30A7						
M4	RESISTOR	30C5						
M5	CONTRACT	30A7						
M6	MTHOLE	30A7						
M7	MTHOLE	30A7						
M8	MTHOLE	30A7						
M9	MTHOLE	30A7						
M10	MTHOLE	30A7						
M11	MTHOLE	30A7						
M12	MTHOLE	30A7						
M13	MTHOLE	30A7						
M14	MTHOLE	30A7						
M15	MTHOLE	30A7						
M16	MTHOLE	30A7						
M17	MTHOLE	30A7						
M18	MTHOLE	30A7						
M19	MTHOLE	30A7						
M20	MTHOLE	30A7						
M21	MTHOLE	30A7						
M22	MTHOLE	30A7						
M23	MTHOLE	30A7						
M24	MTHOLE	30A7						
M25	MTHOLE	30A7						
M26	MTHOLE	30A7						
M27	MTHOLE	30A7						
M28	MTHOLE	30A7						
M29	MTHOLE	30A7						
M30	MTHOLE	30A7						
M31	MTHOLE	30A7						
M32	MTHOLE	30A7						
M33	MTHOLE	30A7						
M34	MTHOLE	30A7						
M35	MTHOLE	30A7						
M36	MTHOLE	30A7						
M37	MTHOLE	30A7						
M38	MTHOLE	30A7						
M39	MTHOLE	30A7						
M40	MTHOLE	30A7						
M41	MTHOLE	30A7						
M42	MTHOLE	30A7						
M43	MTHOLE	30A7						
M44	MTHOLE	30A7						
M45	MTHOLE	30A7						
M46	MTHOLE	30A7						
M47	MTHOLE	30A7						
M48	MTHOLE	30A7						
M49	MTHOLE	30A7						
M50	MTHOLE	30A7						
M51	MTHOLE	30A7						
M52	MTHOLE	30A7						
M53	MTHOLE	30A7						
M54	MTHOLE	30A7						
M55	MTHOLE	30A7						
M56	MTHOLE	30A7						
M57	MTHOLE	30A7						
M58	MTHOLE	30A7						
M59	MTHOLE	30A7						
M60	MTHOLE	30A7						
M61	MTHOLE	30A7						
M62	MTHOLE	30A7						
M63	MTHOLE	30A7						
M64	MTHOLE	30A7						
M65	MTHOLE	30A7						
M66	MTHOLE	30A7						
M67	MTHOLE	30A7						
M68	MTHOLE	30A7						
M69	MTHOLE	30A7						
M70	MTHOLE	30A7						
M71	MTHOLE	30A7						
M72	MTHOLE	30A7						
M73	MTHOLE	30A7						
M74	MTHOLE	30A7						
M75	MTHOLE	30A7						
M76	MTHOLE	30A7						
M77	MTHOLE	30A7						
M78	MTHOLE	30A7						
M79	MTHOLE	30A7						
M80	MTHOLE	30A7						
M81	MTHOLE	30A7						
M82	MTHOLE	30A7						
M83	MTHOLE	30A7						
M84	MTHOLE	30A7						
M85	MTHOLE	30A7						
M86	MTHOLE	30A7						
M87	MTHOLE	30A7						
M88	MTHOLE	30A7						
M89	MTHOLE	30A7						
M90	MTHOLE	30A7						
M91	MTHOLE	30A7						
M92	MTHOLE	30A7						
M93	MTHOLE	30A7						
M94	MTHOLE							

A		B		C		D		E	
1	2	3	4	5	6	7	8	9	10
R1.3	RESISTOR 10A8	R9.0	RESISTOR 16D6	R17.6	RESISTOR 25B5	R28.9	RESISTOR 35D3	R3.9	RESISTOR 35D3
R1.4	RESISTOR 10D7	R9.1	RESISTOR 16D6	R17.7	RESISTOR 29A6	R29.0	RESISTOR 35D2	R3.91	RESISTOR 35D2
R1.5	RESISTOR 10D7	R9.2	RESISTOR 16D6	R17.8	RESISTOR 29A6	R29.2	RESISTOR 35D2	R3.92	RESISTOR 35D2
R1.6	RESISTOR 10D7	R9.3	RESISTOR 16D6	R17.9	RESISTOR 23C6	R29.3	RESISTOR 35D2	R3.93	RESISTOR 35D2
R1.7	RESISTOR 10D7	R9.4	RESISTOR 16A6	R18.0	RESISTOR 23B6	R29.4	RESISTOR 35D2	R3.94	RESISTOR 35D2
R1.8	RESISTOR 10D7	R9.5	RESISTOR 16D5	R18.1	RESISTOR 23C6	R29.5	RESISTOR 35D2	R3.95	RESISTOR 35D2
R1.9	RESISTOR 10D7	R9.6	RESISTOR 16D6	R18.2	RESISTOR 23B6	R29.6	RESISTOR 35D1	R3.96	RESISTOR 35D1
R2.0	RESISTOR 10D7	R9.7	RESISTOR 16D6	R18.3	RESISTOR 23C5	R29.7	RESISTOR 35D1	R3.97	RESISTOR 35D1
R2.1	RESISTOR 10D7	R9.8	RESISTOR 16D6	R18.4	RESISTOR 23B5	R29.8	RESISTOR 16A6	R3.98	RESISTOR 16A6
R2.2	RESISTOR 10D4	R9.9	RESISTOR 16A5	R18.5	RESISTOR 26B5	R29.9	RESISTOR 35D1	R3.99	RESISTOR 35D1
R2.3	RESISTOR 10D4	R1.00	RESISTOR 16A5	R18.6	RESISTOR 27B5	R3.00	RESISTOR 35D1	R3.01	RESISTOR 35D1
R2.4	RESISTOR 10D4	R1.01	RESISTOR 16D5	R18.7	RESISTOR 27B5	R3.02	RESISTOR 35C6	R3.03	RESISTOR 35C6
R2.5	RESISTOR 10D4	R1.02	RESISTOR 15A6	R18.8	RESISTOR 27B4	R3.04	RESISTOR 35C6	R3.05	RESISTOR 35C6
R2.6	RESISTOR 10D4	R1.03	RESISTOR 17C4	R18.9	RESISTOR 27B4	R3.06	RESISTOR 35C5	R3.07	RESISTOR 35C5
R2.7	RESISTOR 10C4	R1.04	RESISTOR 18D4	R19.0	RESISTOR 27A5	R3.08	RESISTOR 35C5	R3.09	RESISTOR 35C5
R2.8	RESISTOR 10C7	R1.05	RESISTOR 18D4	R19.1	RESISTOR 27A5	R3.10	RESISTOR 35C4	R3.11	RESISTOR 35C4
R2.9	RESISTOR 10C7	R1.06	RESISTOR 18D4	R19.2	RESISTOR 27A5	R3.12	RESISTOR 35C4	R3.13	RESISTOR 35C4
R3.0	RESISTOR 10C4	R1.07	RESISTOR 18C4	R19.3	RESISTOR 27A5	R3.14	RESISTOR 35C3	R3.15	RESISTOR 35C3
R3.1	RESISTOR 10C7	R1.08	RESISTOR 24C4	R19.4	RESISTOR 27A5	R3.16	RESISTOR 35C3	R3.17	RESISTOR 35C3
R3.2	RESISTOR 10C7	R1.09	RESISTOR 49C6	R19.5	RESISTOR 27A5	R3.18	RESISTOR 35C4	R3.19	RESISTOR 35C4
R3.3	RESISTOR 10C7	R1.10	RESISTOR 49C5	R19.6	RESISTOR 28D5	R3.20	RESISTOR 35C4	R3.21	RESISTOR 35C4
R3.4	RESISTOR 10C7	R1.11	RESISTOR 49C5	R19.7	RESISTOR 28D5	R3.22	RESISTOR 35C4	R3.23	RESISTOR 35C4
R3.5	RESISTOR 10C7	R1.12	RESISTOR 21B6	R19.8	RESISTOR 28D5	R3.24	RESISTOR 28C6	R3.25	RESISTOR 28C6
R3.6	RESISTOR 10C7	R1.13	RESISTOR 21B6	R19.9	RESISTOR 28D5	R3.26	RESISTOR 28D5	R3.27	RESISTOR 28D5
R3.7	RESISTOR 10B7	R1.20	RESISTOR 21B5	R20.0	RESISTOR 28D5	R3.28	RESISTOR 35B5	R3.29	RESISTOR 35B5
R3.8	RESISTOR 10B7	R1.21	RESISTOR 21B4	R20.1	RESISTOR 31B2	R3.30	RESISTOR 35B3	R3.31	RESISTOR 35B3
R3.9	RESISTOR 10B7	R1.22	RESISTOR 21B4	R20.2	RESISTOR 29D3	R3.32	RESISTOR 35C3	R3.33	RESISTOR 35C3
R4.0	RESISTOR 49B7	R1.23	RESISTOR 21B4	R20.3	RESISTOR 29C5	R3.34	RESISTOR 35C3	R3.35	RESISTOR 35C3
R4.1	RESISTOR 49C5	R1.24	RESISTOR 21B6	R20.4	RESISTOR 31A3	R3.36	RESISTOR 35C3	R3.37	RESISTOR 35C3
R4.2	RESISTOR 49C5	R1.25	RESISTOR 16A5	R20.5	RESISTOR 31A2	R3.38	RESISTOR 35C3	R3.39	RESISTOR 35C3
R4.3	RESISTOR 6A5	R1.30	RESISTOR 16A5	R20.6	RESISTOR 34D5	R3.40	RESISTOR 35C2	R3.41	RESISTOR 35C2
R4.4	RESISTOR 49C5	R1.35	RESISTOR 16D5	R20.7	RESISTOR 34D5	R3.42	RESISTOR 35C2	R3.43	RESISTOR 35C2
R4.5	RESISTOR 49C5	R1.36	RESISTOR 22C6	R20.8	RESISTOR 34D5	R3.44	RESISTOR 35B6	R3.45	RESISTOR 35B6
R4.6	RESISTOR 49C5	R1.37	RESISTOR 22C6	R20.9	RESISTOR 34D5	R3.46	RESISTOR 35B6	R3.47	RESISTOR 35B6
R4.7	RESISTOR 49C5	R1.38	RESISTOR 22C5	R21.0	RESISTOR 34D5	R3.48	RESISTOR 35B6	R3.49	RESISTOR 35B6
R4.8	RESISTOR 49C5	R1.39	RESISTOR 22C5	R21.1	RESISTOR 34D5	R3.50	RESISTOR 35B6	R3.51	RESISTOR 35B6
R4.9	RESISTOR 49C5	R1.40	RESISTOR 22C5	R21.2	RESISTOR 34D5	R3.52	RESISTOR 35B6	R3.53	RESISTOR 35B6
R5.0	RESISTOR 49C5	R1.41	RESISTOR 22C5	R21.3	RESISTOR 34D4	R3.54	RESISTOR 35B6	R3.55	RESISTOR 35B6
R5.1	RESISTOR 49C5	R1.42	RESISTOR 22C5	R21.4	RESISTOR 34D4	R3.56	RESISTOR 35B6	R3.57	RESISTOR 35B6
R5.2	RESISTOR 49C5	R1.43	RESISTOR 22C5	R21.5	RESISTOR 34D4	R3.58	RESISTOR 35B6	R3.59	RESISTOR 35B6
R5.3	RESISTOR 10A5	R1.44	RESISTOR 22C4	R21.6	RESISTOR 34D4	R3.60	RESISTOR 35B6	R3.61	RESISTOR 35B6
R5.4	RESISTOR 10A4	R1.45	RESISTOR 22C4	R21.7	RESISTOR 34D4	R3.62	RESISTOR 35B6	R3.63	RESISTOR 35B6
R5.5	RESISTOR 10A4	R1.46	RESISTOR 23C5	R21.8	RESISTOR 34D3	R3.64	RESISTOR 35B6	R3.65	RESISTOR 35B6
R5.6	RESISTOR 10A4	R1.47	RESISTOR 23C5	R21.9	RESISTOR 34D3	R3.66	RESISTOR 35B6	R3.67	RESISTOR 35B6
R5.7	RESISTOR 10B4	R1.48	RESISTOR 23C5	R22.0	RESISTOR 34D3	R3.68	RESISTOR 35B6	R3.69	RESISTOR 35B6
R5.8	RESISTOR 10B4	R1.49	RESISTOR 23C5	R22.1	RESISTOR 34D3	R3.70	RESISTOR 35B6	R3.71	RESISTOR 35B6
R5.9	RESISTOR 10B4	R1.50	RESISTOR 23C4	R22.2	RESISTOR 34D3	R3.72	RESISTOR 35B6	R3.73	RESISTOR 35B6
R6.0	RESISTOR 10B4	R1.51	RESISTOR 23B4	R22.3	RESISTOR 34D3	R3.74	RESISTOR 35B6	R3.75	RESISTOR 35B6
R6.1	RESISTOR 10B4	R1.52	RESISTOR 23B5	R22.4	RESISTOR 34D2	R3.76	RESISTOR 35B6	R3.77	RESISTOR 35B6
R6.2	RESISTOR 5L6	R1.53	RESISTOR 23B5	R22.5	RESISTOR 34D2	R3.78	RESISTOR 35B6	R3.79	RESISTOR 35B6
R6.3	RESISTOR 10C4	R1.54	RESISTOR 23C5	R22.6	RESISTOR 34D2	R3.80	RESISTOR 35B6	R3.81	RESISTOR 35B6
R6.4	RESISTOR 10B4	R1.55	RESISTOR 24C6	R22.7	RESISTOR 34D2	R3.82	RESISTOR 35B6	R3.83	RESISTOR 35B6
R6.5	RESISTOR 10B4	R1.56	RESISTOR 24C6	R22.8	RESISTOR 34D2	R3.84	RESISTOR 35B6	R3.85	RESISTOR 35B6
R6.6	RESISTOR 10B4	R1.57	RESISTOR 24C6	R22.9	RESISTOR 34D2	R3.86	RESISTOR 35B6	R3.87	RESISTOR 35B6
R6.7	RESISTOR 10B4	R1.58	RESISTOR 24C5	R23.0	RESISTOR 34D2	R3.88	RESISTOR 35B6	R3.89	RESISTOR 35B6
R6.8	RESISTOR 10B4	R1.59	RESISTOR 24C5	R23.1	RESISTOR 34D2	R3.90	RESISTOR 35B6	R3.91	RESISTOR 35B6
R6.9	RESISTOR 10B4	R1.70	RESISTOR 10A4	R23.2	RESISTOR 34D2	R3.92	RESISTOR 35B6	R3.93	RESISTOR 35B6
R7.0	RESISTOR 10A4	R1.71	RESISTOR 24C5	R23.3	RESISTOR 34D2	R3.94	RESISTOR 35B6	R3.95	RESISTOR 35B6
R7.1	RESISTOR 10A4	R1.72	RESISTOR 10A4	R23.4	RESISTOR 34D2	R3.96	RESISTOR 35B6	R3.97	RESISTOR 35B6
R7.2	RESISTOR 6C2	R1.73	RESISTOR 10A4	R23.5	RESISTOR 34D2	R3.98	RESISTOR 35B6	R3.99	RESISTOR 35B6
R7.3	RESISTOR 49C5	R1.74	RESISTOR 10A4	R23.6	RESISTOR 34D2	R4.00	RESISTOR 35B6	R4.01	RESISTOR 35B6
R7.4	RESISTOR 49C5	R1.75	RESISTOR 10A4	R23.7	RESISTOR 34D2	R4.02	RESISTOR 35B6	R4.03	RESISTOR 35B6
R7.5	RESISTOR 49C5	R1.76	RESISTOR 10A4	R23.8	RESISTOR 34D2	R4.04	RESISTOR 35B6	R4.05	RESISTOR 35B6
R7.6	RESISTOR 49C5	R1.77	RESISTOR 10B4	R23.9	RESISTOR 34D2	R4.06	RESISTOR 35B6	R4.07	RESISTOR 35B6
R7.7	RESISTOR 10B4	R1.78	RESISTOR 10B4	R24.0	RESISTOR 34D2	R4.08	RESISTOR 35B6	R4.09	RESISTOR 35B6
R7.8	RESISTOR 10A5	R1.79	RESISTOR 10B4	R24.1	RESISTOR 34D2	R4.10	RESISTOR 35B6	R4.11	RESISTOR 35B6
R7.9	RESISTOR 10B6	R1.80	RESISTOR 10B4	R24.2	RESISTOR 34D2	R4.12	RESISTOR 35B6	R4.13	RESISTOR 35B6
R8.0	RESISTOR 16C6	R1.81	RESISTOR 10B4	R24.3	RESISTOR 34D2	R4.14	RESISTOR 35B6	R4.15	RESISTOR 35B6
R8.1	RESISTOR 10A6	R1.82	RESISTOR 10B4	R24.4	RESISTOR 34D2	R4.16	RESISTOR 35B6	R4.17	RESISTOR 35B6
R8.2	RESISTOR 10B3	R1.83	RESISTOR 10B4	R24.5	RESISTOR 34D2	R4.18	RESISTOR 35B6	R4.19	RESISTOR 35B6
R8.3	RESISTOR 10B3	R1.84	RESISTOR 10B4	R24.6	RESISTOR 34D2	R4.20	RESISTOR 35B6	R4.21	RESISTOR 35B6
R8.4	RESISTOR 10B3	R1.85	RESISTOR 10B4	R24.7	RESISTOR 34D2	R4.22	RESISTOR 35B6	R4.23	RESISTOR 35B6
R8.5	RESISTOR 10B3	R1.86	RESISTOR 10B4	R24.8	RESISTOR 34D2	R4.24	RESISTOR 35B6	R4.25	RESISTOR 35B6
R8.6	RESISTOR 10B3	R1.87	RESISTOR 10B4	R24.9	RESISTOR 34D2	R4.26	RESISTOR 35B6	R4.27	RESISTOR 35B6
R8.7	RESISTOR 10A3	R1.88	RESISTOR 10B4	R25.0	RESISTOR 34D2	R4.28	RESISTOR 35B6	R4.29	RESISTOR 35B6
R8.8	RESISTOR 10A3	R1.89	RESISTOR 10A3	R25.1	RESISTOR 34D2	R4.30	RESISTOR 35A6	R4.31	RESISTOR 35A6
R8.9	RESISTOR 10A3	R1.90	RESISTOR 10A3	R25.2	RESISTOR 34D2	R4.32	RESISTOR 35A6	R4.33	RESISTOR 35A6

Appendix G Data Sheets

This section contains data sheets on selected components. IBM does not recommend the use of components from any particular manufacturer (other than IBM). The following data sheets are included as reference material only. Third party material is reproduced and included by permission of the copyright owner.

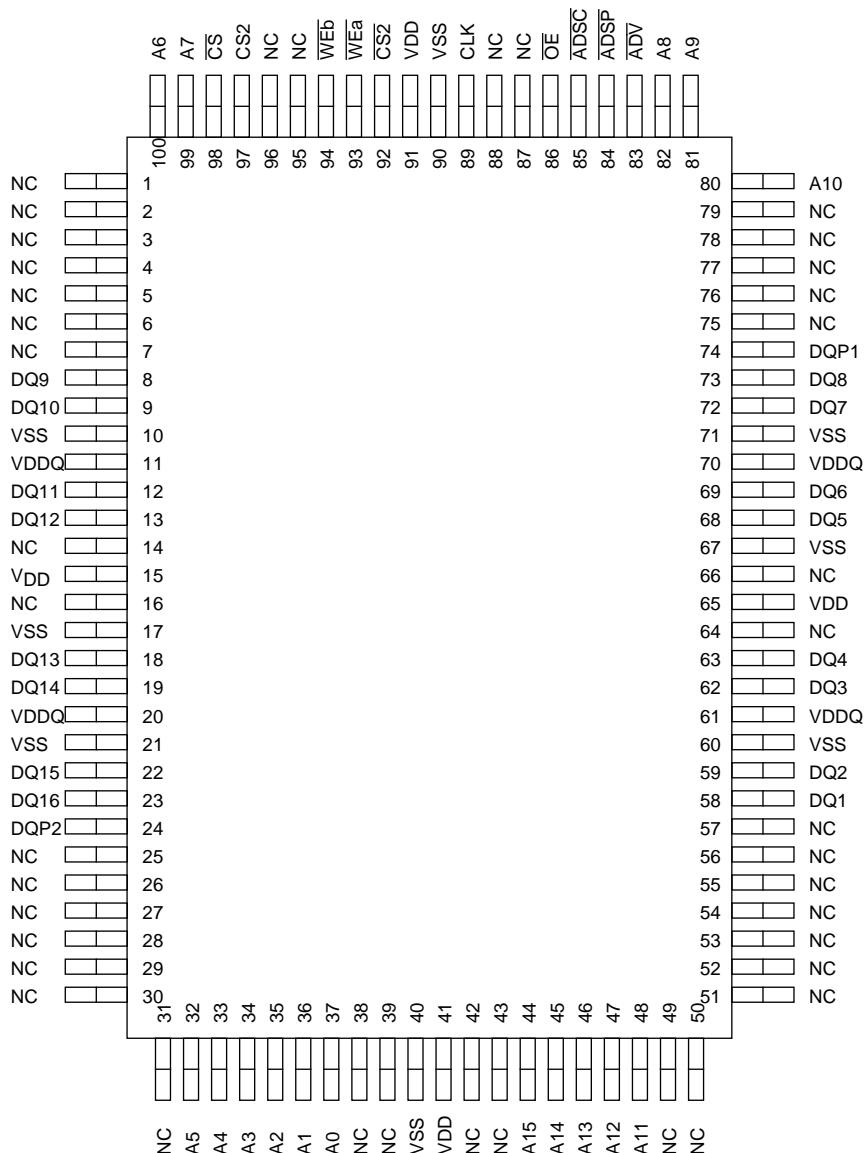
Features

- 64K x 18 Synchronous Burst Mode SRAM
- 0.5µ CMOS Technology
- Synchronous Burst Mode of Operation Compatible with PowerPC™ Processors
- LVTTL I/O Compatible with Common I/O
- Single +3.3 V ± 5% Power Supply and Ground
- Registered Addresses, Data Ins and Control Signals
- 5 V Tolerant I/O
- Asynchronous Output Enable
- Self-Timed Write Operation and Byte Write Capability
- Low Power Dissipation
 - 1.1 W Active at 83MHz
 - 90 mW Standby
- 100 Pin Thin Quad Flat Pack

Description

IBM Microelectronics 1M SRAM is a Synchronous Burstable, high performance CMOS Static RAM that is versatile, wide I/O, and achieves 8 nsec access. A single clock is used to initiate the read/write operation and all internal operations are self-timed. At the rising edge of the Clock, all Addresses, Data Ins and Control Signals are registered internally. Burst mode operation, compatible with PowerPC™ Processor's sequence, is accomplished by integrating input registers, internal 2-bit burst counter and high speed SRAM in a single chip. Burst reads are initiated with either ADSP or ADSC being LOW with a valid address during the rising edge of clock. Data from this address plus the three subsequent addresses will be output. The chip is operated with a single +3.3 V power supply and is compatible with LVTTL I/O interfaces.

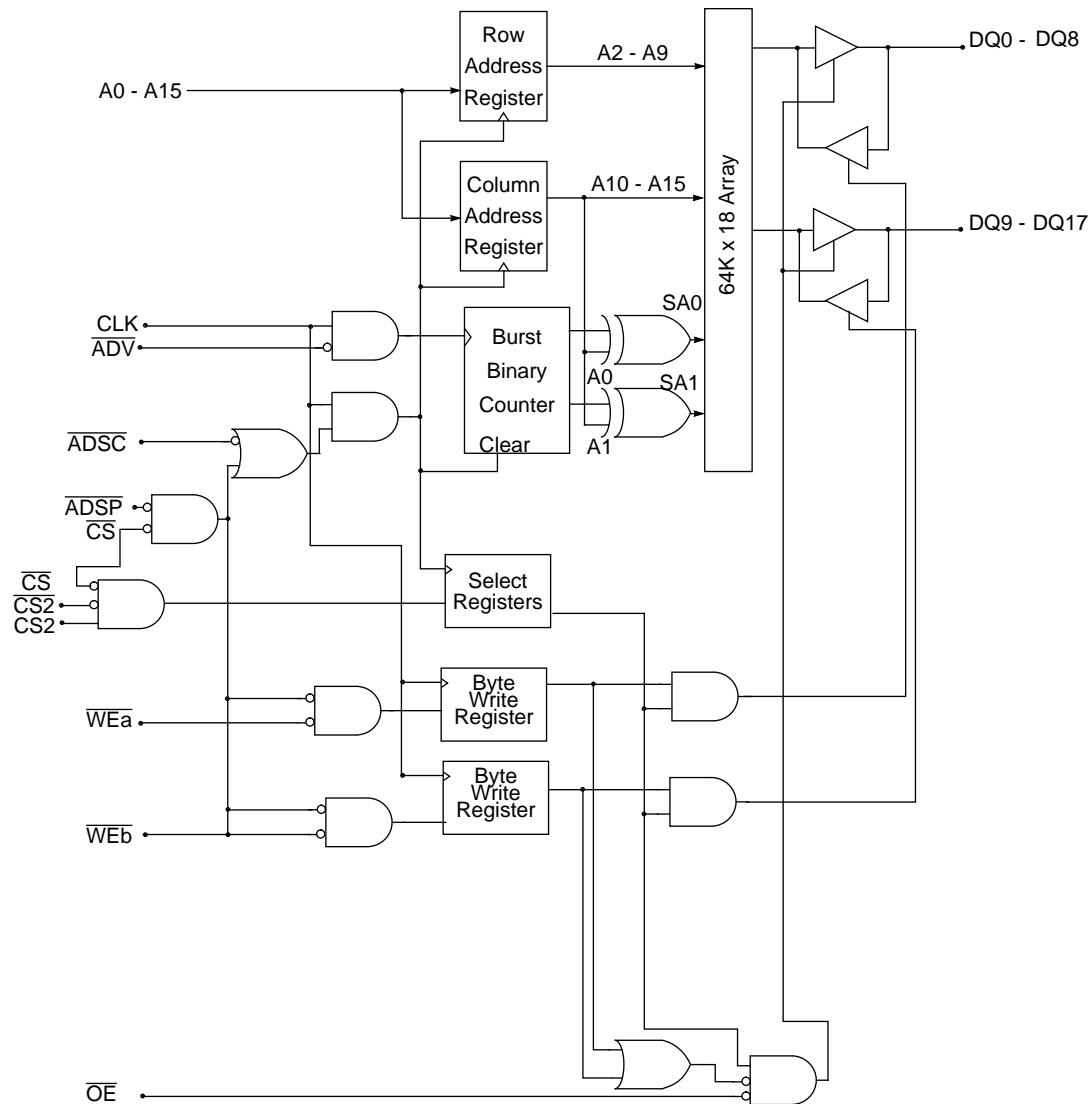
X18 TQFP Pin Array Layout



Pin Description

A0-A15	Address input	$\overline{\text{ADSP}}$	Address Status Processor
DQa - DQb	Data Input/Output (1-8 , 9-16)	$\overline{\text{ADSC}}$	Address status controller
CLK	Clock	$\overline{\text{ADV}}$	Burst Advance Control
$\overline{\text{WEa}}$	Write Enable, Byte a (1 to 8 & DQP1)	$\overline{\text{CS}}$	$\overline{\text{ADSP}}$ - Gated Chip Select
$\overline{\text{WEb}}$	Write Enable, Byte b (9 to 16 & DQP2)	V _{DD}	Power Supply (+3.3V)
$\overline{\text{OE}}$	Output Enable	V _{SS}	Ground
$\overline{\text{CS2}}$, CS2	Chip Selects	V _{DDQ}	Output Power Supply (+3.3V)
DQP1,DQP2	Parity bits for byte a, and byte b.	NC	No Connect

Block Diagram



Ordering Information

Part Number	Organization	Speed	Leads	Notes
IBM041814PQK-8	64K x 18	8 ns Access / 12 ns Cycle	100 pin TQFP	
IBM041814PQK-9	64K x 18	9 ns Access / 12 ns Cycle	100 pin TQFP	
IBM041814PQK-10	64K x 18	10 ns Access / 12 ns Cycle	100 pin TQFP	
IBM041814PQK-11	64K x 18	11 ns Access / 12 ns Cycle	100 pin TQFP	

Burst SRAM Clock Truth Table

CLK	\overline{CS}_2	CS2	\overline{CS}	\overline{ADSP}	\overline{ADSC}	\overline{ADV}	\overline{WE}	\overline{OE}	DQ	Operation
L→H	H	X	L	L	X	X	X	X	High-Z	Deselected Cycle
L→H	X	L	L	L	X	X	X	X	High-Z	Deselected Cycle
L→H	H	X	X	X	L	X	X	X	High-Z	Deselected Cycle
L→H	X	L	X	X	L	X	X	X	High-Z	Deselected Cycle
L→H	L	H	L	L	X	X	X	L	Q	Read from External Address, Begin Burst
L→H	L	H	L	L	X	X	X	H	High-Z	Read from External Address, Begin Burst
L→H	L	H	L	H	L	X	H	L	Q	Read from External Address, Begin Burst
L→H	L	H	L	H	L	X	L	X	D	Write to External Address, Begin Burst
L→H	X	X	X	H	H	L	H	L	Q	Read from next Add., Continue Burst
L→H	X	X	X	H	H	L	L	X	D	Write to next Add., Continue Burst
L→H	X	X	X	H	H	H	H	L	Q	Read from Current Add., Suspend Burst
L→H	X	X	X	H	H	H	L	X	D	Write to Current Add., Suspend Burst
L→H	X	X	H	X	L	X	X	X	High-Z	Deselect Cycle
L→H	X	X	H	X	H	L	H	L	Q	Read from next Add., Continue Burst
L→H	X	X	H	X	H	L	L	X	D	Write to next Add., Continue Burst
L→H	X	X	H	X	H	H	H	L	Q	Read from current Add., Suspend Burst
L→H	X	X	H	X	H	H	L	X	D	Write to current Add., Suspend Burst

- For a write operation preceded by a read cycle, \overline{OE} must be HIGH early enough to allow Input Data Setup, and must be kept HIGH through Input Data Hold Time.
- \overline{WE} refers to WE_A, WE_B.
- \overline{ADSP} is gated by \overline{CS} , and \overline{CS} is used to block \overline{ADSP} when $\overline{CS} = V_{IH}$, as required in applications using Processor Address Pipelining.
- All Addresses, Data In and Control signals are registered on the rising edge of CLK.

Burst Sequence Truth Table

External Address	A15-A2	(A1,A0)				Notes
		(0,0)	(0,1)	(1,0)	(1,1)	
1st Access	A15-A2	(0,0)	(0,1)	(1,0)	(1,1)	
2nd Access	A15-A2	(0,1)	(1,0)	(1,1)	(0,0)	
3rd Access	A15-A2	(1,0)	(1,1)	(0,0)	(0,1)	
4th Access	A15-A2	(1,1)	(0,0)	(0,1)	(1,0)	

Write Enable Truth Table

$\overline{WE_a}$	$\overline{WE_b}$	Byte Written	Notes
H	H	Read All Bytes	
L	L	Write All Bytes	
L	H	Write Byte A (D_{IN} 0 - 8)	
H	L	Write Byte B (D_{IN} 9 - 17)	

Absolute Maximum Ratings

Parameter	Symbol	Rating	Units	Notes
Power Supply Voltage	V_{DD}	-0.5 to 4.6	V	1
Input Voltage	V_{IN}	-0.5 to 6.0	V	1
Output Voltage	V_{OUT}	-0.5 to $V_{DD}+0.5$	V	1
Operating Temperature	T_{OPR}	0 to +70	°C	1
Storage Temperature	T_{STG}	-55 to +125	°C	1
Power Dissipation	P_D	2.0	W	1
Short Circuit Output Current	I_{OUT}	50	mA	1

- Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

Recommended DC Operating Conditions ($T_A=0$ to 70°C)

Parameter	Symbol	Min.	Typ.	Max.	Units	Notes
Supply Voltage	V_{DD}	3.135	3.3	3.465	V	1, 4
Input High Voltage	V_{IH}	2.2	—	5.5	V	1, 2, 4
Input Low Voltage	V_{IL}	-0.3	—	0.8	V	1, 3, 4
Output Current	I_{OUT}	—	5	8	mA	4

- All voltages referenced to V_{SS} . All V_{DD} and V_{SS} pins must be connected.
- $V_{IH}(\text{Max})\text{DC} = 5.5$ V, $V_{IH}(\text{Max})\text{AC} = 6.0$ V (pulse width $\leq 4.0\text{ns}$).
- $V_{IL}(\text{Min})\text{DC} = -0.3$ V, $V_{IL}(\text{Min})\text{AC} = -1.5$ V (pulse width $\leq 4.0\text{ns}$).
- Input Voltage levels are tested to the following DC conditions: 1 microsecond cycle and 200 nanosecond set-up and hold times.

Capacitance ($T_A=0$ to $+70^\circ\text{C}$, $V_{DD}=3.3\text{V} \pm 5\%$, $f=1\text{MHz}$)

Parameter	Symbol	Test Condition	Max	Units	Notes
Input Capacitance	C_{IN}	$V_{IN} = 0\text{V}$	5	pF	
Data I/O Capacitance (DQ0-DQ17)	C_{OUT}	$V_{OUT} = 0\text{V}$	5	pF	

DC Electrical Characteristics ($T_A = 0$ to $+70^\circ\text{C}$, $V_{DD} = 3.3\text{V} \pm 5\%$)

Parameter	Symbol	Min.	Max.	Units	Notes
Operating Current Average Power Supply Operating Current ($\overline{OE} = V_{IH}$, $I_{OUT} = 0$)	I_{DD12}	—	300	mA	2, 3
Standby Current Power Supply Standby Current ($\overline{CS2} = V_{IH}$ or $CS2 = V_{IL}$ or $\overline{CS} = V_{IL}$ All other inputs = V_{IH} or V_{IL} , $I_{OUT} = 0$ Clock @ 83MHz)	I_{SB}	—	25	mA	1, 3
Input Leakage Current Input Leakage Current, any input ($V_{IN} = 0$ & V_{DD})	I_{LI}	—	+1	μA	4
Output Leakage Current ($V_{OUT} = 0$ & V_{DD} , $\overline{OE} = V_{IH}$)	I_{LO}	—	+1	μA	
Output High Level Output "H" Level Voltage ($I_{OH} = -8\text{mA}$ @ 2.4V)	V_{OH}	2.4	—	V	
Output Low Level Output "L" Level Voltage ($I_{OL} = +8\text{mA}$ @ 0.4V)	V_{OL}	—	0.4	V	
1. I_{SB} = Stand-by Current. 2. I_{DD} = Selected Current. 3. I_{OUT} = Chip Output Current. 4. The input leakage current for 5.5V input is 200 μA for Clk, Chip Selects, and Output Enable. Other inputs have 100 μA of leakage current at 5.5V					

AC Test Conditions ($T_A = 0$ to $+70^\circ\text{C}$, $V_{DD} = 3.3\text{V} \pm 5\%$)

Parameter	Symbol	Conditions	Units	Notes
Input Pulse High Level	V_{IH}	3.0	V	
Input Pulse Low Level	V_{IL}	0.0	V	
Input Rise Time	T_R	2.0	ns	
Input Fall Time	T_F	2.0	ns	
Input and Output Timing Reference Level		1.5	V	
Output Load Conditions				1
1. See AC Test Loading figure on page 8.				

AC Characteristics ($T_A=0$ to $+70^\circ\text{C}$, $V_{DD}=3.3\text{V} \pm 5\%$, Units in nsec)

Parameter	Symbol	-8		-9		-10		-11		Notes
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Cycle Time	t_{CYCLE}	12.0	—	12.0	—	12.0	—	12.0	—	
Clock Pulse High	t_{CH}	3.0	—	3.0	—	3.0	—	3.0	—	
Clock Pulse Low	t_{CL}	3.0	—	3.0	—	3.0	—	3.0	—	
Clock to Output Valid	t_{CQ}	—	8.0	—	9.0	—	10.0	—	11.0	3
Address Status Controller Setup Time	t_{ADSCS}	2.5	—	2.5	—	2.5	—	2.5	—	
Address Status Controller Hold Time	t_{ADSCH}	0.5	—	0.5	—	0.5	—	0.5	—	
Address Status Processor Setup Time	t_{ADSPS}	2.5	—	2.5	—	2.5	—	2.5	—	
Address Status Processor Hold Time	t_{ADSPH}	0.5	—	0.5	—	0.5	—	0.5	—	
Advance Setup Time	t_{ADVS}	2.5	—	2.5	—	2.5	—	2.5	—	
Advance Hold Time	t_{ADVH}	0.5	—	0.5	—	0.5	—	0.5	—	
Address Setup Time	t_{AS}	2.5	—	2.5	—	2.5	—	2.5	—	
Address Hold Time	t_{AH}	0.5	—	0.5	—	0.5	—	0.5	—	
Chip Selects Setup Time	t_{CSS}	2.5	—	2.5	—	2.5	—	2.5	—	
Chip Selects Hold Time	t_{CSH}	0.5	—	0.5	—	0.5	—	0.5	—	
Write Enables Setup Time	t_{WES}	2.5	—	2.5	—	2.5	—	2.5	—	
Write Enables Hold Time	t_{WEH}	0.5	—	0.5	—	0.5	—	0.5	—	
Data In Setup Time	t_{DS}	2.5	—	2.5	—	2.5	—	2.5	—	
Data In Hold Time	t_{DH}	0.5	—	0.5	—	0.5	—	0.5	—	
Data Out Hold Time	t_{CQX}	3.0	—	3.0	—	3.0	—	3.0	—	3
Clock High to Output High-Z	t_{CHZ}	—	5.0	—	5.0	—	5.5	—	5.5	1, 2, 4
Clock High to Output Active	t_{CLZ}	2.5	—	2.5	—	2.5	—	2.5	—	1, 2, 4
Output Enable to High-Z	t_{OHZ}	2.0	5.0	2.0	5.5	2.0	6.0	2.0	6.5	1, 4
Output Enable to Low-Z	t_{OLZ}	0.25	—	0.25	—	0.25	—	0.25	—	1, 4
Output Enable to Output Valid	t_{OQ}	—	4.0	—	5.0	—	5.0	—	6.0	3

1. Transitions are measured ± 200 mV from steady state voltage.
 2. At any given voltage and temperature, T_{CHZ} (max) is always less than T_{CTZ} (min) for a given device and from device to device. For any read cycle preceded by a write or deselect cycle, the data bus will transition glitch-free from High-Z to new RAM data.
 3. See AC Test Loading figure 1 on page 8.
 4. See AC Test Loading figure 2 on page 8.

AC Test Loading

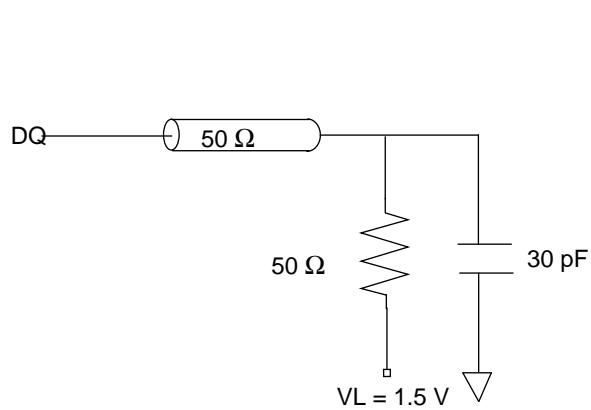


Fig. 1 Test Equivalent Load

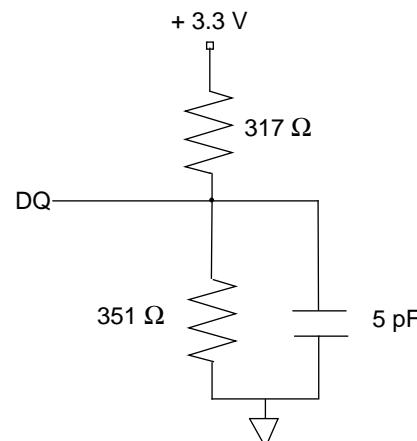
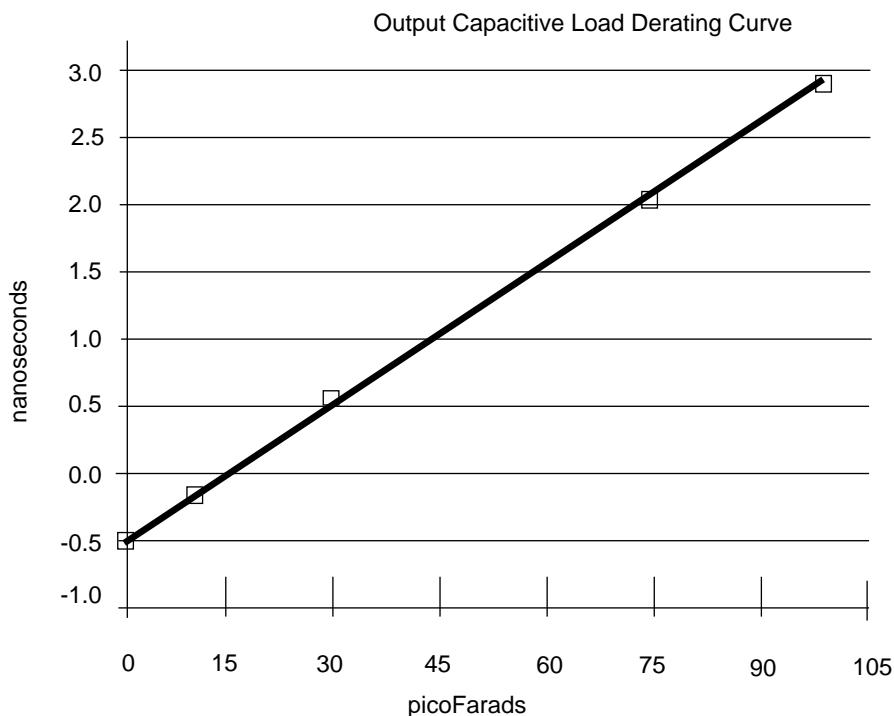
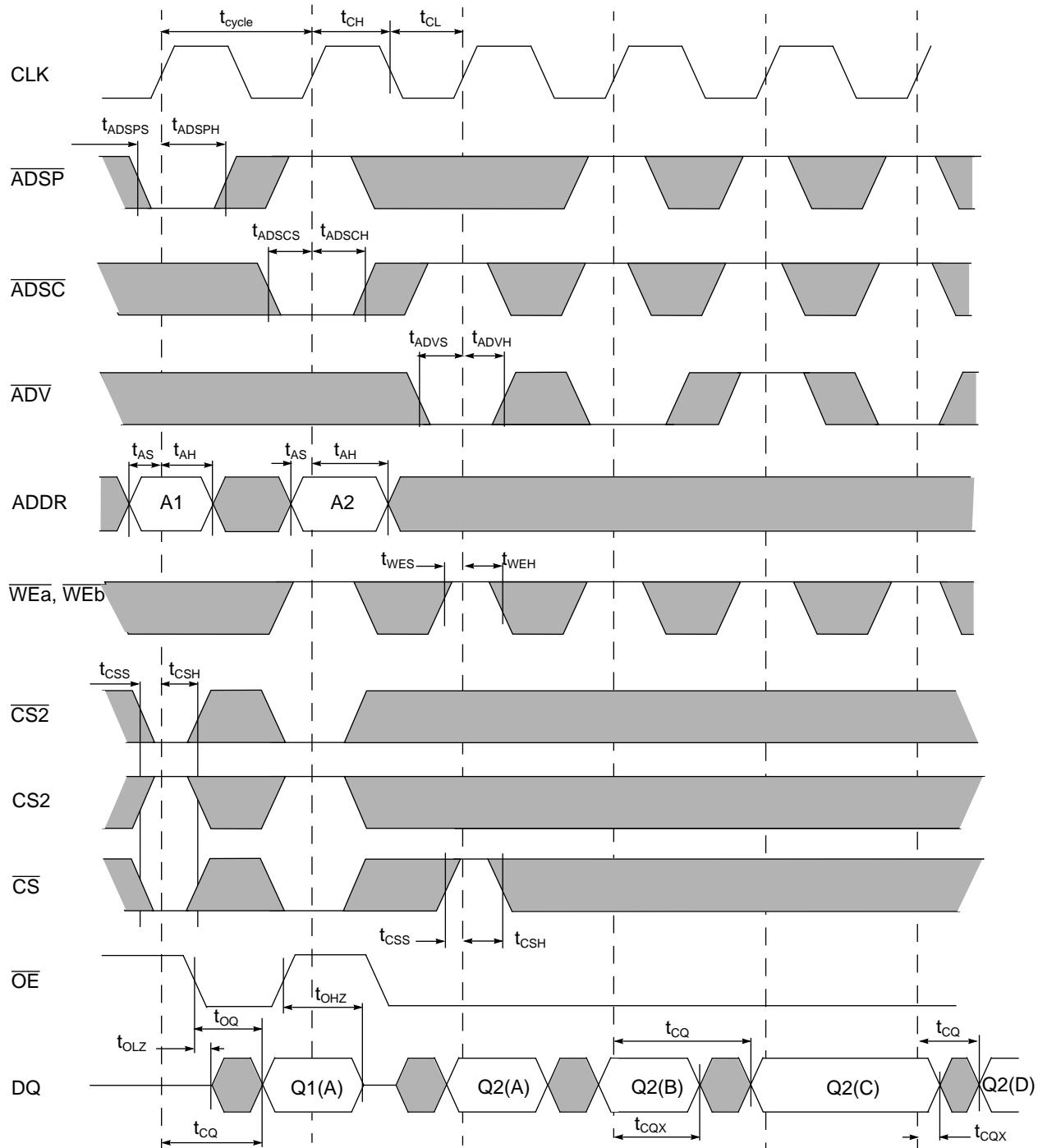


Fig. 2 Test Equivalent Load



The derating curve above is for a purely capacitive load on the output driver. For example, a part specified at 8ns access time will behave as though it has an 8.5 ns access time if a 30 pF load with no DC component was attached to the output driver. The access times guaranteed in the datasheets are based on a 50 ohm terminated test load. For unterminated loads the derating curve should be used. This curve is based on nominal process conditions with worst case parameters $V_{CC} = 3.14$ V, $T_A = 70^\circ$ C.

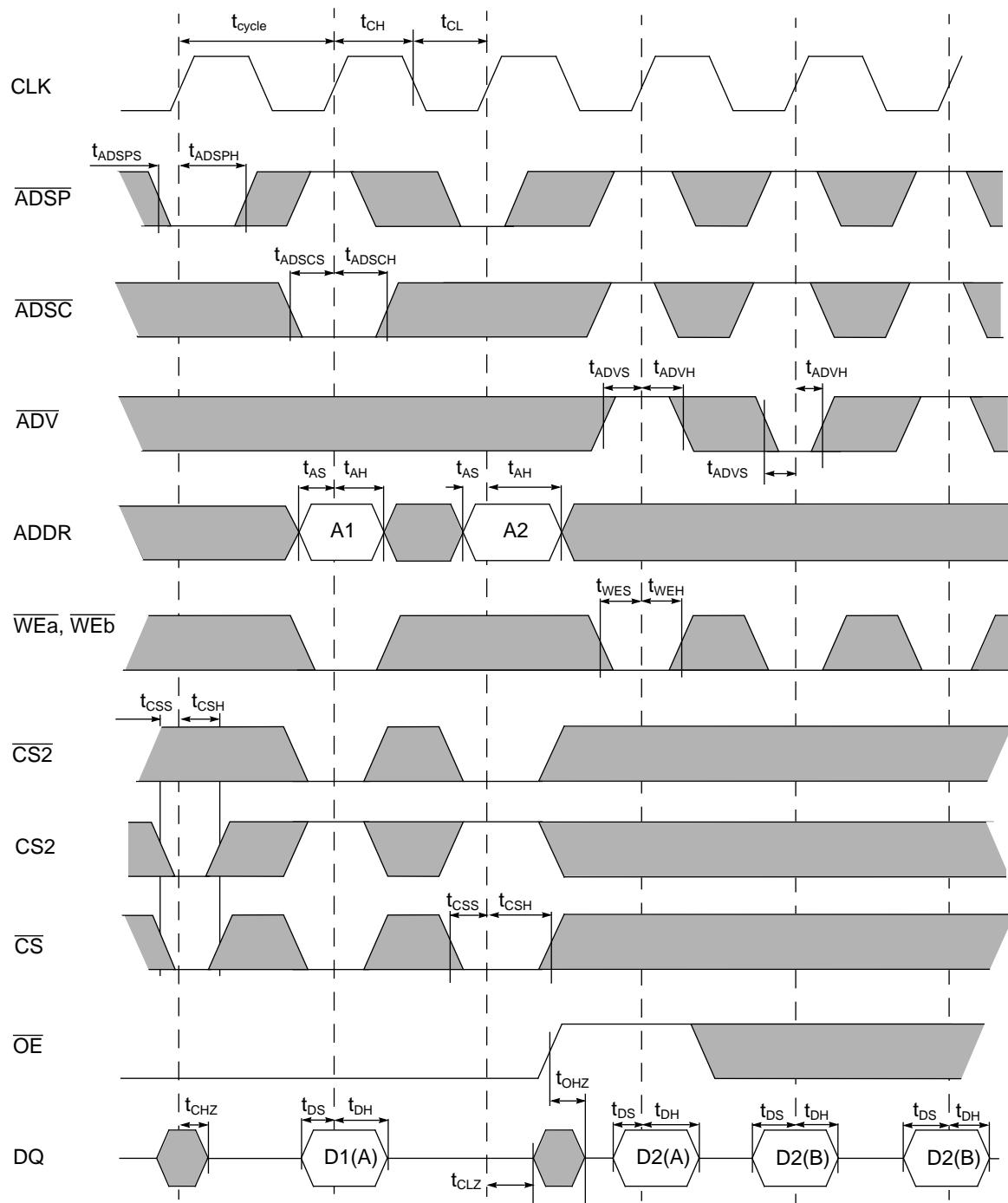
Timing Diagram (Burst Read)



Notes:

1. Q1(A) and Q2(A) refer to data written to address A1 and A2.
2. Q2(B), Q2(C) and Q2(D) refer to data written to subsequent internal burst counter addresses.

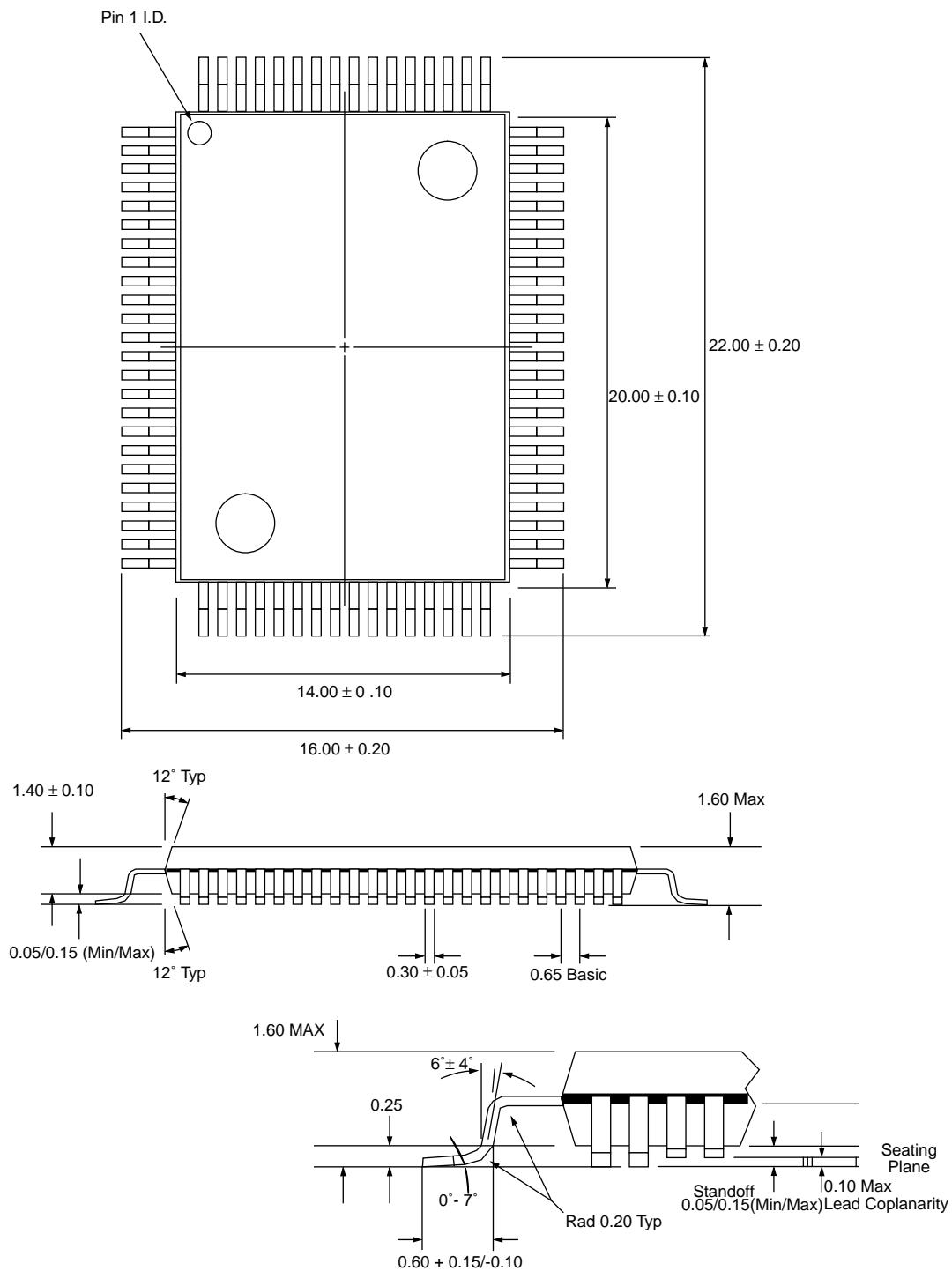
Timing Diagram (Burst Write)



Notes:

1. D1(A) and D2(A) refer to data written to address A1 and A2.
2. D2(B) refers to data written to a subsequent internal burst counter address.
3. \overline{WEa} , \overline{WEb} are don't cares when ADSP is sampled LOW.

100 Pin TQFP Package Diagram



Note: All dimensions in Millimeters

Connect Compatibility for 64K x18 and Future 64K x 16 & 64K x 18

TQFP PIN #	Current Connections (x18)	Future Connections (x16 & x18)	Function
4,27,54,77	NC	V _{DDQ}	Output Power Supply
5,26,55,76	NC	V _{SS}	Ground
14	NC	Low or High, NC for most vendors but Low or High to comply to the JEDEC standard.	\overline{FT} , FFlow thru or Pipeline function, tie Low for flow thru, High for Pipeline
24	DQP2	NC or DQ in x18	Parity bit for second byte
31	NC	Low or High	\overline{LBO} , Linear Burst Order, This pin must be tied low for linear(PowerPC), High for Interleave (Pentium)
64	NC	Low or High. Low allows normal operation.	ZZ , Asynchronous Sleep Mode, Tie to ground for normal function, V _{DDQ} for sleep mode (Low power state)
74	DQP1	NC or DQ in x18	Parity bit for first byte
87	NC	Low or \overline{BWE} , Tie low if function not used	Byte Write Enable, Allows individual bytes to be written.
88	NC	Low or \overline{GW} , Tie High if function not used	Global Write Enable, Allows write of all bytes to occur with single pin.

The IBM041812PQK has the pins connected in the manner indicated in the Current Connections (x18) and is also JEDEC complaint. Future Connections refers to the evolution on the JEDEC standard for subsequent part numbers.

Revision Log

Rev	Contents of Modification
7/95	Initial Release of the 64K x 18 (8/9/10/11) TQFP BURST MODE Application Spec.



© International Business Machines Corp.1996

Printed in the United States of America
All rights reserved

IBM and the IBM logo are registered trademarks of the IBM Corporation.

This document may contain preliminary information and is subject to change by IBM without notice. IBM assumes no responsibility or liability for any use of the information contained herein. Nothing in this document shall operate as an express or implied license or indemnity under the intellectual property rights of IBM or third parties. The products described in this document are not intended for use in implantation or other direct life support applications where malfunction may result in direct physical harm or injury to persons.
NO WARRANTIES OF ANY KIND, INCLUDING, BUT NOT LIMITED TO, THE IMPLIED WARRANTIES OF MERCHANTABILITY OR FITNESS FOR A PARTICULAR PURPOSE, ARE OFFERED IN THIS DOCUMENT.

For more information contact your IBM Microelectronics sales representative or
visit us on World Wide Web at <http://www.chips.ibm.com>



Integrated Device Technology, Inc.

BiCMOS StaticRAM

240K (16K x 15-BIT)

CACHE-TAG RAM

For PowerPC™ and RISC Processors

IDT71216

FEATURES:

- 16K x 15 Configuration
 - 12 TAG Bits
 - 3 Separate I/O Status Bits (Valid, Dirty, Write Through)
- Match output uses Valid bit to qualify MATCH output
- High-Speed Address-to-Match comparison times
 - 8/9/10/12ns over commercial temperature range
- TA circuitry included inside the Cache-Tag for highest speed operation
- Asynchronous Read/Match operation with Synchronous Write and Reset operation
- Separate WE for the TAG bits and the Status bits
- Separate OE for the TAG bits, the Status bits, and TA
- Synchronous RESET pin for invalidation of all Tag entries
- Dual Chip selects for easy depth expansion with no performance degradation
- I/O pins both 5V TTL and 3.3V LVTTL compatible with VccQ pins
- PWRDN pin to place device in low-power mode
- Packaged in a 80-pin Thin Plastic Quad Flat Pack (TQFP)

DESCRIPTION:

The IDT71216 is a 245,760-bit Cache Tag StaticRAM, organized 16K x 15 and designed to support PowerPC and other RISC processors at bus speeds up to 66MHz. There are twelve common I/O TAG bits, with the remaining three bits used as status bits. A 12-bit comparator is on-chip to allow fast comparison of the twelve stored TAG bits and the current Tag input data. An active HIGH MATCH output is generated when these two groups of data are the same for a given address.

This high-speed MATCH signal, with tADM as fast as 8ns, provides the fastest possible enabling of secondary cache accesses.

The three separate I/O status bits (VLD, DTY, and WT) can be configured for either dedicated or generic functionality, depending on the SFUNC input pin. With SFUNC LOW, the status bits are defined and used internally by the device, allowing easier determination of the validity and use of the given Tag data. SFUNC HIGH releases the defined internal status bit usage and control, allowing the user to configure the status bit information to fit his system needs. A synchronous RESET pin, when held LOW at a rising clock edge, will reset all status bits in the array for easy invalidation of all Tag addresses.

The IDT71216 also provides the option for Transfer Acknowledge (TA) generation within the cache tag itself, based upon MATCH, VLD bit, WT bit, and external inputs provided by the user. This can significantly simplify cache controller logic and minimize cache decision time. Match and Read operations are both asynchronous in order to provide the fastest access times possible, while Write operations are synchronous for ease of system timing.

The IDT71216 uses a 5V power supply on Vcc, with separate VccQ pins provided for the outputs to offer compliance with both 5.0V TTL and 3.3V LVTTL Logic levels. The PWRDN pin offers a low-power standby mode to reduce power consumption by 90%, providing significant system power savings.

The IDT71216 is fabricated using IDT's high-performance, high-reliability BiCMOS technology and is offered in a space-saving 80-pin Thin Plastic Quad Flat Pack (TQFP) package.

PIN DESCRIPTIONS

A0 – A13	Address Inputs	Input
CS1, CS2	Chip Selects	Input
WET	Write Enable - Tag Bits	Input
WES	Write Enable - Status Bits	Input
OET	Output Enable - Tag Bits	Input
OES	Output Enable - Status Bits	Input
RESET	Status Bit Reset	Input
PWRDN	Powerdown Mode Control Pin	Input
SFUNC	Status Bit Function Control Pin	Input
TT1	Read/Write Input from Processor	Input
VLDIN / S1IN	Valid Bit / S1 Bit Input	Input
DTYIN / S2IN	Dirty Bit / S2 Bit Input	Input
WTIN / S3IN	Write Through Bit / S3 Bit Input	Input

CLK	System Clock	Input
TAH	TA Force High	Input
TAOE	TA Output Enable	Input
TAIN	Additional TA Input	Input
TA	Transfer Acknowledge	Output
TAG0 – TAG11	Tag Data Input/Outputs	I/O
VLDOUT / S1OUT	Valid Bit / S1 Bit Output	Output
DTYOUT / S2OUT	Dirty Bit / S2 Bit Output	Output
WTOUT / S3OUT	Write Through Bit / S3 Bit Output	Output
MATCH	Match	Output
Vcc	+5V Power	Pwr
VccQ	Output Buffer Power	QPwr
Vss	Ground	Gnd

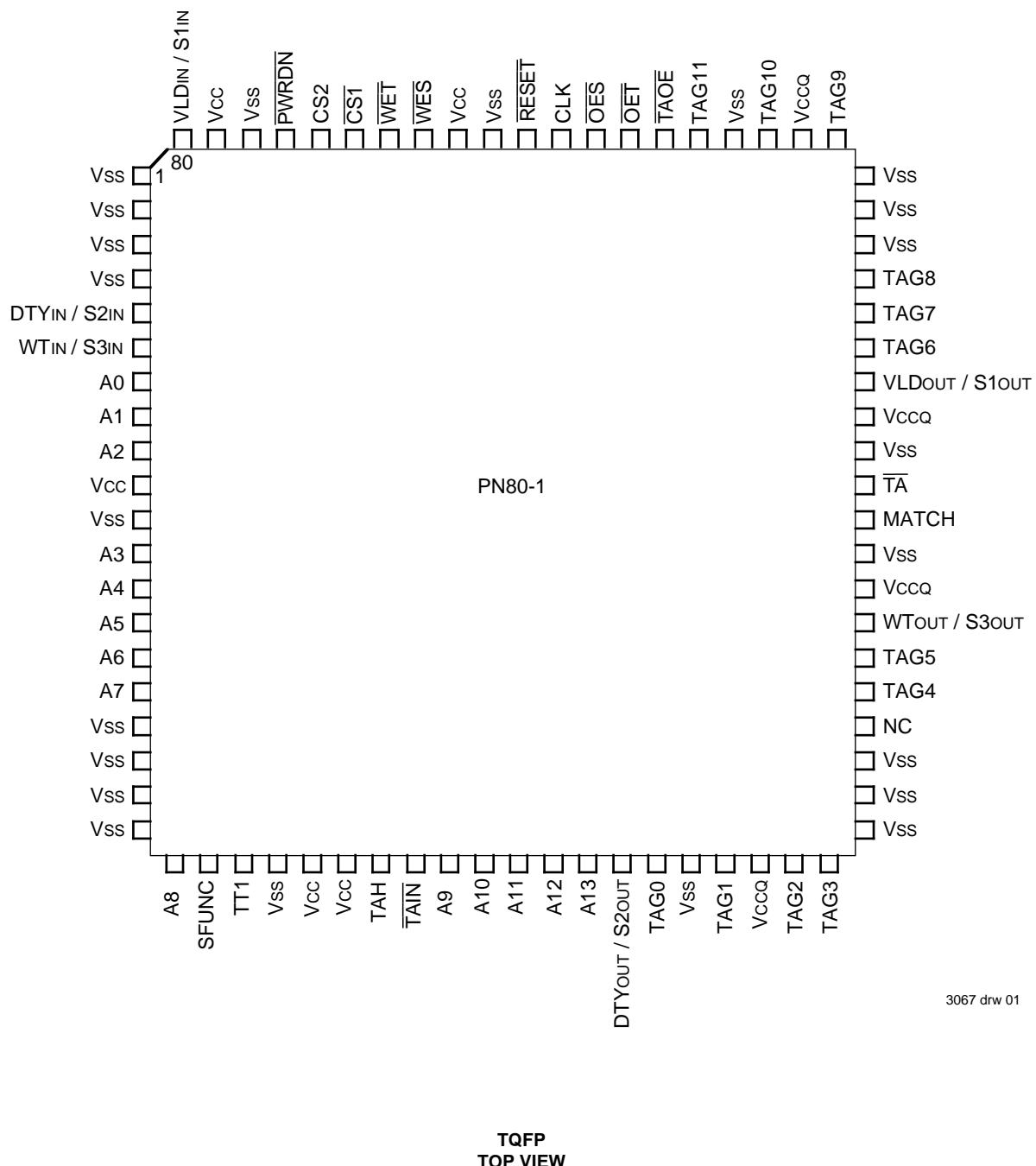
3067 tbl 01

The IDT logo is a registered trademark and CacheRAM is a trademark of Integrated Device Technology, Inc.
PowerPC is a trademark of International Business Machines, Inc.

COMMERCIAL TEMPERATURE RANGE

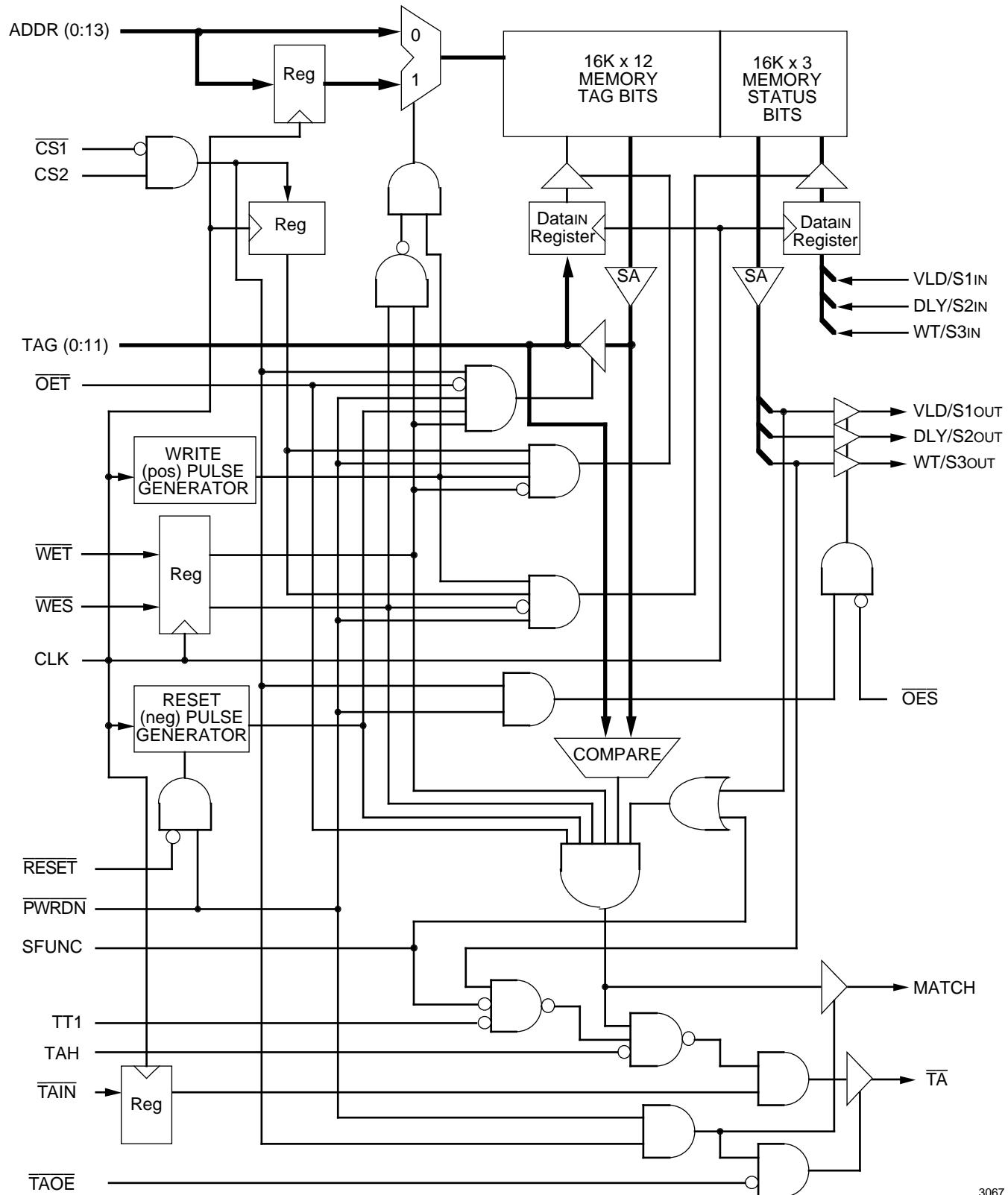
AUGUST 1996

PIN CONFIGURATION



3067 drw 01

FUNCTIONAL BLOCK DIAGRAM



3067 drw 02

TRUTH TABLES

CHIP SELECT, RESET, AND POWER-DOWN FUNCTIONS^(1, 2)

CS1	CS2	RESET	PWRDN	CLK	WET	WES	TAOE	TAG	VLDOUT	DTYOUT	WTOUT	MATCH	TA	OPERATION	POWER
-----	-----	-------	-------	-----	-----	-----	------	-----	--------	--------	-------	-------	----	-----------	-------

CHIP SELECT FUNCTION

H	X	X	H	X	X	X	X	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Deselected	Active
X	L	X	H	X	X	X	X	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Deselected	Active
L	H	X	H	X	X	X	X	-	-	-	-	-	-	Selected	Active

RESET FUNCTION

L	H	L	H	↑	H	H	L	Hi-Z	L ⁽³⁾	L ⁽³⁾	L ⁽³⁾	L ⁽³⁾	H	Reset Status	Active
L	H	L	H	↑	H	H	H	Hi-Z	L ⁽³⁾	L ⁽³⁾	L ⁽³⁾	L ⁽³⁾	Hi-Z	Reset Status	Active
H	X	L	H	↑	H	H	X	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Reset Status	Active
X	L	L	H	↑	H	H	X	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Reset Status	Active
X	X	L	H	↑	L	X	X	-	-	-	-	-	-	Not Allowed	-
X	X	L	H	↑	X	L	X	-	-	-	-	-	-	Not Allowed	-

POWER-DOWN FUNCTION

X	X	X	L	X	H	H	X	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Power-down	Standby
---	---	---	---	---	---	---	---	------	------	------	------	------	------	------------	---------

NOTES:

3067 tbl 02

- "H" = VI_H, "L" = VI_L, "X" = don't care, "—" = unrelated.
- OET, OES, TT1, TAH, TAIN and SFUNC are "X" for this table.
- OES is LOW.

READ AND WRITE FUNCTIONS^(1, 2)

OET	OES	WET	WES	CLK	TT1	TAG	VLDIN	DTYIN	WTIN	VLDOUT	DTYOUT	WTOUT	MATCH	OPERATION
-----	-----	-----	-----	-----	-----	-----	-------	-------	------	--------	--------	-------	-------	-----------

READ FUNCTION

L	X	H	X	X	X	DOUT	-	-	-	-	-	-	DOUT	Read TAG I/O
X	L	X	X	X	X	-	-	-	-	DOUT	DOUT	DOUT	DOUT	Read Status Bits
H	X	X	X	X	X	Hi-Z	-	-	-	-	-	-	DOUT	TAG I/O Disable
X	H	X	X	X	X	-	-	-	-	Hi-Z	Hi-Z	Hi-Z	DOUT	Status Disabled

WRITE FUNCTION

H	X	L	X	↑	X	DIN	-	-	-	DOUT	DOUT	DOUT	L	Write TAG I/O
L	X	L	X	↑	X	-	-	-	-	-	-	-	-	Not Allowed
X	L	X	L	↑	X	-	DIN	DIN	DIN	DOUT ⁽³⁾	DOUT ⁽³⁾	DOUT ⁽³⁾	L	Write Status Bits
X	H	X	L	↑	X	-	DIN	DIN	DIN	Hi-Z	Hi-Z	Hi-Z	L	Write Status Bits

NOTES:

3067 tbl 03

- "H" = VI_H, "L" = VI_L, "X" = don't care, "—" = unrelated.
- This table applies when CS1 is LOW and CS2, RESET, and PWRDN are HIGH. TAOE, TAH, TAIN and SFUNC are "X" for this table.
- DOUT in this case is the same as DIN; that is, the input data is written through to the outputs during the write operation.

TRUTH TABLES (CONT.)

MATCH FUNCTION^(1, 2, 3)

CS1	CS2	SFUNC	OET	WET	WES	TAG	VLD ⁽⁴⁾	DTY ⁽⁴⁾	WT ⁽⁴⁾	MATCH	OPERATION
H	X	X	X	X	X	Hi-Z	-	-	-	Hi-Z	Deselected
X	L	X	X	X	X	Hi-Z	-	-	-	Hi-Z	Deselected
L	H	X	X	X	X	-	-	-	-	DOUT	Selected
L	H	X	L	H	X	DOUT	-	-	-	L	Read Tag I/O
L	H	X	H	L	X	DIN	-	-	-	L	Write Tag I/O
L	H	X	X	X	L	-	DIN	DIN	DIN	L	Write Status Bits
L	H	L	H	H	H	TAGIN	L	-	-	L	Invalid Data - Dedicated Status Bits
L	H	L	H	H	H	TAGIN	H	-	-	M	Match - Dedicated Status Bits
L	H	H	H	H	H	TAGIN	X	-	-	M	Match - Generic Status Bits

NOTES:

3067 tbl 04

- "H" = VIH, "L" = VIL, "X" = don't care, "-" = unrelated.
- M = HIGH if TAGIN equals the memory contents at that address; M = LOW if TAGIN does not equal the memory contents at that address.
- PWRDN and RESET are HIGH for this table. TT1, TAH, TAOE, TAIN, OES, and CLK are "X".
- This column represents the stored memory cell data for the given Status bit at the selected address.

TA FUNCTION^(1, 2, 3, 5)

TAOE	TAIN ⁽⁶⁾	OET	WET	WES	TAH	TT1	SFUNC	VLD ⁽⁴⁾	DTY ⁽⁴⁾	WT ⁽⁴⁾	TAG	MATCH	TA	OPERATION	
H	X	X	X	X	X	X	X	X	-	X	-	-	-	Hi-Z	TA Disabled
L	L	X	X	X	X	X	X	X	-	X	-	X	L	External TA Input ⁽⁷⁾	
L	H	L	X	X	X	X	X	X	-	X	DOUT	L	H	Read TAG	
L	H	X	L	X	X	X	X	X	-	X	DIN	L	H	Write TAG	
L	H	X	X	L	X	X	X	DIN	DIN	DIN	-	L	H	Write Status	
L	H	X	X	X	H	X	X	X	-	X	-	X	H	Force TA HIGH	
L	H	X	X	X	X	X	L	L	-	X	-	L	H	Invalid TAG	
L	H	X	X	X	X	L	L	X	-	H	-	X	H	Write Through	
L	H	H	H	H	L	X	L	H	-	L	TAGIN	M	~M	Compare	
L	H	H	H	H	L	H	L	H	-	X	TAGIN	M	~M	Compare	
L	H	H	H	H	L	X	L	H	-	X	TAGIN	M	~M	Compare	
L	H	H	H	H	L	X	H	X	-	X	TAGIN	M	~M	Compare	

NOTES:

3067 tbl 05

- "H" = VIH, "L" = VIL, "X" = don't care, "-" = unrelated.
- M = HIGH if TAGIN equals the memory contents at that address; M = LOW if TAGIN does not equal the memory contents at that address.
- PWRDN and RESET are HIGH for this table. CLK and OES are "X".
- This column represents the stored memory cell data for the given Status bit at the selected address.
- CS1 is LOW, CS2 is HIGH for this table.
- TAIN is a synchronous input; thus the inputs noted in the table must be applied during a rising CLK edge.
- TA will be a factor in determining the TA output in all cases except when TAH is HIGH and there is a valid MATCH. In that case, TA will be LOW(Valid).

RECOMMENDED DC OPERATING CONDITIONS

Symbol	Parameter	Min.	Typ.	Max.	Unit
Vcc	Supply Voltage	4.75	5.0	5.25	V
VCCQ	5V Output Buffers	4.75	5.0	5.25	V
VCCQ	3.3V Output Buffers	3.0	3.3	3.6	V
Vss	Supply Ground	0	0	0	V
VIH	Input High Voltage	2.2	3.0	VCC+0.3	V
VIHQ	I/O High Voltage	2.2	3.0	VCCQ+0.3	V
VIL	Input Low Voltage	-0.5 ⁽¹⁾	—	0.8	V

NOTE: 3067 tbl 06

1. VIL (min.) = -1.5V for pulse width of less than 10ns, once per cycle.

CAPACITANCE

(TA = +25°C, f = 1.0 MHz)

Symbol	Parameter ⁽¹⁾	Condition	Max.	Unit
CIN	Input Capacitance	VIN = 0V	5	pF
CTAG	TAG Input/Output Capacitance	VI/O = 0V	7	pF
COUT	Output Capacitance	VOUT = 0V	7	pF

NOTE: 3067 tbl 07

1. This parameter is determined by device characterization but is not production tested.

DC ELECTRICAL CHARACTERISTICS OVER THE OPERATING TEMPERATURE AND SUPPLY VOLTAGE RANGE

(VCC = 5.0V ± 5%, VCCQ = 5.0V ± 5% OR 3.3V ± 0.3V)

Symbol	Parameter	Test Condition	Min.		Max.		Unit
			Com'l.	Mil.	Com'l.	Mil.	
I _L	Input Leakage Current	VCC = Max., VIN = 0V to VCC	—	—	5	—	µA
I _{LO}	Output Leakage Current	CS1 ≥ VIH, CS2 ≤ VIL, OE ≥ VIH, VCC = Max. VOUT = 0V to VCCQ, VCCQ = Max.	—	—	5	—	µA
V _{OL}	Output Low Voltage	I _{OL} = 4mA, VCC = Min.	—	—	0.4	—	V
V _{OH}	Output High Voltage	I _{OH} = -4mA, VCC = Min.	2.4	—	—	—	V

3067 tbl 09

DC ELECTRICAL CHARACTERISTICS OVER THE OPERATING TEMPERATURE AND SUPPLY VOLTAGE RANGE^(1, 2) (VCC = 5.0V ± 5%)

Symbol	Parameter	Test Condition	71216S8		71216S9		71216S10		71216S12		Unit
			Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	
I _{CC}	Operating Power Supply Current	PWRDN ≥ VIH Outputs Open, VCC = Max., f = f _{MAX} ⁽³⁾	330	—	300	—	290	—	280	—	mA
I _{SB}	Standby Power Supply Current	PWRDN ≤ VIL, VIN ≥ VIH or ≤ VIL VCC = Max., f = f _{MAX} ⁽³⁾	30	—	30	—	30	—	30	—	mA
I _{SB1}	Full Standby Power Supply Current	PWRDN ≤ VIL, VIN ≥ VHC or ≤ VLC ⁽⁴⁾ VCC = Max., f = 0 ⁽³⁾	25	—	25	—	25	—	25	—	mA

NOTES:

3067 tbl 10

1. All values are maximum guaranteed values.

2. CS1 ≤ VIL, CS2 ≥ VIH.

3. f_{MAX} = 1/t_{CYC} (all address inputs are cycling at f_{MAX}). f = 0 means no address input lines are changing.

4. V_{HC} = VCC - 0.2V, VLC = 0.2V

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

Symbol	Rating	Value	Unit
VTERM	Terminal Voltage with Respect to GND	-0.5 to +7.0 ⁽²⁾	V
TA	Operating Temperature	-0 to +70	°C
TBIAS	Temperature Under Bias	-65 to +135	°C
TSTG	Storage Temperature	-65 to +150	°C
PT	Power Dissipation	1.7	W
I _{OUT}	DC Output Current	20	mA

NOTES:

3067 tbl 08

1. Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

2. VIN should not exceed VCC+0.5V. All pins should not exceed 7.0V. V_{CCA} should never exceed V_{CC}, and V_{CC} should never exceed V_{CCQ} + 4.0V.

AC ELECTRICAL CHARACTERISTICS

(V_{CC} = 5.0V ± 5%, V_{CCQ} = 5.0V ± 5% OR 3.3V ± 0.3V, TA = 0 to 70°C)

Symbol	Parameter	IDT71216S8		IDT71216S9		IDT71216S10		IDT71216S12		Unit
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Read Cycle										
taAT	Address Access Time Tag Bits	—	10	—	11	—	12	—	14	ns
tACST	Chip Select Access Time Tag Bits	—	8	—	9	—	10	—	12	ns
tCLZ ⁽¹⁾	Chip Select to Tag and Status Bits in Low-Z	1	—	1	—	1	—	1	—	ns
tCHZ ⁽¹⁾	Chip Select to Tag and Status Bits in High-Z	1	5	1	6	1	6	1	7	ns
toET	Output Enable to Tag Bits Valid	—	5	—	6	—	6	—	7	ns
totLZ ⁽¹⁾	Output Enable to Tag Bits in Low-Z	0	—	0	—	0	—	0	—	ns
totHZ ⁽¹⁾	Output Enable to Tag Bits in High-Z	1	5	1	6	1	6	1	7	ns
tTOH	Tag Bit Hold from Address Change	2	—	2	—	2	—	2	—	ns
toES	Output Enable to Status Bits Valid	—	5	—	6	—	6	—	7	ns
tosLZ ⁽¹⁾	Output Enable to Status Bits in Low-Z	0	—	0	—	0	—	0	—	ns
tosHZ ⁽¹⁾	Output Enable to Status Bits in High-Z	1	5	1	6	1	6	1	7	ns
taAS	Address Access Time Status Bits	—	8	—	9	—	10	—	12	ns
tACSS	Chip Select Access Time Status Bits	—	6	—	7	—	8	—	10	ns
tSOH	Status Bit Hold from Address Change	2	—	2	—	2	—	2	—	ns

NOTE:

3067 tbl 11

1. This parameter is guaranteed with the AC Load (Figure 3) by device characterization, but is not production tested.

AC ELECTRICAL CHARACTERISTICS (1)

(V_{CC} = 5.0V ± 5%, V_{CCQ} = 5.0V ± 5% OR 3.3V ± 0.3V, TA = 0 to 70°C)

Symbol	Parameter	IDT71216S8		IDT71216S9		IDT71216S10		IDT71216S12		Unit
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Reset and Power Down Cycles										
tsR	RESET Set-up Time	4	—	4	—	4	—	4	—	ns
tHR	RESET Hold Time	1	—	1	—	1	—	1	—	ns
tsRST	Status Bit Reset Time	—	50	—	60	—	60	—	70	ns
tSHRS	Status Bit Hold from RESET LOW	2	—	2	—	2	—	2	—	ns
trsMI	RESET LOW to MATCH and $\overline{T_A}$ Invalid	—	9	—	10	—	10	—	12	ns
trsMV	RESET HIGH to MATCH and $\overline{T_A}$ Valid	—	110	—	120	—	120	—	130	ns
trSHZ ⁽²⁾	RESET LOW to TAG High-Z	—	9	—	10	—	10	—	12	ns
trSLZ ⁽²⁾	RESET HIGH to TAG Low-Z	—	90	—	100	—	100	—	110	ns
tpDSR	PWRDN Set-up to RESET LOW	30	—	30	—	30	—	30	—	ns
trHPL	RESET HIGH to PWRDN LOW	1	—	1	—	1	—	1	—	CLK
trHWL	RESET HIGH to \overline{WET} and \overline{WES} LOW	90	—	95	—	95	—	105	—	ns
tpD ⁽²⁾	PWRDN LOW to Low Power Mode	—	50	—	50	—	50	—	50	ns
tpU ⁽²⁾	PWRDN HIGH to Active Power Mode	0	—	0	—	0	—	0	—	ns
tpDHZ ⁽²⁾	PWRDN LOW to Outputs in High-Z	—	9	—	10	—	10	—	12	ns
tpDLZ ⁽²⁾	PWRDN HIGH to Outputs in Low-Z	0	—	0	—	0	—	0	—	ns
tpUV	PWRDN HIGH to Outputs Valid	—	50	—	50	—	50	—	50	ns
tWHPL ⁽²⁾	\overline{WET} and \overline{WES} HIGH to PWRDN LOW	5	—	5	—	5	—	5	—	ns
tPUWL	PWRDN HIGH to \overline{WET} and \overline{WES} Active	50	—	50	—	50	—	50	—	ns

NOTES:

3067 tbl 12

1. Power-down mode is intended to be used during extended time periods of device inactivity.
2. This parameter is guaranteed with the AC Load (Figure 3) by device characterization, but is not production tested.

AC ELECTRICAL CHARACTERISTICS (1)

(V_{CC} = 5.0V ± 5%, V_{CCQ} = 5.0V ± 5% OR 3.3V ± 0.3V, T_A = 0 to 70°C)

Symbol	Parameter	IDT71216S8		IDT71216S9		IDT71216S10		IDT71216S12		Unit
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Write Cycle and Clock Parameters										
t _{CYC}	Clock Cycle Time	15	—	15	—	15	—	16.6	—	ns
t _{CH} ^(2, 3)	Clock Pulse HIGH	4.5	—	4.5	—	4.5	—	5	—	ns
t _{CL} ^(2, 3)	Clock Pulse LOW	4.5	—	4.5	—	4.5	—	5	—	ns
t _S	WET, WES, Chip Select, and Input Data Set-up Time	3	—	3	—	3	—	3	—	ns
t _H	WET, WES, Chip Select, and Input Data Hold Time	1	—	1	—	1	—	1	—	ns
t _{SA}	Address Set-up Time	3	—	3	—	3	—	3	—	ns
t _{HA}	Address Hold Time	1	—	1	—	1	—	1	—	ns
t _{WMI}	CLK HIGH Write to MATCH and TA Invalid	—	6	—	7	—	7	—	8	ns
t _{CKLZ} ⁽³⁾	CLK HIGH Read to Outputs in Low-Z	1.5	—	1.5	—	1.5	—	1.5	—	ns
t _{CTV} ⁽⁴⁾	CLK HIGH Read to Tag Bits Valid	—	9	—	10	—	10	—	12	ns
t _{CSV} ⁽⁴⁾	CLK HIGH Write to Status Outputs Valid	—	8	—	9	—	9	—	10	ns
t _{CSH} ⁽³⁾	Status Output Hold from CLK HIGH Write	0	—	0	—	0	—	0	—	ns
t _{WHPL}	WET and WES HIGH to PWRDN LOW	5	—	5	—	5	—	5	—	ns
t _{PWUL}	PWRDN HIGH to WET and WES Active	50	—	50	—	50	—	50	—	ns

NOTES:

1. All Write cycles are synchronous and referenced from rising CLK.
2. This parameter is measured as a HIGH time above 2.0V and a LOW time below 0.8V.
3. This parameter is guaranteed with the AC Load (Figure 3) by device characterization, but is not production tested.
4. Addresses are stable prior to CLK transition HIGH.

3067 tbl 14

AC ELECTRICAL CHARACTERISTICS

(V_{CC} = 5.0V ± 5%, V_{CCQ} = 5.0V ± 5% OR 3.3V ± 0.3V, T_A = 0 to 70°C)

Symbol	Parameter	IDT71216S8		IDT71216S9		IDT71216S10		IDT71216S12		Unit
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
MATCH and TA Cycles										
tADM	Address to MATCH Valid	—	8	—	9	—	10	—	12	ns
tdAM	Data Input to MATCH Valid	—	8	—	9	—	10	—	12	ns
tCSM	Chip Select to MATCH Valid	—	8	—	9	—	10	—	12	ns
tCMLZ ⁽¹⁾	Chip Select to MATCH in Low-Z	1	—	1	—	1	—	1	—	ns
tCMHZ ⁽¹⁾	Chip Select to MATCH in High-Z	1	5	1	6	1	6	1	7	ns
tMHA	MATCH Valid Hold from Address	2	—	2	—	2	—	2	—	ns
tMHD	MATCH Valid Hold from Data	2	—	2	—	2	—	2	—	ns
tbHA	TA Valid Hold from Address	2	—	2	—	2	—	2	—	ns
tbHD	TA Valid Hold from Data	2	—	2	—	2	—	2	—	ns
tADB	Address to TA Valid	—	9	—	10	—	11	—	13	ns
tDAB	Data Input to TA Valid	—	9	—	10	—	11	—	13	ns
tCSB	Chip Select LOW to TA Valid	—	9	—	10	—	11	—	13	ns
toEBV	TAOE LOW to TA Valid	—	6	—	6	—	7	—	8	ns
toBLZ ⁽¹⁾	TAOE LOW to TA in Low-Z	0	—	0	—	0	—	0	—	ns
toBHZ ⁽¹⁾	TAOE HIGH to TA in High-Z	1	5	1	6	1	6	1	7	ns
tBYFH	TAH HIGH to Force TA HIGH	—	5	—	5	—	5	—	6	ns
tBYHV	TAH LOW to TA Valid	—	5	—	5	—	5	—	6	ns
tsB	TAIN Set-up Time	4	—	4	—	4	—	4	—	ns
tHB	TAIN Hold Time	1.5	—	1.5	—	1.5	—	1.5	—	ns
tBIBL	CLK HIGH TAIN LOW to TA LOW	—	6	—	6	—	7	—	8	ns
tBIBV	CLK HIGH TAIN HIGH to TA Valid	—	6	—	6	—	7	—	8	ns
toEMI	OET LOW to MATCH and TA Invalid	—	6	—	7	—	7	—	8	ns
toEMV	OET HIGH to MATCH and TA Valid	—	7	—	8	—	8	—	10	ns
tWRBH ⁽²⁾	W/R HIGH to TA HIGH	—	6	—	7	—	7	—	8	ns
tWRBV ⁽²⁾	W/R LOW to TA Valid	—	6	—	7	—	7	—	8	ns
twMI	CLK HIGH Write to MATCH and TA Invalid	—	7	—	7	—	7	—	8	ns
tWMV ⁽³⁾	CLK HIGH Read to MATCH and TA Valid	—	8	—	9	—	10	—	12	ns

NOTES:

1. This parameter is guaranteed with the AC Load (Figure 3) by device characterization, but is not production tested.
2. These parameters only apply when SFUNC is LOW and the internal WT bit is HIGH.
3. tADM, tdAM, tCSM and tADB, tDAB, tCSB must also be satisfied.

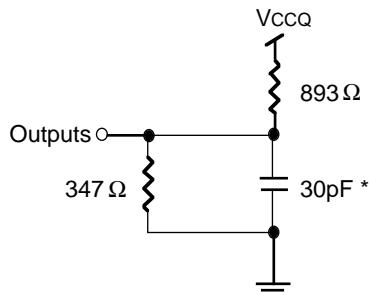
3067 tbl 15

AC TEST CONDITIONS

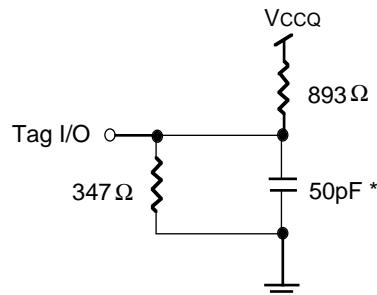
Input Pulse Levels	GND to 3.0V
Input Rise/Fall Times	3ns
Input Timing Reference Levels	1.5V
Output Timing Reference Levels	1.5V
AC Test Load	See Figs. 1, 2, 3, & 4

3067 tbl 16

AC TEST LOADS



3067 drw 03

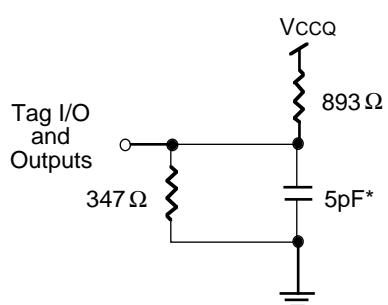


3067 drw 04

Figure 1. AC Test Load

Figure 2. Tag I/O AC Test Load

* Including scope and jig capacitance



3067 drw 05

Figure 3. AC Test Load
(for tHZ and tLZ parameters)

* Including scope and jig capacitance

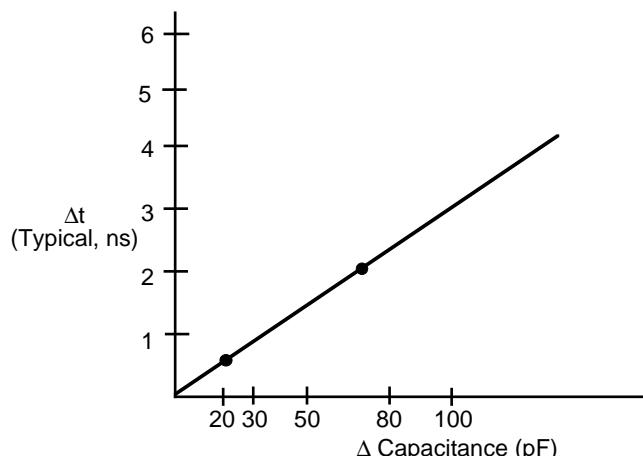
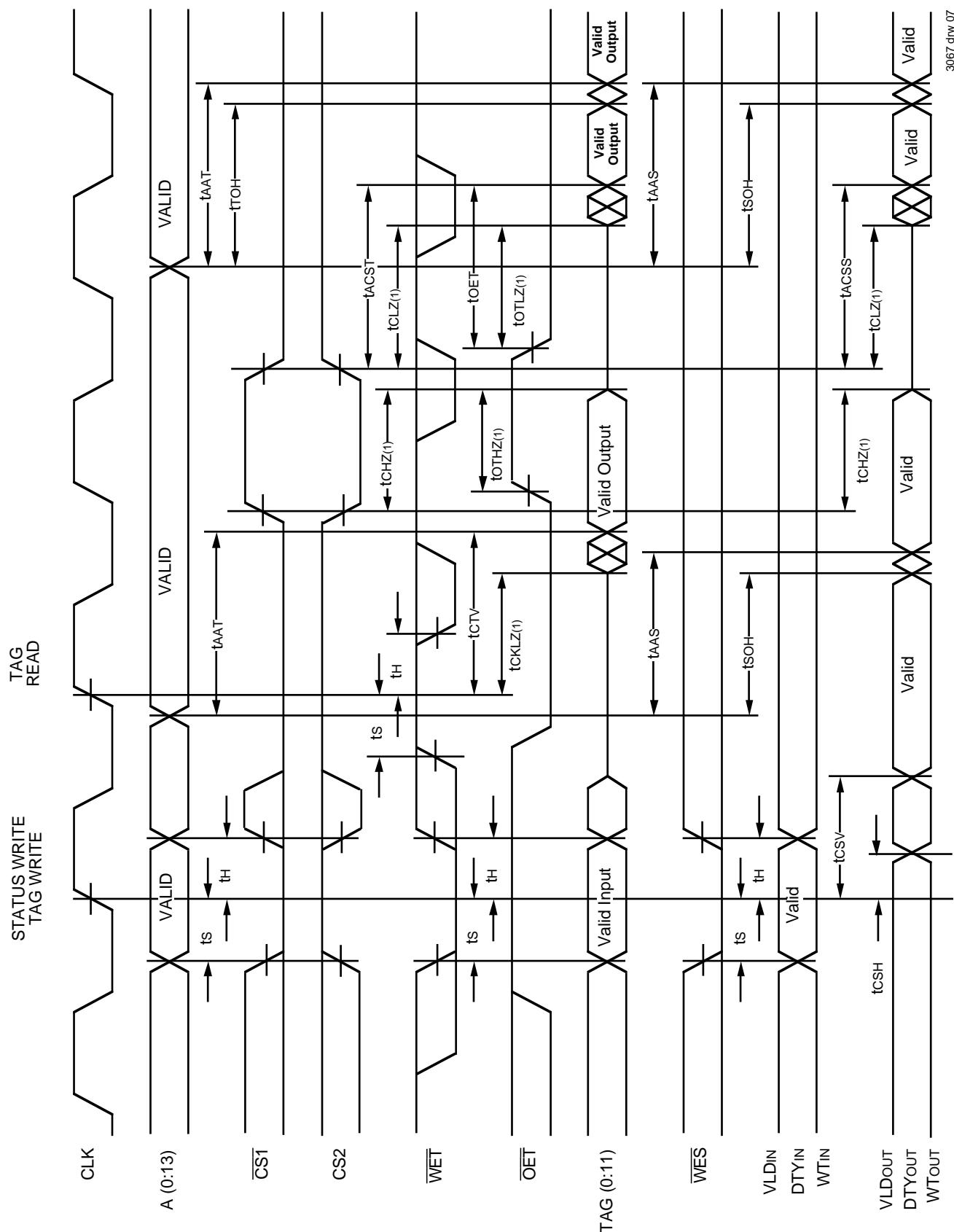


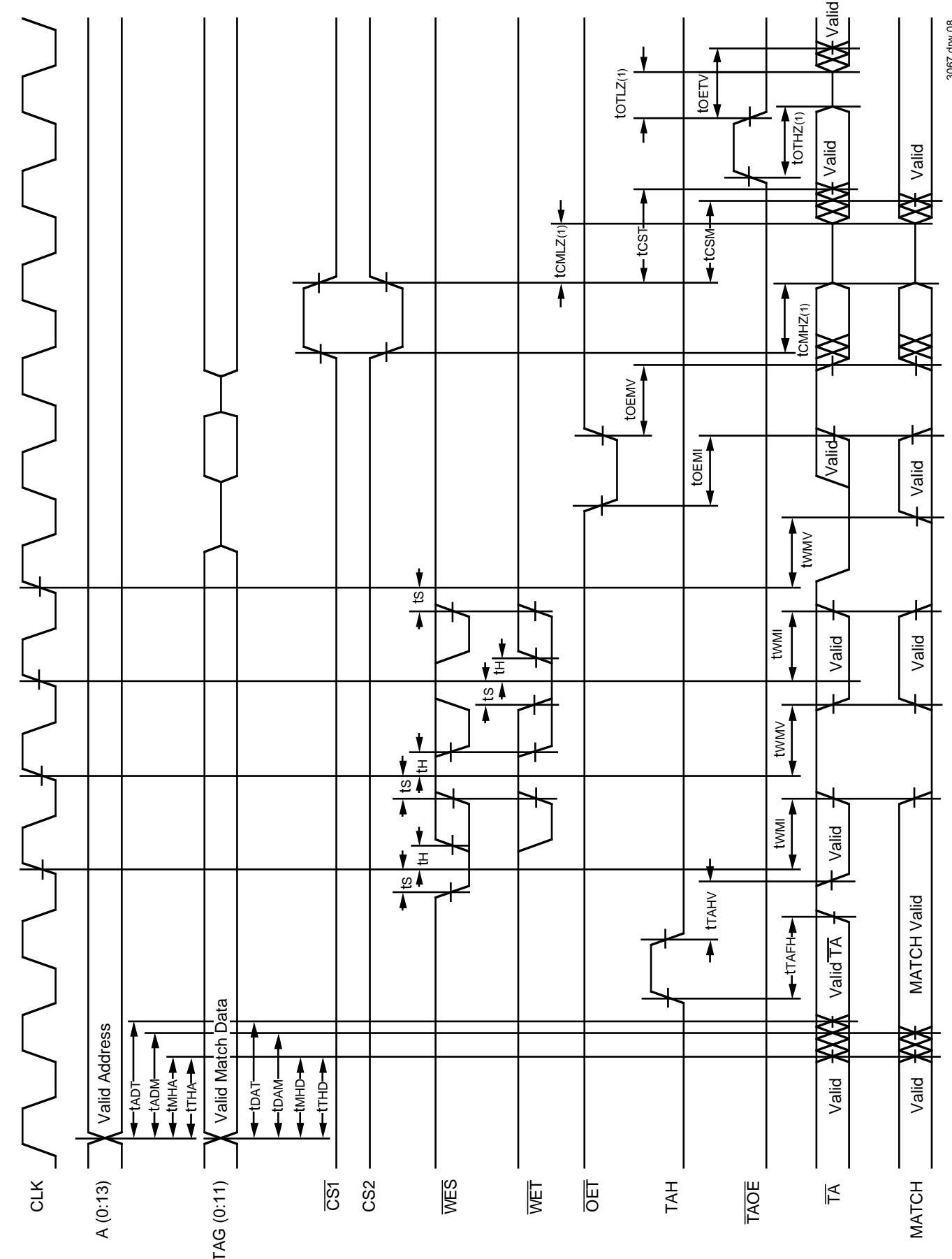
Figure 4. Lumped Capacitance Load, Typical Derating

TIMING WAVEFORMS OF WRITE AND READ CYCLES



NOTE:
1. Transition is measured $\pm 200\text{mV}$ from steady state.

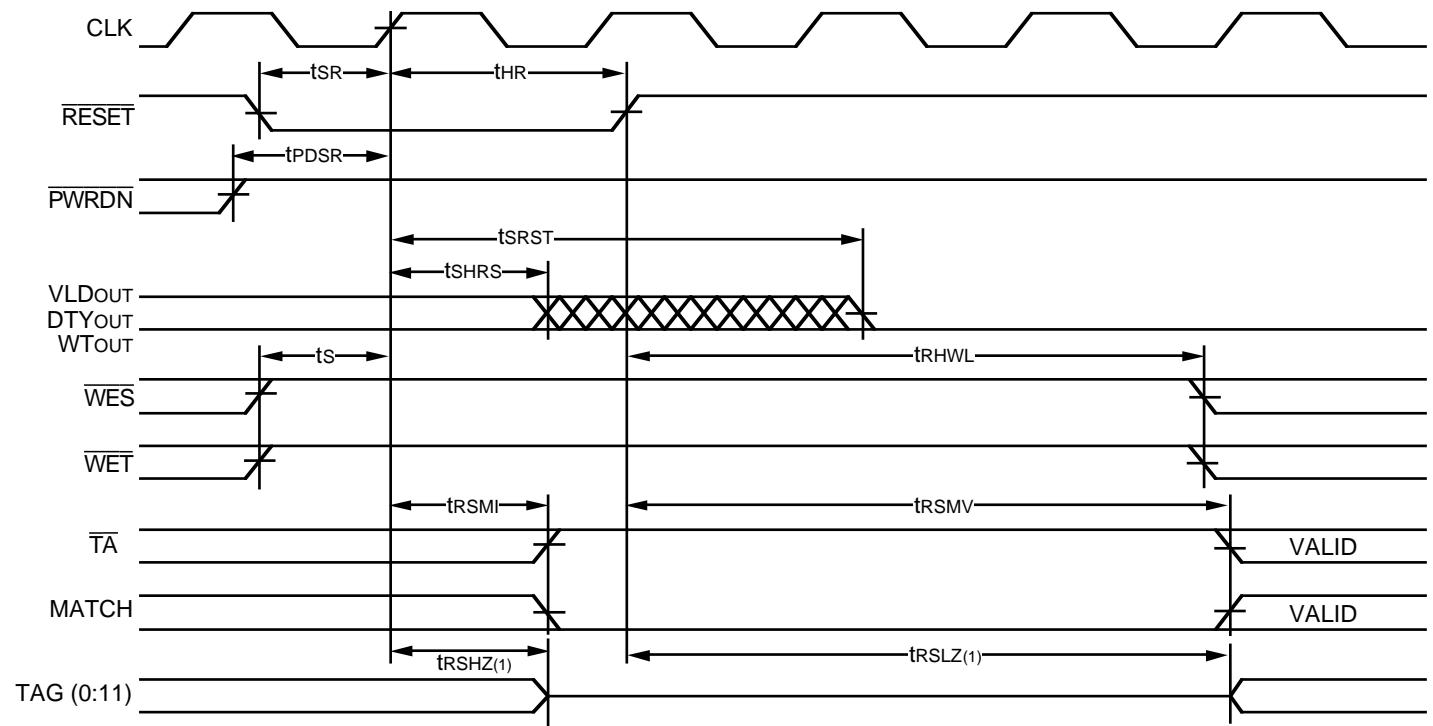
TIMING WAVEFORMS OF MATCH AND $\overline{\text{TA}}$ FUNCTIONS



NOTE:

1. Transition is measured $\pm 200\text{mV}$ from steady state.

TIMING WAVEFORMS OF $\overline{\text{RESET}}$ FUNCTION

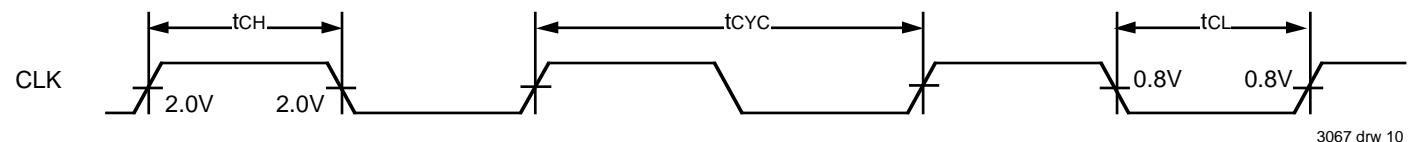


3067 drw 09

NOTE:

1. Transition is measured $\pm 200\text{mV}$ from steady state.

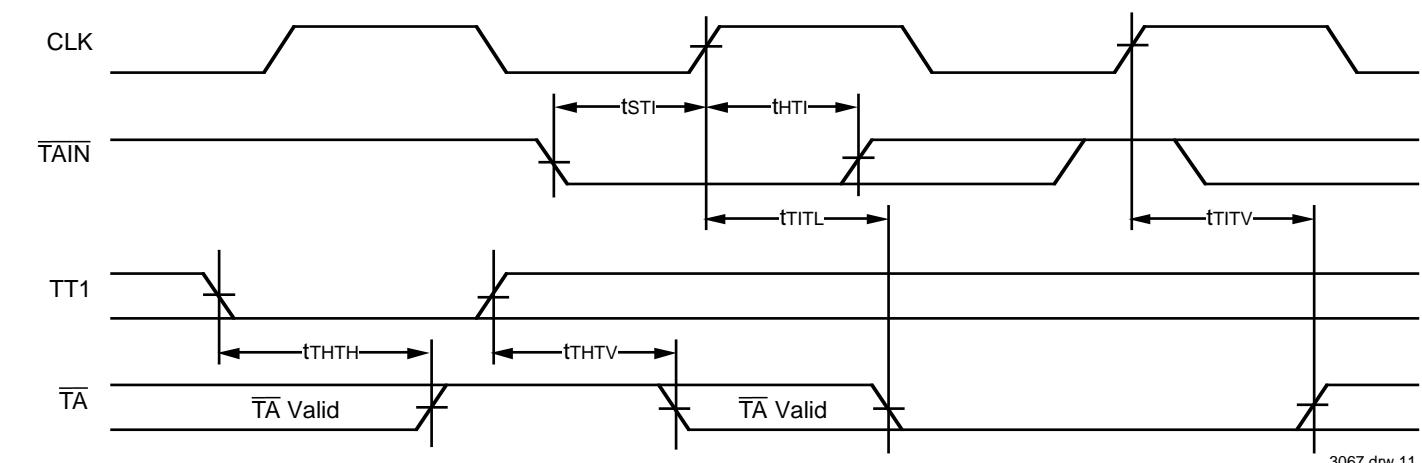
CLOCK TIMING WAVEFORM



3067 drw 10

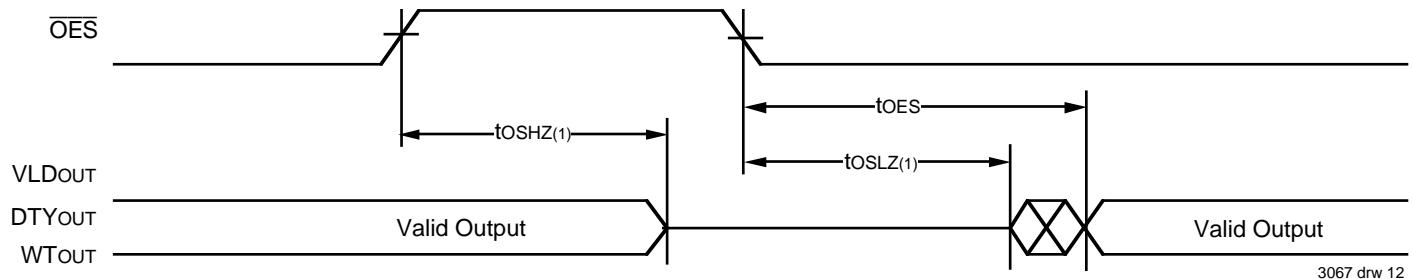
TIMING WAVEFORMS OF $\overline{\text{TA}}$ AND TT1 SIGNAL

Applies when SFUNC is LOW, and the internal WT bit is HIGH



3067 drw 11

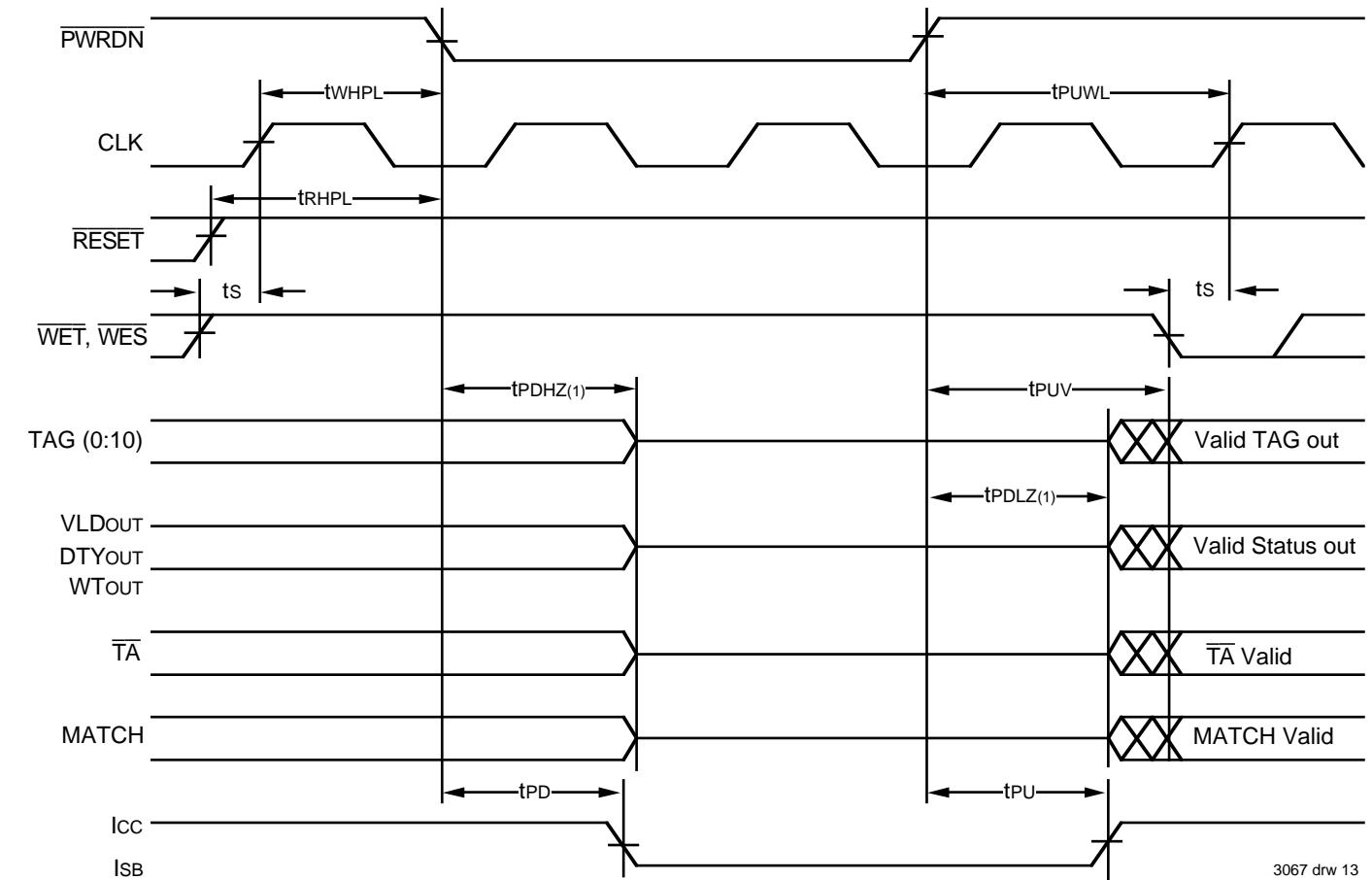
TIMING WAVEFORMS OF $\overline{\text{OES}}$ FUNCTION



NOTE:

1. Transition is measured $\pm 200\text{mV}$ from steady state.

TIMING WAVEFORMS OF POWER DOWN FUNCTION



NOTE:

1. Transition is measured $\pm 200\text{mV}$ from steady state.

ORDERING INFORMATION

IDT	71216	S	XX	PF	
	Device Type	Power	Speed	Package	
				PF	Plastic Thin Quad Flatpack (PN80-1)
			8 9 10 12		Speed in nanoseconds

3067 drw 14

Low Voltage PLL Clock Driver

The MPC970 is a 3.3V compatible, PLL based clock driver devices targeted for high performance RISC or CISC processor based systems.

- Fully Integrated PLL
- Output Frequency Up to 250MHz
- Compatible with PowerPC™ and Pentium™ Processors
- Output Frequency Configuration
- On-Board Crystal Oscillator
- 52-Lead TQFP Packaging
- ±50ps Typical Cycle-to-Cycle Jitter

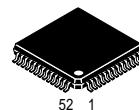
The MPC970 was designed specifically to drive today's PowerPC 601 and Pentium processors while providing the necessary performance to address higher frequency PowerPC 601 as well as PowerPC 603 and PowerPC 604 applications. The 2x_PCLK output can toggle at up to 250MHz while the remaining outputs can be configured to drive the other system clocks for MPC 601 based systems. As the processor based clock speeds increase the processor bus will likely run at one third or even one fourth the processor clock. The MPC970 supports the necessary waveforms to drive the BCLKEN input signal of the MPC 601 when the processor bus is running at a lower frequency than the processor. The MPC970 uses an advanced PLL design which minimizes the jitter generated on the outputs. The jitter specification is well within the requirements of the Pentium processor and meets the stringent preliminary specifications of the PowerPC 603 and PowerPC 604 processors. The application section of this data sheet deals in more detail with driving PowerPC and Pentium processor based systems.

The external feedback option of the MPC970 provides for a near zero delay between the reference clock input and the outputs of the device. This feature is required in applications where a master clock is being picked up off the backplane and regenerated and distributed on a daughter card. The advanced PLL of the MPC970 eliminates the dead zone of the phase detector and minimizes the jitter of the PLL so that the phase error variation is held to a minimum. This phase error uncertainty makes up a major portion of the part-to-part skew of the device.

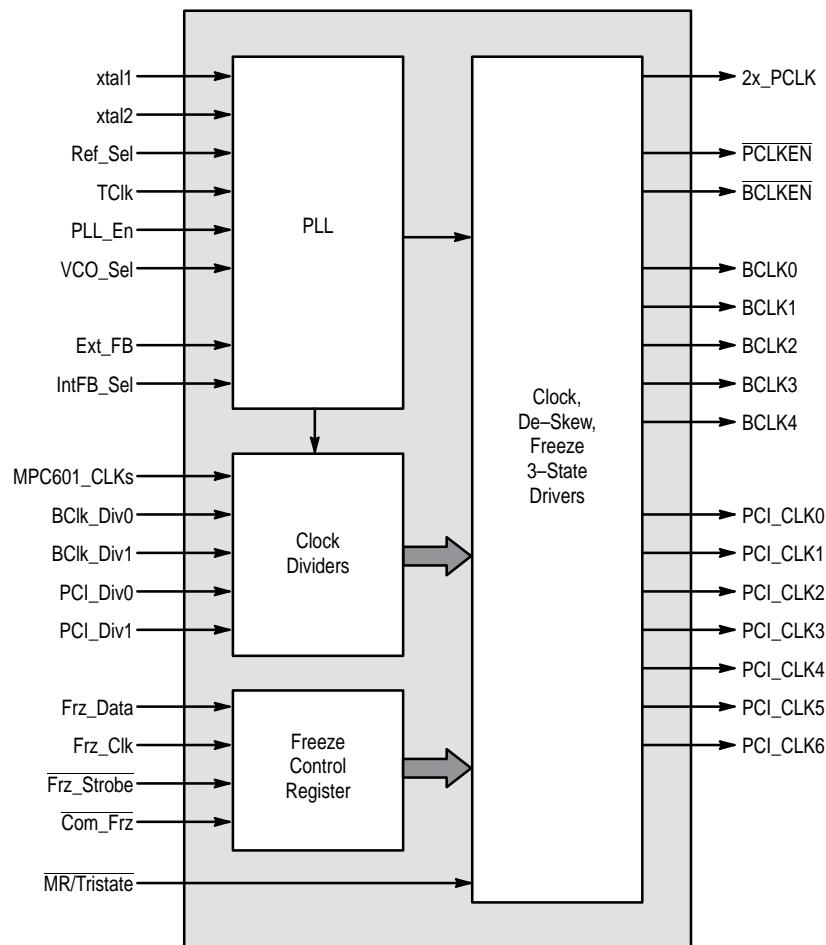
For single clock driver applications the MPC970 provides an internal oscillator and internal feedback to simplify board layout and minimize system cost. By using the on-board crystal oscillator the MPC970 acts as both the clock generator and distribution chip. The external component is a relatively inexpensive crystal rather than a more expensive oscillator. Since in single board applications the delay between the input reference and the outputs is inconsequential an internal feedback option is offered. The internal feedback simplifies board design in that the system designer need not worry about noise being coupled into the feedback line due to board parasitics and layout. The internal feedback is a fixed divide by 32 of the VCO. This divide ratio ensures that the input crystals will be ≤20MHz, thus keeping the crystal costs down and ensuring availability from multiple vendors.

MPC970

**LOW VOLTAGE
PLL CLOCK DRIVER**



52 1
FA SUFFIX
TQFP PACKAGE
CASE 848D-03

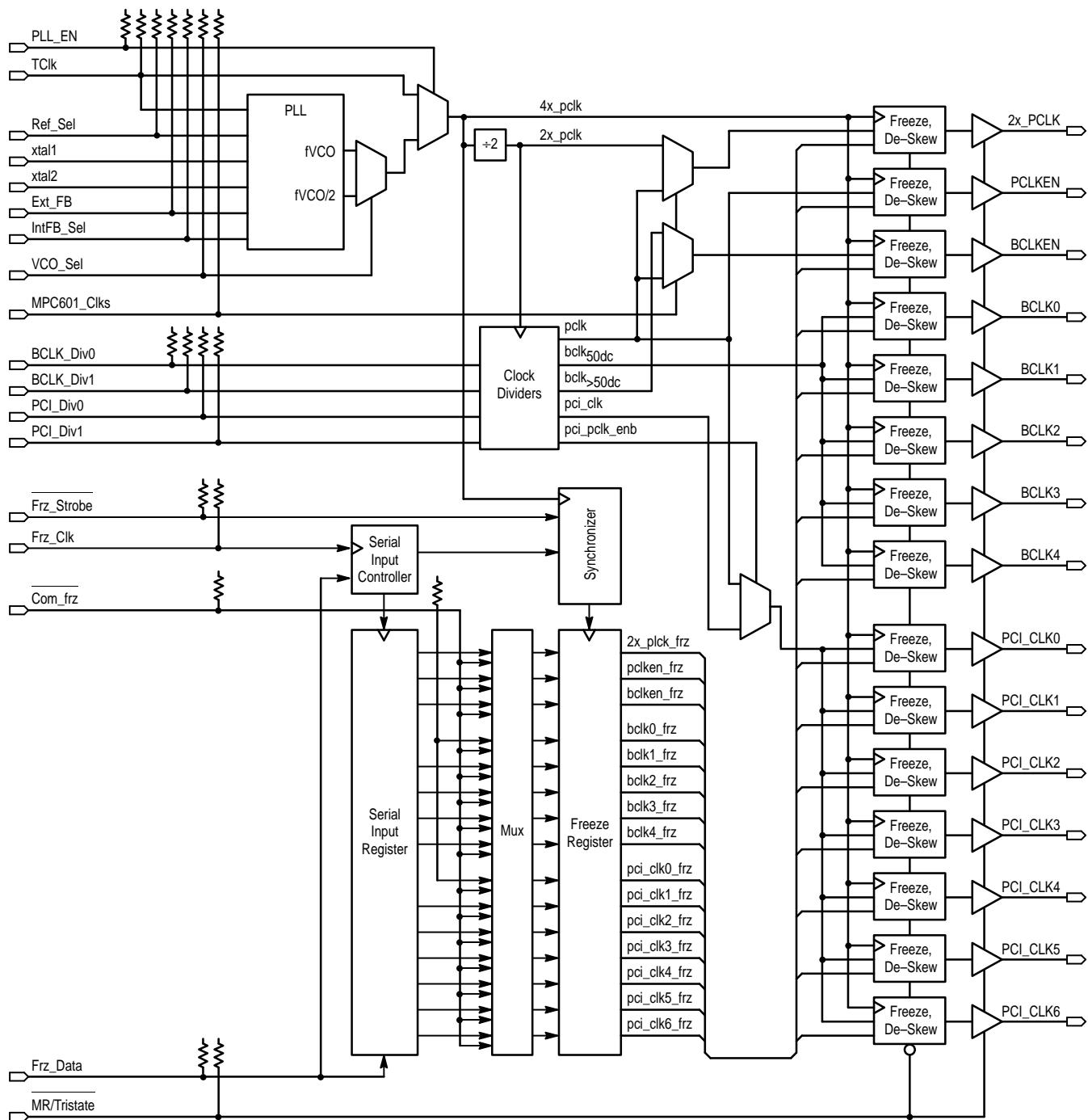


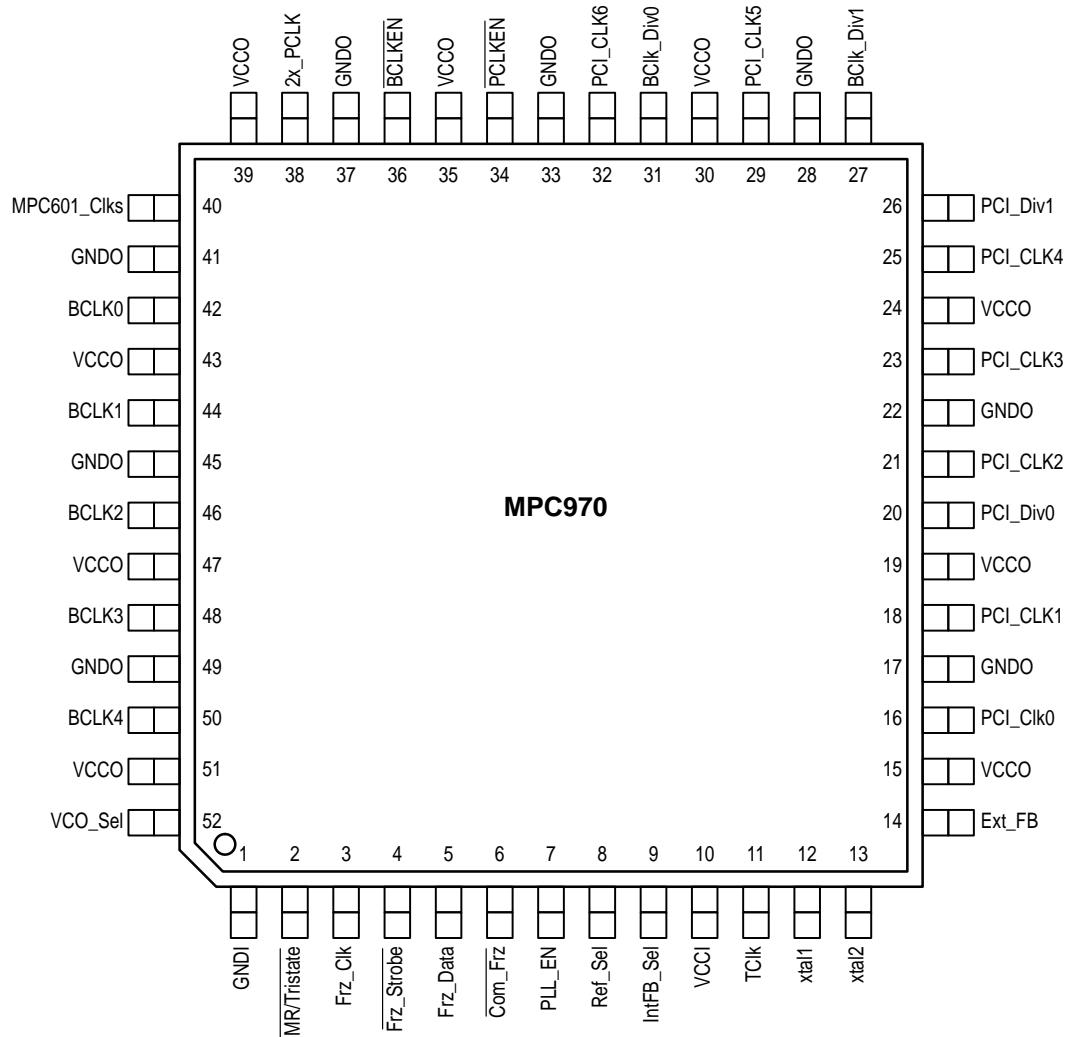
The MPC970 offers a very flexible output enable/disable scheme. This enable/disable scheme helps facilitate system debug as well as provide unique opportunities for system power down schemes to meet the requirements of "green" class machines. The MPC970 allows for the enabling of each output independently via a serial input port or a common enable/disable of all outputs simultaneously via a parallel control pin. When disabled or "frozen" the outputs will be locked in the "LOW" state, however the internal state machines will continue to run. Therefore when "unfrozen" the outputs will activate synchronous and in phase with those outputs which were not frozen. The freezing and unfreezing of outputs occurs only when they are already in the "LOW" state, thus the possibility of runt pulse generation is eliminated. A power-on reset will ensure that upon power up all of the outputs will be active.

For IC and board level testing a MR/Tristate input is provided. When pulled "LOW" all outputs will tristate and all internal flip flops will be reset. In addition the internal PLL can be bypassed and the fanout dividers and output buffers can be driven directly by the TClk input pin. Note that in this mode it will take a number of input clock pulses to cause output transitions as the TClk is fed through the internal dividers.

The MPC970 is fully 3.3V (3.6V for PowerPC 601 designs) compatible and requires no external loop filter components. All inputs accept LVCMS/LVTTL compatible levels while the outputs provide LVCMS levels with the capability to drive 50Ω transmission lines. For series terminated lines each MPC970 output can drive two 50Ω lines in parallel thus effectively doubling the fanout of the device.

SIMPLIFIED BLOCK DIAGRAM





FUNCTION TABLE 1

MPC601_Clks	2x_PCLK	PCLKEN	BCLKEN	BCLK	PCI_CLK
0	VCO/4	VCO/4	VCO/4	X	X
1	VCO/2	VCO/4	BCLK*	X	X

* Output is purposely delayed vs 2x_PCLK output.

FUNCTION TABLE 2

PCI_Div1	PCI_Div0	PCI_CLK	BCLK_Div1	BCLK_Div0	BCLK
0	0	BCLK	0	0	PCLKEN
0	1	BCLK/2	0	1	PCLKEN/2
1	0	BCLK/3	1	0	PCLKEN/3
1	1	PCLKEN	1	1	PCLKEN/4

FUNCTION TABLE 3

Control Pin	Logic '0'	Logic '1'
VCO_Sel	fVCO/2	fVCO
Ref_Sel	TCLK	Crystal Osc
PLL_En	Bypass PLL	Enable PLL
IntFB_Sel	Ext Feedback	Int Feedback

ABSOLUTE MAXIMUM RATINGS*

Symbol	Parameter	Min	Max	Unit
V _{CC}	Supply Voltage	-0.3	4.6	V
V _I	Input Voltage	-0.3	V _{DD} + 0.3	V
I _{IN}	Input Current	TBD	TBD	mA
T _{Stor}	Storage Temperature Range	-40	125	°C

* Absolute maximum continuous ratings are those values beyond which damage to the device may occur. Exposure to these conditions or conditions beyond those indicated may adversely affect device reliability. Functional operation under absolute-maximum-rated conditions is not implied.

DC CHARACTERISTICS (T_A = 0 to 70°C)

Symbol	Characteristic	Min	Max	Unit	Condition
V _{CC}	Power Supply Voltage	3.0	3.8	V	
I _{CC}	Quiescent Power Supply		250	mA	
V _{IL}	Input Voltage LOW	LVC MOS Inputs	0.3V _{DD}	V	
V _{IH}	Input Voltage HIGH	LVC MOS Inputs	0.7V _{DD}	TBD	V
I _{IH}	Input Current HIGH	LVC MOS Inputs	-100	µA	V _{IN} = V _{CC}
I _{IL}	Input Current LOW		-200	µA	V _{IN} = GND
V _{OH}	Output Voltage HIGH	V _{DD} -0.2		V	I _{OH} = -20mA ¹
V _{OL}	Output Voltage LOW		0.2	V	I _{OL} = 20mA ¹
I _{OZ}	Tristate Output Leakage Current	-10	10	µA	V _{OH} = V _{CC} or GND
C _{IN}	Input Capacitance		4	pF	
C _{pd}	Power Dissipation Capacitance			pF	
C _{OUT}	Output Capacitance		8	pF	

1. The MPC970 outputs can drive series or parallel terminated 50Ω (or 50Ω to V_{CC}/2) transmission lines on the incident edge (see Applications Info section).

PLL INPUT REFERENCE CHARACTERISTICS (T_A = 0 to 70°C)

Symbol	Characteristic	Min	Max	Unit	Condition
t _r , t _f	TCLK Input Rise/Falls		3.0	ns	
f _{ref}	Reference Input Frequency	10	Note 1	MHz	
f _{refDC}	Reference Input Duty Cycle	25	75	%	

1. Maximum input reference is limited by the VCO lock range and the feedback divider.

AC CHARACTERISTICS ($T_A = 0$ to 70°C)

Symbol	Characteristic	Min	Typ	Max	Unit	Condition	
fXtal	Crystal Oscillator Frequency	10		25	MHz	Note 4	
Fout ²	Maximum 2x_PCLK Output Frequency			200	MHz		
tDC ^{1,2}	Output Duty Cycle	45 35		55 65	%	Fout < 200MHz Fout ≥ 200MHz	
V _{OHAC}	AC Output HIGH Voltage	2.4 2.2			V	Fout < 200MHz Fout ≥ 200MHz	
V _{OLAC}	AC Output LOW Voltage			0.4 0.6	V	Fout < 200MHz Fout ≥ 200MHz	
t _{pw} ^{1,2}	2x_PCLK Pulse Width	1.75	2.27		ns	Fout = 200MHz	
t _{per}	Minimum Clock Out Period	4.85	4.91		ns	Fout = 200MHz	
t _{pd} ³	SYNC to Feedback Propagation Delay	Feedback = VCO/4 Feedback = VCO/8 Feedback = VCO/12 Feedback = VCO/16 Feedback = VCO/24 Feedback = VCO/36 Feedback = VCO/48	X ₁ ±150 X ₂ ±150 X ₃ ±150 X ₄ ±150 X ₅ ±150 X ₆ ±150 X ₇ ±150		ps		
f _{VCO}	VCO Lock Range	Feedback = VCO/4 Feedback = VCO/8 Feedback = VCO/12 Feedback = VCO/16 Feedback = VCO/24 Feedback = VCO/32 Feedback = VCO/36 Feedback = VCO/36	TBD TBD TBD TBD TBD TBD TBD TBD	200–700 200–700 200–700 200–700 200–700 200–700 200–700 200–700	TBD TBD TBD TBD TBD TBD TBD TBD	MHz	Ext Feedback Ext Feedback Ext Feedback Ext or Int Feedback Ext Feedback Int Feedback Ext Feedback Ext Feedback
t _{jitter} ^{1,2}	Output Jitter		±50 110 76	±100 190 210	ps	PLL Jitter 2x_P Period Variation Period Variation (Other)	
t _{skew} ^{1,2}	Output-to-Output Skew 2x_PCLK, PCLKEN, BCLKEN, BCLK 2x_PCLK, PCLKEN, BCLK PCI_CLK BCLK All			550 550 450 550 800	ps	MPC601_Clk = '0' MPC601_Clk = '1'	
t _{delay}	Propagation Delay 2x_PCLK to BCLKEN	100		850	ps	MPC601_Clk = '1'	
t _r , t _f ^{1,2}	Output Rise/Fall Time	0.15		1.5	ns	0.8 to 2.0V	
t _{lock}	PLL Lock Time			10	ms		
t _{PZL}	Output Enable Time MR/Tristate to Outputs			8	ns		
t _{PHZ} , t _{PLZ}	Output Disable Time MR/Tristate to Outputs			10	ns		
f _{MAX}	Maximum Frz_Clk Frequency			20	MHz		
t _s	Setup Time Frz_Data to Frz_Clk Com_Frz to Frz_Strobe	8 5					
t _h	Hold Time Frz_Clk to Frz_Data Frz_Strobe to Com_Frz	8 5					

1. Measured at 1.4V.

2. Drive 50Ω transmission lines.

3. X₁, X₂, X₃, X₄, X₅, X₆, and X₇ all to be determined. The specs hold only when the 970 or 971 is used in the external feedback mode.

4. See Applications Info section for more crystal information.

DETAILED PIN DESCRIPTIONS

The following gives a brief description of the functionality of the MPC970 I/O. Unless explicitly stated all inputs are LVCMOS/LVTTL compatible with internal pull up resistors. All outputs are LVCMOS level outputs which are capable of driving two series terminated 50Ω transmission lines on the incident edge.

xtal1, xtal2

These input pairs are the pins which differentiate the MPC970 and the MPC971. For the MPC970 the xtal1 and xtal2 pins represent the external crystal connections to the internal oscillator. The crystal oscillator is completely self contained, there are no external components required. The oscillator is specified to function for crystals of up to 50MHz. Exact crystal specifications are outlined in the applications section.

VCO_Sel

The VCO_Sel pin allows the user to further divide the internal VCO frequency for the generation of lower frequencies at the outputs. The VCO_Sel pin should be used to set the VCO into its most optimum range. Refer to the applications section for more details on the VCO frequency range. A logic '1' on the VCO_Sel pin will bypass the internal +2.

TClk

The TClk input serves a dual purpose; it can be used as either a reference clock input for the PLL from an external frequency source or it can be used as a board level test clock in the PLL bypass mode.

PLL_En

The PLL_En pin allows the TClk input to be routed around the PLL for system test and debug. When pulled low the MPC970 will be placed in the test mode. Note that the TClk input will be routed through the divider chain. For instance in the PowerPC 601 microprocessor clock generation mode the TClk input will toggle twice for each toggle on the 2x_PCLK output. Depending on the states of the frequency divider select pins this ratio may be higher.

Frz_Data

Frz_Data is the serial data input for the output freeze function of the device. Refer to the applications section for more information on the freeze functionality.

Frz_Clk

Frz_Clk is the serial freeze logic clock input. Refer to the applications section for more information on the freeze functionality.

Frz_Strobe

The Frz_Strobe input is used to freeze or unfreeze all of the outputs simultaneously. Refer to the applications section for more information on the freeze functionality.

Com_Frz

The Com_Frz input allows the user to enable/disable all of the outputs with the control of a single pin. The action will take place upon a high to low transition of the Frz_Strobe input.

BClk_Div0:1

The BClk_Div inputs are used to program the VCO divide ratio for the BCLK outputs. These inputs also set the divide ratio of the BCLKEN output to be equal in frequency to the BCLK outputs when the device is in the MPC601_Clks mode. The BClk_Div inputs set the frequency as follows:

BClk_Div1	BClk_Div0	BCLK Frequency
0	0	PCLKEN
0	1	PCLKEN/2
1	0	PCLKEN/3
1	1	PCLKEN/4

In most applications these inputs will be strapped to the appropriate power rails.

PCI_Div0:1

The PCI_Div inputs set the division ratio for the PCI_CLKs. The PCI_CLKs are set relative to the BCLK or the PCLKEN output such that you can upgrade the processor bus and maintain the PCI bus frequency in the currently defined ≤ 33MHz range. The PCI_Div inputs set the PCI_CLKs as follows:

PCI_Div1	PCI_Div0	PCI_CLK Frequency
0	0	BCLK
0	1	BCLK/2
1	0	BCLK/3
1	1	PCLKEN

In a typical application these inputs will be strapped to the appropriate power rail.

MPC601_Clks

The MPC601_Clks input will configure the outputs to drive the PowerPC 601 microprocessor when pulled HIGH or left open. When pulled LOW it will configure the 2xPCLK, PCLKEN and BCLKEN all into a VCO/4 mode. In this mode the MPC970 will have three more outputs available to drive clock loads on the processor bus for PowerPC 603, PowerPC 604 or Pentium microprocessor based systems.

Ext_FB

The Ext_FB pin is an input to the phase detector of the PLL which is tied to an external feedback output. Typically this feedback will be one of the lowest frequency outputs of the MPC970.

IntFB_Sel

The IntFB_Sel input selects whether the internal feedback signal or an external feedback signal is routed to the phase detector of the PLL. The default mode, pulled HIGH via the internal pull up resistor, is to select the internal feedback.

MPC970

MR/Tristate

The MR/Tristate input when pulled LOW will reset all of the internal flip flops and also tristate all of the clock outputs. This input is used primarily for IC and board level test.

Ref_Sel

The Ref_Sel input allows the user to choose between two sources for the PLL reference frequency. For the MPC970, LOW on Ref_Sel will choose the LVCMS TCLK input. For the MPC970, a HIGH on Ref_Sel will choose the crystal oscillator input.

2x_PCLK

In general the outputs are named based on the implementation in a PowerPC 601 microprocessor based system. In the MPC601_Clk mode the 2x_PCLK will run at half the internal VCO frequency. With a maximum internal VCO frequency of 1000MHz this output could theoretically toggle at 500MHz, in practice however the output can toggle only as fast as 300MHz. This frequency will be required on future enhancements to the MPC 601 microprocessor. When the MPC970 is taken out of the MPC601_Clk mode the 2xPCLK will run at a VCO/4 frequency. This divide ratio will place this output frequency in the present and future processor bus speeds of the PowerPC 603, PowerPC 604 and Pentium microprocessors. The 2x_PCLK output is a 50% duty cycle LVCMS output.

PCLKEN

The PCLKEN output is designed to drive the PCLKEN input of the PowerPC 601 microprocessor when the MPC970 is in the MPC601_Clk mode. The PCLKEN output frequency is one half that of the 2x_PCLK output, a divide by four of the internal VCO. The PCLKEN output runs at the same frequency regardless of the state of the frequency divide controls. The toggle frequency of this output is well placed for driving the PowerPC 603, PowerPC 604 and Pentium processor buses. The PCLKEN output is a 50% duty cycle LVCMS output.

BCLKEN

The BCLKEN output is designed to drive the BCLKEN input of the PowerPC 601 microprocessor when the MPC970 is in the MPC601_Clk mode. The BCLKEN toggles at the same frequency as the BCLK outputs as described earlier. However when the BCLKEN output is a divide by three or a divide by four of the PCLKEN output the duty cycle is 66/33 and 75/25 respectively per the requirement of the MPC 601 processor. In addition to meet the HOLD time spec for the BCLKEN input of the MPC 601 the BCLKEN output of the MPC970 lags the 2xPCLK output by no less than 100ps. When the MPC970 is not in the MPC601_Clk mode the BCLKEN output is set at a fixed divide by four from the internal VCO. In addition in this mode the BCLKEN output does NOT lag the other outputs, but rather is synchronous within the Output-to-Output skew spec of the device.

BCLK0:4

The BCLK outputs are designed to drive the clock loads on the processor bus of either the PowerPC or Pentium microprocessors. The most common practice in "non MPC 601" applications will be to place these outputs in the PCLKEN/1 mode and combine them with the above outputs to drive all of the loads on the processor bus. The division ratios do allow for the swap of these outputs with the PCI_CLK outputs if more clocks are needed to drive the processor bus. For PowerPC 601 microprocessor based systems the division ratios allow the processor internal speeds to be increased while maintaining reasonable speeds for the L2 cache and the PCI bridge chip. The BCLK outputs are 50% duty cycle LVCMS outputs.

PCI_Clk0:6

As the name would suggest the PCI_CLK outputs are designed to drive the PCI bus clock loads in a typical microprocessor based system. The division ratios allow for these outputs to remain in the $\leq 33\text{MHz}$ PCI bus speeds for various common processor bus speeds as well as higher future processor bus speeds. These outputs can also be programmed to run at the processor bus speeds if more processor bus clocks are required. The PCI_CLK outputs are 50% duty cycle LVCMS outputs.

APPLICATIONS INFORMATION

Programming the MPC970

The MPC970 is very flexible in the programming of the frequency relationships of the various outputs as well as the relationships between the input references and outputs. The purpose of this section is to outline the various relationships. Although not exhaustive the hope is that enough information is supplied to allow the customers to tailor the I/O relationships for their specific applications.

The VCO used in the MPC970 is a differential ring oscillator. The VCO exhibits a very wide frequency range to allow for a great deal of flexibility to the end user. Special design techniques were used in the overall PLL design to

keep the relatively high gain of the VCO from significantly impacting the jitter of the PLL.

Table 1 tabulates the various output frequencies for the different modes defined by the division select input pins. In this table the VCO_Sel pin is high so that the +2 prescaler is bypassed. Note that the +32 feedback is always fed directly from the VCO and is thus unaffected by the level on the VCO_Sel input. Table 1 shows each of the output frequencies as a function of the VCO frequency. The two VCO ranges can be used to plug in values to get the actual frequencies. When the internal feedback option is used the multiplication factor of the device will equal 32 divided by the

output divide ratio. If the VCO_Sel pin is "LOW" the multiplication factor will be reduced further by 2. (See "Using the On-Board Crystal Oscillator" section of this datasheet.)

Using the External Feedback Feature of the MPC970/71

In applications where the relationship between the output waveforms and the input waveforms are critical the external feedback option will likely be used. Table 1 and Table 2 are still appropriate for establishing the potential output frequency relationships. The input reference frequency for external feedback applications will be equal to the frequency of the feedback signal. As a result the use of the external feedback yields a number of potential input to output frequency multiplication factors which are not available using the internal feedback. Using the external feedback the device can function as a zero delay buffer and could multiply the input from 4 to as much as 48. In practice however the multiplication factor is limited by the loop dynamics of the PLL. The MPC970 PLL was optimized for an input reference frequency or greater than 10MHz. Frequencies lower than 10MHz will tend to pass through the filter and add jitter to the PLL. In addition the PLL was optimized for feedback divide ratios of between 8 and 64. The user should avoid using the device with feedback divide ratios outside of this range. For the external feedback case the feedback divide ratio will include the +2 (if VCO_Sel is LOW) plus the output divider for

the feedback output. If, for instance the MPC970 is to be used as a zero delay buffer the VCO_Sel pin should be pulled LOW and all of the outputs should be set in a VCO/4 mode. This would produce a feedback ratio of +8. Several potential configurations using the external feedback are pictured in Figure 1 through Figure 4.

The external feedback option of the MPC970 is critical for applications in which more than one clock driver need to be synchronized. The external feedback option ensures that the feed through delay is the same as the feedback delay. This functionality removes propagation delay variation as a factor in the determination of part to part skew. The low jitter PLL used in the MPC970 has a near zero deadband phase detector and very little part to part variability. The result is a very low phase error variability in the product. When coupled with the output to output skew the phase error variability accounts for the part to part skew of the device. From the specification table one sees that the worst case part to part skew of the device is 800ps, assuming that there is zero skew in the multiple reference inputs. For multiple MPC970 applications if the lowest generated output frequency is used as the feedback signal the devices will be guaranteed to be synchronized. For applications where the lowest frequency is not used as the reference or where the internal feedback is used there is no way to guarantee that the multiple devices will be phase synchronized.

Table 1. Programmable Output Frequency Relationships (MPC601_Clk = 'HIGH'; VCO_Sel = 'HIGH')

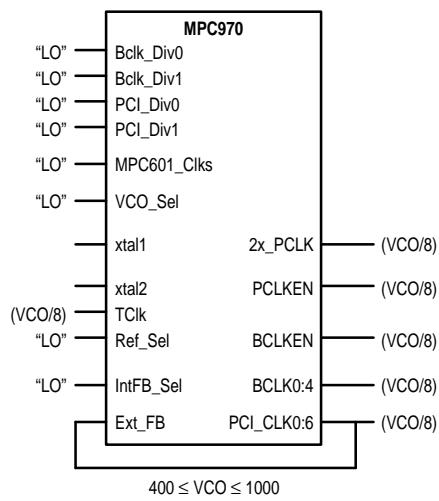
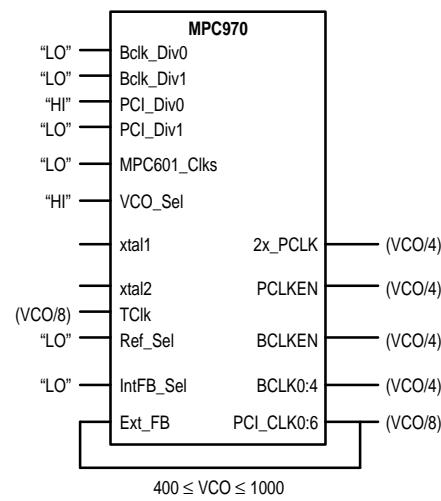
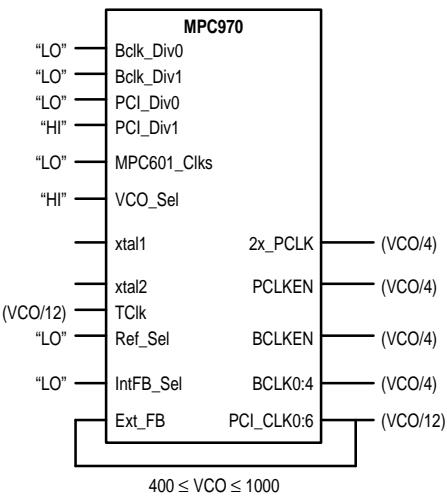
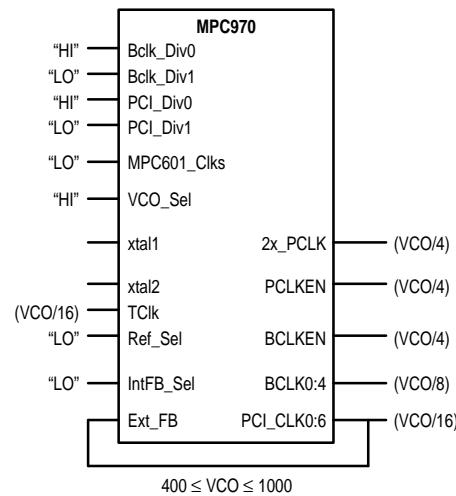
INPUTS				OUTPUTS				
PCI_Div1	PCI_Div0	BCLK_Div1	BCLK_Div0	2x_PCLK	PCLKEN	BCLKEN*	BCLK	PCI_CLK
0	0	0	0	VCO/2	VCO/4	VCO/4	VCO/4	VCO/4
0	0	0	1	VCO/2	VCO/4	VCO/8	VCO/8	VCO/8
0	0	1	0	VCO/2	VCO/4	VCO/12	VCO/12	VCO/12
0	0	1	1	VCO/2	VCO/4	VCO/16	VCO/16	VCO/16
0	1	0	0	VCO/2	VCO/4	VCO/4	VCO/4	VCO/8
0	1	0	1	VCO/2	VCO/4	VCO/8	VCO/8	VCO/16
0	1	1	0	VCO/2	VCO/4	VCO/12	VCO/12	VCO/24
0	1	1	1	VCO/2	VCO/4	VCO/16	VCO/16	VCO/32
1	0	0	0	VCO/2	VCO/4	VCO/4	VCO/4	VCO/12
1	0	0	1	VCO/2	VCO/4	VCO/8	VCO/8	VCO/24
1	0	1	0	VCO/2	VCO/4	VCO/12	VCO/12	VCO/36
1	0	1	1	VCO/2	VCO/4	VCO/16	VCO/16	VCO/48
1	1	0	0	VCO/2	VCO/4	VCO/4	VCO/4	VCO/4
1	1	0	1	VCO/2	VCO/4	VCO/8	VCO/8	VCO/4
1	1	1	0	VCO/2	VCO/4	VCO/12	VCO/12	VCO/4
1	1	1	1	VCO/2	VCO/4	VCO/16	VCO/16	VCO/4

* BCLK_En output is delayed relative to other outputs

Table 2. Programmable Output Frequency Relationships (MPC601_Clks = 'LOW'; VCO_Sel = 'HIGH')

INPUTS				OUTPUTS				
PCI_Div1	PCI_Div0	BCLK_Div1	BCLK_Div0	2x_PCLK	PCLKEN	BCLKEN*	BCLK	PCI_CLK
0	0	0	0	VCO/4	VCO/4	VCO/4	VCO/4	VCO/4
0	0	0	1	VCO/4	VCO/4	VCO/4	VCO/8	VCO/8
0	0	1	0	VCO/4	VCO/4	VCO/4	VCO/12	VCO/12
0	0	1	1	VCO/4	VCO/4	VCO/4	VCO/16	VCO/16
0	1	0	0	VCO/4	VCO/4	VCO/4	VCO/4	VCO/8
0	1	0	1	VCO/4	VCO/4	VCO/4	VCO/8	VCO/16
0	1	1	0	VCO/4	VCO/4	VCO/4	VCO/12	VCO/24
0	1	1	1	VCO/4	VCO/4	VCO/4	VCO/16	VCO/32
1	0	0	0	VCO/4	VCO/4	VCO/4	VCO/4	VCO/12
1	0	0	1	VCO/4	VCO/4	VCO/4	VCO/8	VCO/24
1	0	1	0	VCO/4	VCO/4	VCO/4	VCO/12	VCO/36
1	0	1	1	VCO/4	VCO/4	VCO/4	VCO/16	VCO/48
1	1	0	0	VCO/4	VCO/4	VCO/4	VCO/4	VCO/4
1	1	0	1	VCO/4	VCO/4	VCO/4	VCO/8	VCO/4
1	1	1	0	VCO/4	VCO/4	VCO/4	VCO/12	VCO/4
1	1	1	1	VCO/4	VCO/4	VCO/4	VCO/16	VCO/4

* BCLK_En output is coincident with other outputs

**Figure 1. External Feedback Configuration 1****Figure 2. External Feedback Configuration 2****Figure 3. External Feedback Configuration 3****Figure 4. External Feedback Configuration 4**

Using the On-Board Crystal Oscillator

The MPC970 features an on-board crystal oscillator to allow for seed clock generation as well as final distribution. The on-board oscillator is completely self contained so that the only external component required is the crystal. As the oscillator is somewhat sensitive to loading on its inputs the user is advised to mount the crystal as close to the MPC970 as possible to avoid any board level parasitics. To facilitate co-location surface mount crystals are recommended, but not required. In addition, with crystals with a higher shunt capacitance, it may be necessary to place a 1k resistor across the two crystal leads.

The oscillator circuit is a series resonant circuit as opposed to the more common parallel resonant circuit, this eliminates the need for large on-board capacitors. Because the design is a series resonant design for the optimum frequency accuracy a series resonant crystal should be used (see specification table below). Unfortunately most off the shelf crystals are characterized in a parallel resonant mode. However a parallel resonant crystal is physically no different than a series resonant crystal, a parallel resonant crystal is simply a crystal which has been characterized in its parallel resonant mode. Therefore in the majority of cases a parallel specified crystal can be used with the MPC970 with just a minor frequency error due to the actual series resonant frequency of the parallel resonant specified crystal. Typically a parallel specified crystal used in a series resonant mode will exhibit an oscillatory frequency a few hundred ppm lower than the specified value. For most processor implementations a few hundred ppm translates into kHz inaccuracies, a level which does not represent a major issue.

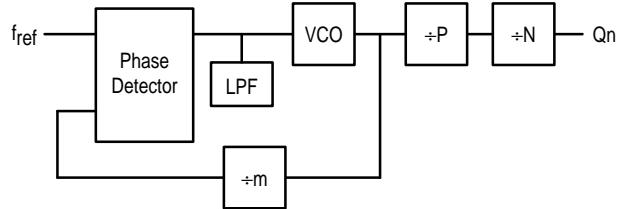
Table 3. Crystal Specifications

Parameter	Value
Crystal Cut	Fundamental at Cut
Resonance	Series Resonance*
Frequency Tolerance	$\pm 75\text{ppm}$ at 25°C
Frequency/Temperature Stability	$\pm 150\text{ppm}$ 0 to 70°C
Operating Range	0 to 70°C
Shunt Capacitance	5–7pF
Equivalent Series Resistance (ESR)	50 to 80Ω
Correlation Drive Level	100 μW
Aging	5ppm/Yr (First 3 Years)

* See accompanying text for series versus parallel resonant discussion.

The MPC970 is a clock driver which was designed to generate outputs with programmable frequency relationships and not a synthesizer with a fixed input frequency. As a result the crystal input frequency is a function of the desired output frequency. For a design which utilizes the external feedback to the PLL the selection of the crystal frequency is straight forward; simply chose a crystal which is equal in frequency to the fed back signal. To determine the crystal required to

produce the desired output frequency for an application which utilizes internal feedback the block diagram of Figure 5 should be used. The P and the M values for the MPC970 are also included in Figure 5. The M values can be found in the configuration tables included in this applications section.



$$f_{\text{ref}} = \frac{f_{\text{VCO}}}{m}, \quad f_{\text{VCO}} = f_{\text{Qn}} \cdot N \cdot P$$

$$\therefore f_{\text{ref}} = \frac{f_{\text{Qn}} \cdot N \cdot P}{m}$$

$$m = 32$$

$$P = 1 (\text{VCO_Sel}='1'), 2 (\text{VCO_Sel}='0')$$

Figure 5. PLL Block Diagram

For the MPC970 clock driver, the following will provide an example of how to determine the crystal frequency required for a given design.

Given:

$$\begin{aligned} 2x_{\text{PCLK}} &= 200\text{MHz} \\ \text{PCLKEN} &= 100\text{MHz} \\ \text{BCLK} &= 50\text{MHz} \\ \text{PCI_CLK} &= 25\text{MHz} \\ \text{VCO_SEL} &= '1' \end{aligned}$$

$$f_{\text{ref}} = \frac{f_{\text{Qn}} \cdot N \cdot P}{m}$$

From Table 3

$$\text{PCI_CLK} = \text{VCO}/16 \text{ then } N = 16$$

or

$$\text{PCLKEN} = \text{VCO}/4 \text{ then } N = 4$$

From Figure 5

$$m = 32 \text{ and } P = 1$$

$$f_{\text{ref}} = \frac{25 \cdot 16 \cdot 1}{32} = 12.5\text{MHz} \text{ or } \frac{100 \cdot 4 \cdot 1}{32} = 12.5\text{MHz}$$

Driving Transmission Lines

The MPC970 clock driver was designed to drive high speed signals in a terminated transmission line environment. To provide the optimum flexibility to the user the output drivers were designed to exhibit the lowest impedance possible. With an output impedance of less than 10Ω the drivers can drive either parallel or series terminated transmission lines. For more information on transmission lines the reader is referred to application note AN1091 in the Timing Solutions brochure (BR1333/D).

In most high performance clock networks point-to-point distribution of signals is the method of choice. In a point-to-point scheme either series terminated or parallel

MPC970

terminated transmission lines can be used. The parallel technique terminates the signal at the end of the line with a 50Ω resistance to VCC/2. This technique draws a fairly high level of DC current and thus only a single terminated line can be driven by each output of the MPC970 clock driver. For the series terminated case however there is no DC current draw, thus the outputs can drive multiple series terminated lines in parallel. Figure 6 illustrates an output driving a single series terminated line vs two series terminated lines in parallel. When taken to its extreme the fanout of the MPC970 clock driver is effectively doubled due to its capability to drive multiple lines.

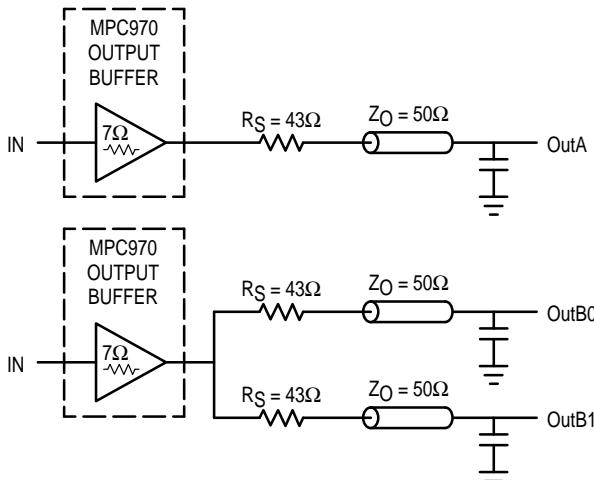


Figure 6. Single versus Dual Transmission Lines

The waveform plots of Figure 7 show the simulation results of an output driving a single line vs two lines. In both cases the drive capability of the MPC970 output buffers is more than sufficient to drive 50Ω transmission lines on the incident edge. Note from the delay measurements in the simulations a delta of only 43ps exists between the two differently loaded outputs. This suggests that the dual line driving need not be used exclusively to maintain the tight output-to-output skew of the MPC970. The output waveform in Figure 7 shows a step in the waveform, this step is caused by the impedance mismatch seen looking into the driver. The parallel combination of the 43Ω series resistor plus the output impedance does not match the parallel combination of the line impedances. The voltage wave launched down the two lines will equal:

$$VL = VS \left(\frac{Z_0}{R_s + R_o + Z_0} \right) = 3.0 \left(\frac{25}{53.5} \right) = 1.40V$$

At the load end the voltage will double, due to the near unity reflection coefficient, to 2.8V. It will then increment towards the quiescent 3.0V in steps separated by one round trip delay (in this case 4.0ns).

Since this step is well above the threshold region it will not cause any false clock triggering, however designers may be uncomfortable with unwanted reflections on the line. To better match the impedances when driving multiple lines the situation in Figure 8 should be used. In this case the series terminating resistors are reduced such that when the parallel combination is added to the output buffer impedance the line impedance is perfectly matched.

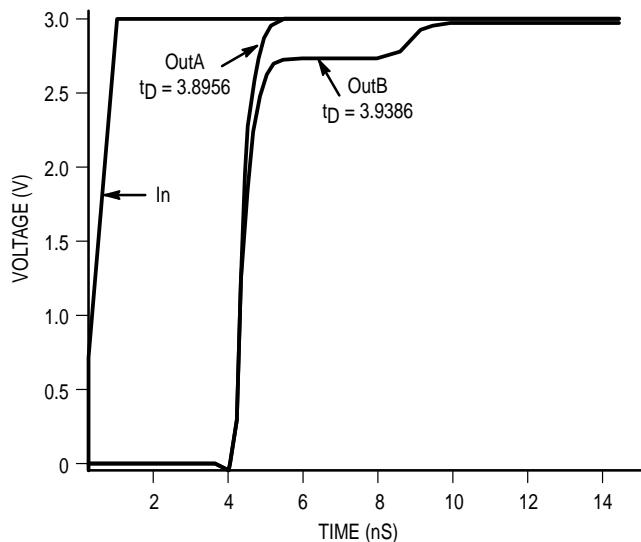


Figure 7. Single versus Dual Waveforms

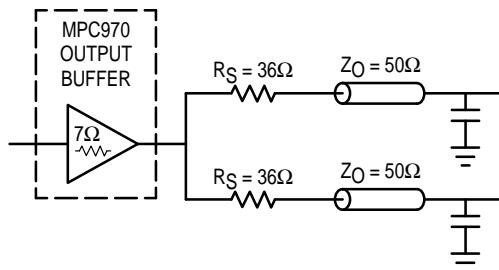


Figure 8. Optimized Dual Line Termination

SPICE level output buffer models are available for engineers who want to simulate their specific interconnect schemes. In addition IV characteristics are in the process of being generated to support the other board level simulators in general use.

Using the Output Freeze Circuitry

With the recent advent of a "green" classification for computers the desire for unique power management among system designers is keen. The individual output enable control of the MPC970 allows designers, under software control, to implement unique power management schemes into their designs. Although useful, individual output control at the expense of one pin per output is too high, therefore a simple serial interface was derived to economize on the control pins.

The freeze control logic provides two mechanisms through which the MPC970 clock outputs may be frozen (stopped in the logic '0' state):

The first freeze mechanism allows serial loading of the 13-bit Serial Input Register, this register contains one programmable freeze enable bit for 13 of the 15 output clocks. The BCLK0 and PCI_CLK0 outputs cannot be frozen with the serial port, this avoids any potential lock up situation

should an error occur in the loading of the Serial Input Register. The user may programmably freeze an output clock by writing logic '0' to the respective freeze enable bit. Likewise, the user may programmably unfreeze an output clock by writing logic '1' to the respective enable bit.

The second freeze mechanism allows all 15 clocks to be frozen simultaneously by placing a logic '0' on the Com_Frz input and then issuing a low going pulse on the Frz_Strobe input. Likewise, all 15 clocks can be simultaneously unfrozen by placing logic '1' on the Com_Frz input and then issuing a low-going pulse on the Frz_Strobe input. Note that all 15 clocks are affected by the Frz_Strobe freeze logic.

The freeze logic will never force a newly-frozen clock to a logic '0' state before the time at which it would normally transition there. The logic simply keeps the frozen clock at logic '0' once it is there. Likewise, the freeze logic will never force a newly-unfrozen clock to a logic '1' state before the time at which it would normally transition there. The logic re-enables the unfrozen clock during the time when the respective clock would normally be in a logic '0' state, eliminating the possibility of 'runt' clock pulses.

The user may write to the Serial Input register through the Frz_Data input by supplying a logic '0' start bit followed serially by 13 NRZ freeze enable bits. After the 13th freeze enable bit the Frz_Data signal must be left in (or returned to) a logic '1' state (Figure 9). The period of each Frz_Data bit equals the period of the free-running Frz_Clk signal. The Frz_Data serial transmission should be timed so the MPC970 can sample each Frz_Data bit with the rising edge of the free-running Frz_Clk signal.

Start Bit	D0	D1	D2	D3	D4	D5	D6	D7	D8	D9	D10	D11	D12
-----------	----	----	----	----	----	----	----	----	----	----	-----	-----	-----

D0 is the control bit for 2x_PCLK
D1 is the control bit for PCLKEN
D2 is the control bit for BCLKEN
D3-D6 are the control bits for BCLK1-BCLK4
D7-D12 are the control bits for PCI_CLK1-PCI_CLK6

Figure 9. Freeze Data Input Protocol

The user can combine the two freeze capabilities to simplify system level implementation. The serial input port can be used to establish the freeze mask to disable the appropriate outputs. The Frz_Strobe input can then be used to unfreeze the outputs without having to serially load an "all unfrozen" freeze mask.

Driving the PowerPC 601 Microprocessor

The MPC601 processor requires three clock inputs from the MPC970 clock driver. A 2x_PCLK input at twice the internal MPC 601 clock rate and the PCLKEN and BCLKEN signals used to mask internal clock edges. The PCLKEN signal always runs at one half the 2x_PCLK signal while the BCLKEN signal can run at 1x, 1/2x, 1/3x or 1/4x the PCLK input signal depending on the speed of the processor bus. When the BCLKEN signal is running at 1/3 or 1/4 the PCLK input the input duty cycle must be 66/33 and 75/25 respectively. In addition, as shown in Figure 10, to satisfy the BCLKEN to 2x_PCLK Hold specification the BCLKEN signal must be at least coincident with the 2x_PCLK edge. To simplify board level implementation it would be desirable that the BCLKEN signal actually lag the 2x_PCLK by a few hundred picoseconds. The MPC970 insures that its BCLKEN output always lags the 2x_PCLK input by at least 300ps.

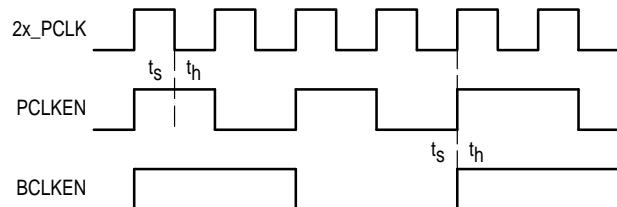


Figure 10. MPC601 Setup and Hold Times

Table 4 illustrates some typical MPC 601 system frequencies which can be realized using the MPC970 clock driver.

Table 4. Common MPC601 System Frequencies

2x_PCLK	PCLK	BCLK	PCI_CLK
240	120	60(1/2x)	30(1/2x)
240	120	40(1/3x)	20(1/2x)
240	120	30(1/4x)	30(1x)
200	100	50(1/2x)	25(1/2x)
200	100	33(1/3x)	33(1x)
200	100	25(1/4x)	25(1x)
160	80	40(1/2x)	20(1/2x)
160	80	20(1/4x)	20(1x)
132	66	66(1x)	33(1/2x)
132	66	33(1/2x)	33(1x)

MPC970

Driving the PowerPC 603, PowerPC 604 and Pentium Microprocessors

The PowerPC 603, PowerPC 604 and Pentium processors differ from the MPC 601 processor in that the processor input clocks are at the same frequency as the processor bus. A typical system for these processors will include 8 – 16 clock loads on the processor bus. When the MPC970 is taken out of the MPC601_Clk mode there are a total of 8 “non PCI_CLK” outputs which can be run at the processor bus speeds for these microprocessors. Since each output can drive two series terminated transmission lines the MPC970 can support point to point clock distribution for up to 16 loads on the processor bus. In addition there will be 7 PCI_CLK outputs which can drive up to 14 loads on the PCI bus.

If more clock loads are present on the processor bus the

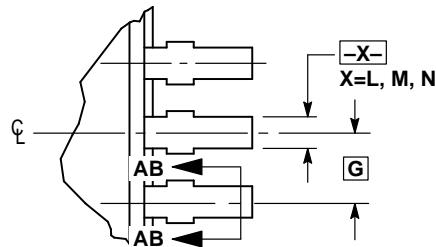
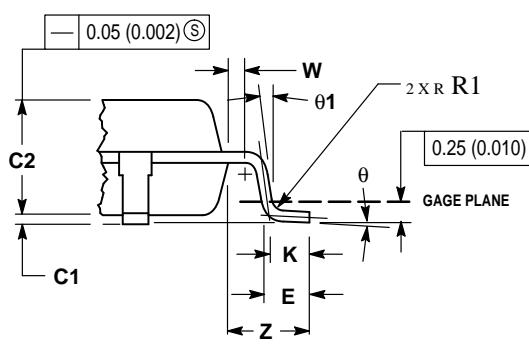
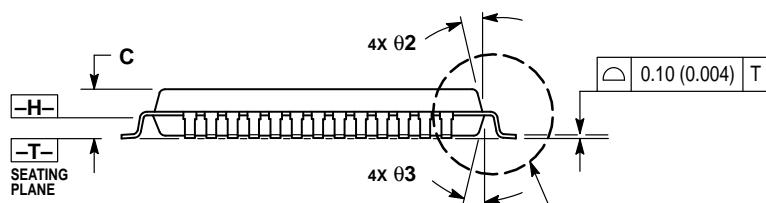
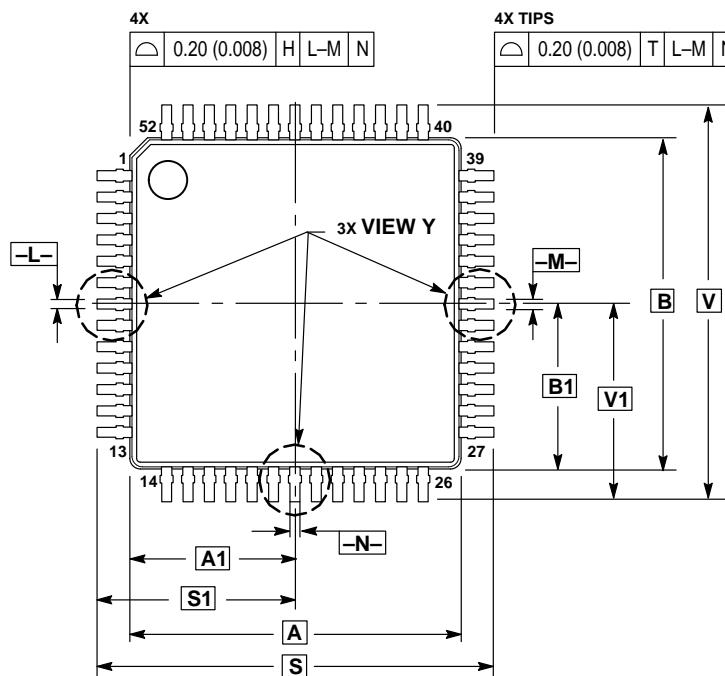
PCI_CLKs can be configured to drive processor bus clock loads in addition to the BCLKs or alternatively the clocking roles of the BCLKs and PCI_CLKs can be reversed. Table 3 illustrates some useful frequency combinations for driving PowerPC 603, PowerPC 604 or Pentium microprocessor based systems.

Table 5. Common PowerPC 603, PowerPC 604 and Pentium System Frequencies

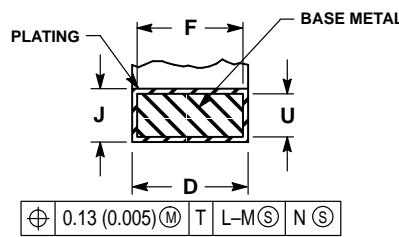
2x_PCLK	PCLK	BCLK	PCI_CLK
80	80	80(1x)	26(1/3x)
75	75	75(1x)	25(1/3x)
66	66	66(1x)	33(1/2x)
66	66	66(1x)	66(1x)
66	66	33(1/2x)	66(PCLKEN)
60	60	60(1x)	30(1/2x)

OUTLINE DIMENSIONS

FA SUFFIX
TQFP PACKAGE
CASE 848D-03
ISSUE C



VIEW Y



**SECTION AB-AB
ROTATED 90° COUNTERCLOCKWISE**

NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: MILLIMETER.
 3. DATUM PLANE $-H-$ IS LOCATED AT BOTTOM OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE BOTTOM OF THE PARTING LINE.
 4. DATUMS $-L-, -M-$ AND $-N-$ TO BE DETERMINED AT DATUM PLANE $-H-$.
 5. DIMENSIONS S AND V TO BE DETERMINED AT SEATING PLANE $-T-$.
 6. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.25 (0.010) PER SIDE. DIMENSIONS A AND B DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE $-H-$.
 7. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED 0.46 (0.018). MINIMUM SPACE BETWEEN PROTRUSION AND ADJACENT LEAD OR PROTRUSION 0.07 (0.003).

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	10.00	BSC	0.394	BSC
A1	5.00	BSC	0.197	BSC
B	10.00	BSC	0.394	BSC
B1	5.00	BSC	0.197	BSC
C	—	1.70	—	0.067
C1	0.05	0.20	0.002	0.008
C2	1.30	1.50	0.051	0.059
D	0.20	0.40	0.008	0.016
E	0.45	0.75	0.018	0.030
F	0.22	0.35	0.009	0.014
G	0.65	BSC	0.026	BSC
J	0.07	0.20	0.003	0.008
K	0.50	REF	0.020	REF
R1	0.08	0.20	0.003	0.008
S	12.00	BSC	0.472	BSC
S1	6.00	BSC	0.236	BSC
U	0.09	0.16	0.004	0.006
V	12.00	BSC	0.472	BSC
V1	6.00	BSC	0.236	BSC
W	0.20	REF	0.008	REF
Z	1.00	REF	0.039	REF
θ	0°	7°	0°	7°
Ø1	0°	—	0°	—
Ø2	12°	REF	12°	REF
Ø3	5°	13°	5°	13°

Motorola reserves the right to make changes without further notice to any products herein. Motorola makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does Motorola assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation consequential or incidental damages. "Typical" parameters can and do vary in different applications. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. Motorola does not convey any license under its patent rights nor the rights of others. Motorola products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other applications intended to support or sustain life, or for any other application in which the failure of the Motorola product could create a situation where personal injury or death may occur. Should Buyer purchase or use Motorola products for any such unintended or unauthorized application, Buyer shall indemnify and hold Motorola and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that Motorola was negligent regarding the design or manufacture of the part. Motorola and  are registered trademarks of Motorola, Inc. Motorola, Inc. is an Equal Opportunity/Affirmative Action Employer.

How to reach us:

USA/EUROPE: Motorola Literature Distribution;
P.O. Box 20912; Phoenix, Arizona 85036. 1-800-441-2447

MFAX: RMFAX0@email.sps.mot.com -TOUCHTONE (602) 244-6609
INTERNET: <http://Design-NET.com>

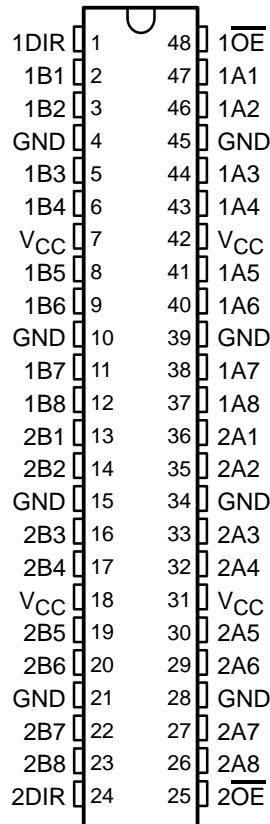
JAPAN: Nippon Motorola Ltd.; Tatsumi-SPD-JLDC, Toshikatsu Otsuki,
6F Seibu-Butsuryu-Center, 3-14-2 Tatsumi Koto-Ku, Tokyo 135, Japan. 03-3521-8315

HONG KONG: Motorola Semiconductors H.K. Ltd.; 8B Tai Ping Industrial Park,
51 Ting Kok Road, Tai Po, N.T., Hong Kong. 852-26629298



- State-of-the-Art Advanced BiCMOS Technology (ABT) Design for 3.3-V Operation and Low-Static Power Dissipation
- Members of the Texas Instruments *Widebus™* Family
- High-Impedance State During Power Up and Power Down
- Support Mixed-Mode Signal Operation (5-V Input and Output Voltages With 3.3-V V_{CC})
- Support Unregulated Battery Operation Down to 2.7 V
- Typical V_{OLP} (Output Ground Bounce) < 0.8 V at $V_{CC} = 3.3$ V, $T_A = 25^\circ\text{C}$
- ESD Protection Exceeds 2000 V Per MIL-STD-883, Method 3015; Exceeds 200 V Using Machine Model ($C = 200$ pF, $R = 0$)
- Latch-Up Performance Exceeds 500 mA Per JEDEC Standard JESD-17
- Bus-Hold Data Inputs Eliminate the Need for External Pullup/Pulldown Resistors
- Power Off Disables Inputs/Outputs, Permitting Live Insertion
- Package Options Include Plastic 300-mil Shrink Small-Outline (DL) and Thin Shrink Small-Outline (DGG) Packages and 380-mil Fine-Pitch Ceramic Flat (WD) Packages Using 25-mil Center-to-Center Spacings

**SN54LVTH16245A . . . WD PACKAGE
SN74LVTH16245A . . . DGG OR DL PACKAGE
(TOP VIEW)**



description

The 'LVTH16245A are 16-bit (dual-octal) noninverting 3-state transceivers designed for low-voltage (3.3-V) V_{CC} operation, but with the capability to provide a TTL interface to a 5-V system environment.

These devices can be used as two 8-bit transceivers or one 16-bit transceiver. They allow data transmission from the A bus to the B bus or from the B bus to the A bus, depending on the logic level at the direction-control (DIR) input. The output-enable (OE) input can be used to disable the devices so that the buses are effectively isolated.

Active bus-hold circuitry is provided to hold unused or floating data inputs at a valid logic level.

When V_{CC} is between 0 and 1.5-V, the devices are in the high-impedance state during power up or power down. However, to ensure the high-impedance state above 1.5-V, OE should be tied to V_{CC} through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

Widebus is a trademark of Texas Instruments Incorporated.

UNLESS OTHERWISE NOTED this document contains PRODUCTION DATA information current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



Copyright © 1996, Texas Instruments Incorporated

SN54LVTH16245A, SN74LVTH16245A 3.3-V ABT 16-BIT BUS TRANSCEIVERS WITH 3-STATE OUTPUTS

SCBS143F – MAY 1992 – REVISED AUGUST 1996

description (continued)

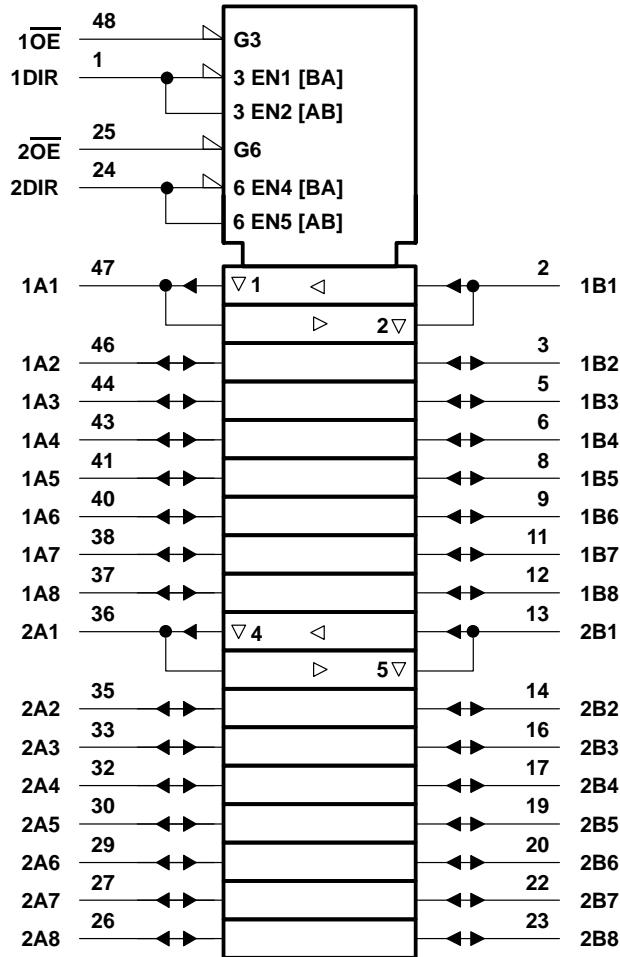
The SN74LVTH16245A is available in TI's shrink small-outline (DL) and thin shrink small-outline (DGG) packages, which provide twice the I/O pin count and functionality of standard small-outline packages in the same printed-circuit-board area.

The SN54LVTH16245A is characterized for operation over the full military temperature range of -55°C to 125°C . The SN74LVTH16245A is characterized for operation from -40°C to 85°C .

FUNCTION TABLE
(each 8-bit section)

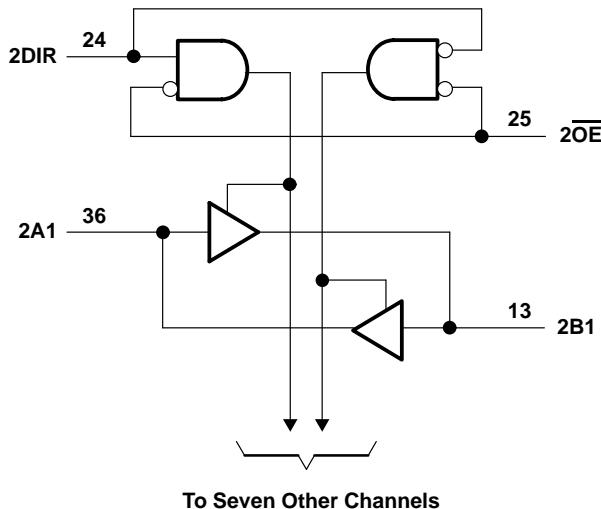
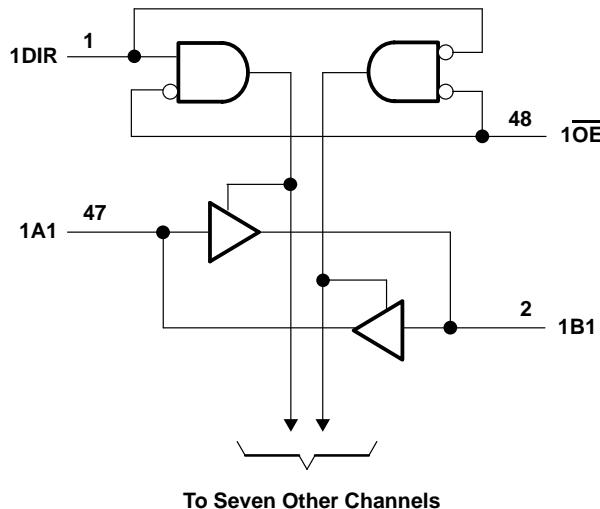
INPUTS		OPERATION
OE	DIR	
L	L	B data to A bus
L	H	A data to B bus
H	X	Isolation

logic symbol†



† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

logic diagram (positive logic)



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage range, V_{CC}	-0.5 V to 4.6 V	
Input voltage range, V_I (see Note 1)	-0.5 V to 7 V	
Voltage range applied to any output in the high state or power-off state, V_O (see Note 1)	-0.5 V to 7 V	
Current into any output in the low state, I_O : SN54LVTH16245A	96 mA	
	SN74LVTH16245A	128 mA
Current into any output in the high state, I_O (see Note 2): SN54LVTH16245A	48 mA	
	SN74LVTH16245A	64 mA
Input clamp current, I_{IK} ($V_I < 0$)	-50 mA	
Output clamp current, I_{OK} ($V_O < 0$)	-50 mA	
Maximum power dissipation at $T_A = 55^\circ\text{C}$ (in still air) (see Note 3): DGG package	0.85 W	
	DL package	1.2 W
Storage temperature range, T_{stg}	-65°C to 150°C	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output clamp-current ratings are observed.
2. This current flows only when the output is in the high state and $V_O > V_{CC}$.
3. The maximum package power dissipation is calculated using a junction temperature of 150°C and a board trace length of 750 mils. For more information, refer to the *Package Thermal Considerations* application note in the *ABT Advanced BiCMOS Technology Data Book*.

**SN54LVTH16245A, SN74LVTH16245A
3.3-V ABT 16-BIT BUS TRANSCEIVERS
WITH 3-STATE OUTPUTS**

SCBS143F – MAY 1992 – REVISED AUGUST 1996

recommended operating conditions (see Note 4)

		SN54LVTH16245A		SN74LVTH16245A		UNIT
		MIN	MAX	MIN	MAX	
V _{CC}	Supply voltage	2.7	3.6	2.7	3.6	V
V _{IH}	High-level input voltage	2		2		V
V _{IL}	Low-level input voltage		0.8		0.8	V
V _I	Input voltage		5.5		5.5	V
I _{OH}	High-level output current		-24		-32	mA
I _{OL}	Low-level output current		48		64	mA
Δt/Δv	Input transition rise or fall rate	Outputs enabled		10	10	ns/V
Δt/ΔV _{CC}	Power-up ramp rate			200	200	μs/V
T _A	Operating free-air temperature	-55	125	-40	85	°C

NOTE 4: Unused control inputs must be held high or low to prevent them from floating.

PRODUCT PREVIEW information concerns products in the formative or design phase of development. Characteristic data and other specifications are design goals. Texas Instruments reserves the right to change or discontinue these products without notice.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

SN54LVTH16245A, SN74LVTH16245A
3.3-V ABT 16-BIT BUS TRANSCEIVERS
WITH 3-STATE OUTPUTS

SCBS143F – MAY 1992 – REVISED AUGUST 1996

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS			SN54LVTH16245A	SN74LVTH16245A	UNIT
				MIN	TYP†	
V _{IK}	V _{CC} = 2.7 V,	I _I = -18 mA		-1.2		-1.2
V _{OH}	V _{CC} = MIN to MAX‡,	I _{OH} = -100 µA		V _{CC} -0.2	V _{CC} -0.2	V
	V _{CC} = 2.7 V,	I _{OH} = -8 mA		2.4	2.4	
	V _{CC} = 3 V	I _{OH} = -24 mA		2		
		I _{OH} = -32 mA			2	
V _{OL}	V _{CC} = 2.7 V	I _{OL} = 100 µA		0.2	0.2	V
		I _{OL} = 24 mA		0.5	0.5	
	V _{CC} = 3 V	I _{OL} = 16 mA		0.4	0.4	
		I _{OL} = 32 mA		0.5	0.5	
		I _{OL} = 48 mA		0.55		
		I _{OL} = 64 mA			0.55	
I _I	V _{CC} = 3.6 V,	V _I = V _{CC} or GND	Control inputs	±1	±1	µA
	V _{CC} = 0 or MAX‡,	V _I = 5.5 V		10	10	
	V _{CC} = 3.6 V	V _I = 5.5 V	A or B ports§	20	20	
		V _I = V _{CC}		1	1	
		V _I = 0		-5	-5	
I _{off}	V _{CC} = 0,	V _I or V _O = 0 to 4.5 V			±100	µA
I _{I(hold)}	V _{CC} = 3 V	V _I = 0.8 V	A or B ports	75	75	µA
		V _I = 2 V		-75	-75	
I _{OZPU} ¶	V _{CC} = 0 to 1.5 V,	V _O = 0.5 V to 3 V, $\overline{OE} = 0$		±100	±100	µA
I _{OZPD} ¶	V _{CC} = 0 to 1.5 V,	V _O = 0.5 V to 3 V, $\overline{OE} = 0$		±100	±100	µA
I _{CC} #	V _{CC} = 3.6 V, I _O = 0, V _I = V _{CC} or GND	Outputs high		0.19	0.19	mA
		Outputs low		5	5	
		Outputs disabled		0.19	0.19	
ΔI _{CC}	V _{CC} = 3 V to 3.6, One input at V _{CC} - 0.6 V, Other inputs at V _{CC} or GND			0.3	0.2	mA
C _i	V _I = 3 V or 0			4	4	pF
C _{io}	V _O = 3 V or 0			10	10	pF

† All typical values are at V_{CC} = 3.3 V, T_A = 25°C.

‡ For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

§ Unused pins at V_{CC} or GND

¶ This parameter is warranted by characterization but not production tested.

This is the increase in supply current for each input that is at the specified TTL voltage level rather than V_{CC} or GND.

PRODUCT PREVIEW information concerns products in the formative or design phase of development. Characteristic data and other specifications are design goals. Texas Instruments reserves the right to change or discontinue these products without notice.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**SN54LVTH16245A, SN74LVTH16245A
3.3-V ABT 16-BIT BUS TRANSCEIVERS
WITH 3-STATE OUTPUTS**

SCBS143F – MAY 1992 – REVISED AUGUST 1996

switching characteristics over recommended operating free-air temperature range, $C_L = 50 \text{ pF}$ (unless otherwise noted) (see Figure 1)

PARAMETER	FROM (INPUT)	TO (OUTPUT)	SN54LVTH16245A		SN74LVTH16245A				UNIT	
			$V_{CC} = 3.3 \text{ V} \pm 0.3 \text{ V}$		$V_{CC} = 2.7 \text{ V}$		$V_{CC} = 3.3 \text{ V} \pm 0.3 \text{ V}$			
			MIN	MAX	MIN	MAX	MIN	TYP†	MAX	
t_{PLH}	A or B	B or A	1.5	3.5	3.9	1.5	2.3	3.3	3.7	ns
t_{PHL}			1.3	3.5	3.7	1.3	2.1	3.3	3.5	
t_{PZH}	\overline{OE}	A or B	1.5	4.8	5.6	1.5	2.8	4.5	5.3	ns
t_{PZL}			1.6	4.8	5.4	1.6	2.9	4.6	5.2	
t_{PHZ}	\overline{OE}	A or B	2.3	5.5	5.8	2.3	3.7	5.1	5.5	ns
t_{PLZ}			2.2	5.5	5.7	2.2	3.5	5.1	5.4	

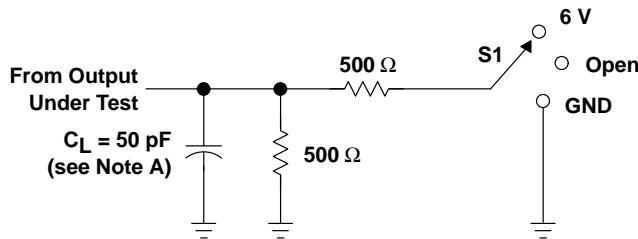
† All typical values are at $V_{CC} = 3.3 \text{ V}$, $T_A = 25^\circ\text{C}$.

PRODUCT PREVIEW information concerns products in the formative or design phase of development. Characteristic data and other specifications are design goals. Texas Instruments reserves the right to change or discontinue these products without notice.



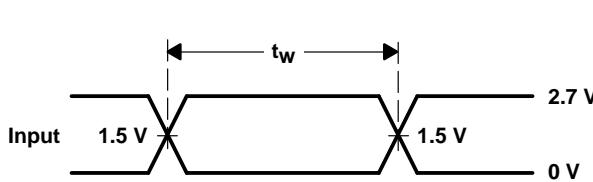
POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

PARAMETER MEASUREMENT INFORMATION

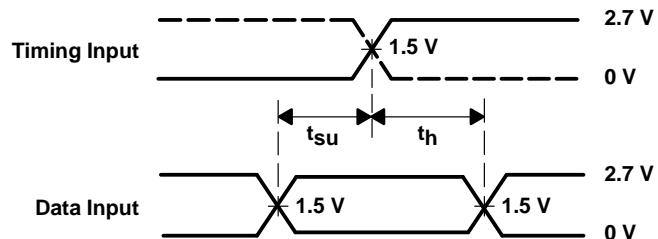


TEST	S1
tPLH/tPHL	Open
tPLZ/tPZL	6 V
tPHZ/tPZH	GND

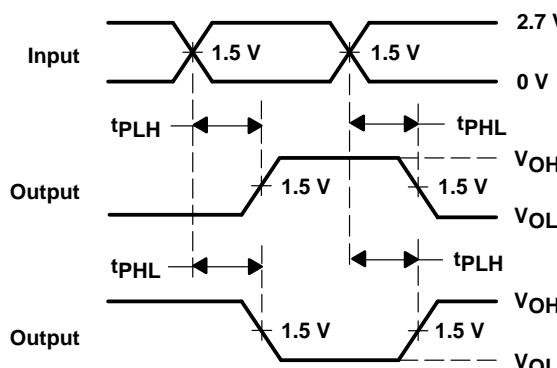
LOAD CIRCUIT



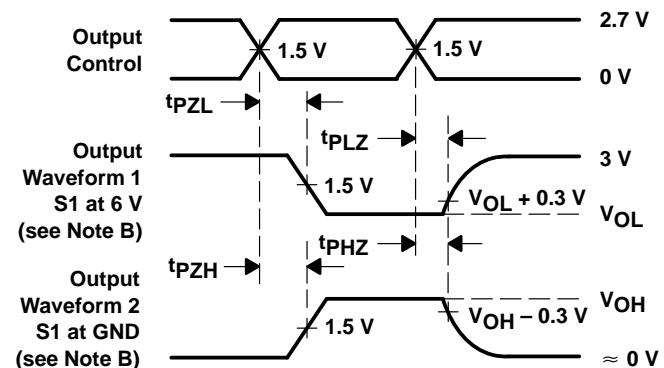
VOLTAGE WAVEFORMS
 PULSE DURATION



VOLTAGE WAVEFORMS
 SETUP AND HOLD TIMES



VOLTAGE WAVEFORMS
 PROPAGATION DELAY TIMES
 INVERTING AND NONINVERTING OUTPUTS



VOLTAGE WAVEFORMS
 ENABLE AND DISABLE TIMES
 LOW- AND HIGH-LEVEL ENABLING

- NOTES:
- A. C_L includes probe and jig capacitance.
 - B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
 - C. All input pulses are supplied by generators having the following characteristics: $PRR \leq 10 \text{ MHz}$, $Z_O = 50 \Omega$, $t_r \leq 2.5 \text{ ns}$, $t_f \leq 2.5 \text{ ns}$.
 - D. The outputs are measured one at a time with one transition per measurement.

Figure 1. Load Circuit and Voltage Waveforms

IMPORTANT NOTICE

Texas Instruments (TI) reserves the right to make changes to its products or to discontinue any semiconductor product or service without notice, and advises its customers to obtain the latest version of relevant information to verify, before placing orders, that the information being relied on is current.

TI warrants performance of its semiconductor products and related software to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

Certain applications using semiconductor products may involve potential risks of death, personal injury, or severe property or environmental damage ("Critical Applications").

TI SEMICONDUCTOR PRODUCTS ARE NOT DESIGNED, INTENDED, AUTHORIZED, OR WARRANTED TO BE SUITABLE FOR USE IN LIFE-SUPPORT APPLICATIONS, DEVICES OR SYSTEMS OR OTHER CRITICAL APPLICATIONS.

Inclusion of TI products in such applications is understood to be fully at the risk of the customer. Use of TI products in such applications requires the written approval of an appropriate TI officer. Questions concerning potential risk applications should be directed to TI through a local SC sales office.

In order to minimize risks associated with the customer's applications, adequate design and operating safeguards should be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance, customer product design, software performance, or infringement of patents or services described herein. Nor does TI warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such semiconductor products or services might be or are used.